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THE BOOSTER-TO-AGS BEAM TRANSFER FAST KICKER MODULATORS¹

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ABSTRACT

This work describes modulator developments for the Brookhaven Booster extraction and the AGS injection fast kickers. The modulators are projected for both proton and heavy ion operation. The equivalent load inductance is about 2.1 to 2.3 μH for each modulator. The PFN voltage is required to be below 40 kVfor operation in air. The rise time of the pulse for proton beam transfer is 120 ns up to 97% of full current (1000 A), and for heavy ion beam transfer, the requirement is 160 ns up to 98% of full current (1615 A). During the fourth batch transfer of the proton beam from the Booster to AGS, the pulse fall time of the AGS injection fast kicker has to be very fast (< 140 ns), so that it does not appreciably deflect the first batch of injected protons that is circulating in the AGS. To achieve the design specifications, an extensive development effort has been pursued, including distributed parameter estimation and measurement, computer aided analysis and design, pulse shaping and tail-biting circuit test, proto-type construction, etc. The test results will be presented.

INTRODUCTION

The Brookhaven AGS Booster will serve as a multifunction synchrotron injector for the AGS, capable of accelerating protons from 200 Mev, the Linac operating energy, to 1.5 Gev, at a maximum repetition rate of 7.5 Hz (4 pulses/AGS pulse). The Booster is also capable of accelerating heavy ions to a magnetic rigidity equal to 17.52 Tesla-meters, at < 1 Hz repetition rate (1 pulse/AGS pulse). Beam transfer from the Booster to the AGS will be bucket to bucket. There are three RF accelerating buckets in the Booster, and twelve in the AGS. During each pulse of proton beam transfer, three Booster proton bunches will be transferred to three of the twelve AGS buckets.

The AGS injection and Booster extraction kickers are the complementary system to each other.

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Centrally-fed picture frame lumped inductance ferrite magnets are used for both kickers. The kicker parameters are listed in Table 1 and 2. The Booster magnet must be capable of being baked up to temperatures of 250 °C, since the Booster operating vacuum is in the 10^{-11} Torr range.

	Proton	Heavy Ion
Rigidity	7.51 $T-m$	11 T - m
Strength	3 mrad	3 mrad
Rise time	120 ns (to 97%)	160 ns (to 98%)
Pulse length	600 ns	1000 ns
Fall time	140 ns	<2300 ns
Pulse overshoot	3%	2%
Flat top ripple	<3%	$<\!2\%$
PFN voltage	30 kV	$42 \ kV$

Table 1 - AGS INJECTION KICKER PARAMETERS

Table 2 - BOOSTER EXTRACTION KICKER PARAMETERS

1006 A

Peak current

1473 A

	Proton	Heavy Ion
Rigidity	7.51 $T-m$	$11 \ T - m$
Strength	5 mrad	3.8 mrad
Rise time	120 ns (to 97%)	160 ns (to 98%)
Pulse length	>600 ns	>1000 ns
Pulse overshoot	3%	2%
Flat top ripple	<3%	$<\!2\%$
PFN voltage	30 kV	40 kV
Peak current	1000 A	1615 A

The AGS injection kicker consists of three lumped magnet sections with equal inductance, which will be driven by three identical pulsers. The Booster extraction kicker has a similar arrangement of four identical subsystems. The load and loop stray inductance is about 2.1 μH for each AGS injection subsystem and 2.35 μH for each Booster extraction subsystem.

¹ Work performed under the auspices of the U.S. Department of Energy.

SYSTEM DESCRIPTION

The modulators are basically E-type pulse forming networks with equal capacitance and equal inductance for each section. The PFN voltage is required to be about 40 kV or below for operation in air. To meet the different kicker current and rise time requirements of proton and heavy ion operation, a terminating resistor is switched to match and mismatch the pulse forming network impedance.

The pulse front edge sharpening is accomplished by using two R-C compensation networks. The $R_a - C_a$ in parallel with the first PFN capacitor is an energy compensation network. It provides additional energy needed to build up the current in the load magnet during the later portion of the pulse rise period. The proper selection of $R_a - C_a$ will give a fast rising front edge current with an acceptable overshoot or even without an overshoot.

A sink network $R_b - C_b$ is used in parallel with the matching resistor. The capacitance C_b together with the magnet inductance constitute a resonant network that reduces the resistance to the energy discharging of the PFN front edge capacitor during the pulse rising period. The damping resistor R_b is used to avoid possible oscillation in the L-C network, and its value should be much smaller than that of the impedance matching resistor. During the flat top of the pulse, both C_a and C_b hold at a constant voltage, therefore the pulse flat top is not affected. Figure 1 (a) shows the current waveform of the proto-module of the Booster extraction kicker, and (b) the magnetic field waveform of the proto-magnet. The Booster extraction kicker modulators have been constructed, and are presently in the installation process.

In addition to the fast rise time and low ripple flat top, the pulse fall time of the AGS injection kicker has to be less than 140 ns, for proton injection. A test pulser, based on a circuit improved from a schematic



Figure 1 (a). Current pulse waveform.



Figure 1 (b). Magnetic field waveform.



Figure 3. AGS injection fast kicker modulator.

in [2], has been simulated, constructed and tested. The EEV CX-1154 thyratrons were used as the main switch and tail-bitting switch. This test unit has been powered up to $25 \ kV$, and about 900 A including the tail-bitting function. Figure 2 shows a typical current pulse fall time obtained in this test unit. The construction unit will use two gap thyratrons to increase voltage hold-off capability. Figure 3 is the schematic diagram of the AGS injection fast kicker modulator.

The three modules of the AGS injection fast kicker are being built. Figure 4 shows a current waveform from one of the modules with fast fall time. However, the unit was tested only up to 500 A, due to noise problems. Since the modulators will be mounted inside the AGS ring, limited space makes it very difficult for package design. The width of each modulator is only 1 foot. The present package design eliminated all thyratron shieldings, which may result in lowering the noise immunity of the floating deck structure of the tail-bitting thyratron. Other factors such as grounding design and trigger scheme may also contribute to the noise problems. Some design changes including an auxiliary trigger circuit of the tail-bitting thyratron, grounding, packaging, and the possible interchange of circuit locations of the main and tailbitting sections are being considered. Testing will resume soon after the Booster installation is completed.

Computer aided design and analysis have been used heavily for the modulator development. The pulse distortion due to distributed capacitance and inductance has been investigated by computer simulation and actual test. The test results and simulations have been agreed very well. Some results are included in references [3, 5]. MICRO-CAP (I,II,III) and $MATRIX_X$ are used for computer simulation².

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Figure 2. Current palse falling edge with tail-bitting.



Figure 4. Current pulse waveform with tail-bitting.

 $^{^2}$ MICRO-CAP is a product of Spectrum Software, Inc., Sunnyvale, CA 94086. $MA\,TRIX_X$ is a trademark of Integrated Systems Inc., Santa Clara, CA 95054.