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# Estimation of the Mode Losses in Complex RF Vacuum Devices

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#### Abstract

It is mandatory to reduce the mode losses in superconducting beam-tube structures due to the high costs of low temperature cooling. For the analysis of the HERA beam-position pick-up, as a complex three-dimensional rf vacuum device, the new 3D MAFIA code is used. Extensive checks with the 2D URMEL code as well as bench measurements in the frequency domain completed the estimation of the mode losses.

### Introduction

The beam position in the HERA proton storage ring will be measured by about 250 directional-coupler pick-up stations [1]. They are build into the 4 K cryostat of the superconducting magnets. Due to the high cost of the low temperature cooling it is important to minimize the introduced heat. For the HERA pick-up the losses have to be kept in the subwatt range. A considerable fraction of the heat losses is due to the so called mode losses, caused by excitation of electromagnetic field resonances in the pick-up by the beam<sup>1</sup>. This paper is concerned with the determination of these resonances.

The losses can be determinated in two complementary ways, in the time and in the frequency domain. Here the computation is done in the frequency domain, because wakefield effects can be estimated more easily and only a small number of resonances have to be considered<sup>2</sup>. The measurement is also done in the frequency domain, because this method determines each resonance separately with a high precision<sup>3</sup>.

### **Power Losses of Resonance Modes**

A particle beam excites resonance modes in a deformed section of the vacuum chamber. Number, type and quality factor of the possible modes are characteristic properties of the specific rf vacuum device. The excitation strength from a centered beam<sup>4</sup> is defined by the bunch spectrum.

Only a fraction of the modes has to be considered. In a cylindrically symmetrical device these are the TM monopole modes. Furthermore we used all relevant modes up to the beam pipe cutoff frequency, which is well above the 3 dB frequency of our shortest bunch.

Next we describe the relation between the normalized power loss and the device parameters for a single mode. The normalized voltage across the RF-structure is given by [3]:

$$v_{n,\lambda}(t) \approx 2 q_b k_{n,\lambda} \exp(-\frac{t}{Q_\lambda/\omega_\lambda}) \cos(\omega_\lambda t)$$
 (1)

Here  $\omega_{\lambda}$  is the resonance frequency,  $Q_{\lambda}$  the quality factor and  $k_{n,\lambda} = (\omega_{\lambda}/2) (R_{\lambda}/Q_{\lambda})$  the normalized loss factor – with the shunt impedance  $R_{\lambda}$  - of an excited mode. Finally  $q_b$  is the test charge of a "dirac bunch".

For a bunch train with a time spacing  $T_b$  the voltages add up to

$$V_{n,\lambda} = 2 q_b k_{n,\lambda} \frac{1 - \exp(-\tau_\lambda) \cos(\omega_\lambda T_b)}{1 - 2 \exp(-\tau_\lambda) \cos(\omega_\lambda T_b) + \exp(-2\tau_\lambda)}, \qquad (2)$$

<sup>1</sup>The effect of the resonance modes on the stability of the beam was not our major concern.

<sup>3</sup>With stimulating pulse techniqes a resolution of only several watts was obtained [2].
 <sup>4</sup>This simplification is valid for mode loss estimations.

	URMEL	MAFIA <sup>a</sup>
CPU time	13min	24 h
elapse time	10 15 min	2030h
memory	2 M B	4 MB
meshpoints <sup>b</sup>	4000	30000

"In the MAFIA code set the E31 eigenvalue finder is used.  $R_{\lambda}$  and  $R_{\lambda}/Q_{\lambda}$  are computed interactively with P3.

For URMEL problems we used a fourfold (except example No. IV) and for MAFIA an eightfold symmetry.

Table 1: Typical program data and computing performance

with 
$$\tau_{\lambda} \approx \frac{T_b}{Q_{\lambda}/\omega_{\lambda}}$$
.

The  $\omega_{\lambda}T_{b}$  term describes the phase between the bunch under test and the wakefields of the previous bunches. We get an upper limit of the mode losses by setting  $\cos(\omega_{\lambda}T_b) = 1$ :

$$V_{n,\lambda_{max}} = 2 q_b k_{n,\lambda} \frac{1}{1 - \exp(-\tau_\lambda)}$$
(3)

For the power loss per mode we get:

$$P_{n,\lambda_{max}} = \frac{1}{2} V_{n,\lambda_{max}} I_0 , \text{ with } I_0 = \frac{q_b}{T_b}$$
$$= \frac{I_0^2}{2} R_\lambda \frac{\tau_\lambda}{1 - \exp(-\tau_\lambda)}$$
$$\approx \frac{I_0^2}{2} R_\lambda , \text{ for small } \tau_\lambda \qquad (4)$$

The total power loss is given by the sum of the losses in every single mode n, where the normalized bunch spectrum  $\rho$  has to be taken into account:

$$P_{tot_{max}} = \sum_{\lambda} P_{n,\lambda_{max}} \cdot \rho^2(\omega_{\lambda}) \tag{5}$$

## **Computations for a** practical RF Vacuum Device

The main task of the computation is to find all the relevant resonance modes of the structure. This is done by discretising the Maxwell equations and solving the eigenvalue problem. For cylinder symmetrical examples we used the URMEL code [4] and for three-dimensional ones the more general MAFIA codes [5].

The programs supply the modes with their resonance frequencies and electromagnetic fields. The  $Q_{\lambda}$  and  $R_{\lambda}/Q_{\lambda}$  values are computed from the field vectors. In the URMEL code this is done in a canonical way, in the MAFIA codes an integration path has to be specified -which is in our cases the symmetry axis.

Both codes run on the DESY central computer (IBM 3084-Q). Table 1 gives an impression of the typical program data of the examples under discussion and the computing performance. Practical experience shows that the eigenvalue finder SAP [7] of the codes is very sensitive to its main input parameter: the maximum resonance frequency of interest. Furthermore, closely spaced substructures lead to

We consider only resonances below the beam pipe cutoff frequency.

problems, which have to be tackled by fine tuning of the SAP parameters.

Next we explore the influence of geometry variations on the mode losses of the beam-position pick-up for the HERA proton storage ring [1]. For all comparisons a rather extreme HERA operation condition was chosen: a beam current of 0.163 A and a bunch length of 0.3 m with a parabolic shape in phase space<sup>5</sup> [6]. As a pessimistic approximation we used the envelope of the corresponding bunch spectrum  $\rho(\omega_{\lambda})$ .

An inspection of different pick-up body structures (geometries I, II and III of Table 2) shows that a smooth – even though short – tapering reduces the mode losses by a factor up to two. The losses vary substantially for different antenna suspensions (Tab. 2, geometries IV, V and VI), where the antennas always enhance the dissipated power (II). The power losses are particularly high for structures with shorted antennas (geometry IV), due to a *coaxial transmission-line effect*. The best solution is simply hanging the antennas into the straight beam pipe (IV). For the HERA proton storage ring this is not possible due to the unavoidable reduction in acceptance. Therefore we used an antenna suspension in a beam pipe indentation (VI) with reasonably low power losses compared to the no antenna case (II).

The URMEL code, used for these cylinder symmetrical calculations, took neither our finite antenna width nor the complex shaped pick-up body structure into account<sup>6</sup>. In order to minimize the mode losses we used the smallest mechanically tolerable width. For the HERA beamposition pick-up (Tab. 2, geometry VIII) the maximum power loss is estimated to be:

$$P_{ioi_{max}} \approx 0.18 W$$

### Detailed Comparisons of Power Loss Estimation Methods

For the relative simple rf vacuum structure II we compared the results of the different power loss estimation methods in detail. Twodimensional – with the URMEL code – and three-dimensional calculations – with the MAFIA codes – as well as measurements for all relevant modes were performed (see Table 3).

The measurement of the resonance frequencies  $\omega_{\lambda}$  and the corresponding Q-values  $Q_{\lambda}$  were done with the transmission method [8]. The shunt impedance  $R_{\lambda}$  is related in a simple way to the electric field component  $E_{z_{\lambda}}$  along the axis of interest:  $R_{\lambda}/Q_{\lambda} = 1/2 \left( \int (E_{z_{\lambda}}/\sqrt{P_{\lambda}} Q_{\lambda}) dz \right)^2$ , where  $P_{\lambda}$  is the power loss of the mode. The term  $E_{z_{\lambda}}/\sqrt{P_{\lambda}} Q_{\lambda}$  is measured with the frequency shift method [9], and we get:

$$\frac{R_{\lambda}}{Q_{\lambda}} = \frac{2}{\omega_{p,\lambda} \epsilon c_E} \left[ \left( \int_0^\ell \sqrt{\delta} \cos \frac{z \, \omega_{p,\lambda}}{c} \, dz \right)^2 + \left( \int_0^\ell \sqrt{\delta} \sin \frac{z \, \omega_{p,\lambda}}{c} \, dz \right)^2 \right]$$

 $\delta = \frac{\omega_{\mathbf{p},\lambda} - \omega_{\lambda}}{\omega_{\lambda}}$  relative frequency shift

 $c_E$  shape factor of the perturbing object

- $\epsilon$  dielectric constant of the perturbing object
- $\omega_{\lambda}$  frequency of the unperturbed resonance mode
- $\omega_{p,\lambda}$  frequency of the perturbed resonance mode

### Conclusion

In this paper we have presented frequency domain methods to compute and measure the power loss of vacuum devices. Mode losses in the *subwatt* range can be estimated. These methods can be applied to structures of almost any shape.

	<u>ہ</u>	$f_{\lambda}[MHz]$	$Q_{\lambda}$	$R_\lambda/Q_\lambda\left[\Omega ight]$
	1	3059 8 3044 8 3072 c	29895 31182 19820	0.0859 0.1114 0.2
:	2	\$101 \$086 \$110	29872 31134 23930	0.0680 0.0477 0.7
:	3	\$169 \$152 \$179	29897 31077 23560	0. <b>33</b> 04 0.4520 0.2
	4	3265 3248 8277	\$0029 \$1171 2\$410	0.8859 0.94 <b>32</b> 1.5
	5	\$387 \$370 \$397	\$0297 \$1445 17880	0.0724 0.0621 0.1
	6	8581 3515 3543	30702 31874 21480	0.5309 0.7157 0.6
,	7	3896 3680 3708	31223 32446 20050	1.56 <b>33</b> 1.88 <b>35</b> 1.6
	8	3878 3863 3891	31802 33121 19470	1.2152 1.1626 1.6
1	9	4070 4055 4085	\$2017 \$3600 13620	0.4529 0.1808 0.7

2D	cal	cul	atio	n
• 9 Th	1		- + ÷ -	

'rough measurement

Table 3: Comparison of the different estimation methods (results for geometry II of Tab. 2)

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<sup>&</sup>lt;sup>5</sup> In real space the shape has the form:  $i_{norm}(t) = 16 c \left[1 - (2 c t/\ell)^2\right]^{2/2} / (3 \pi \ell)$ <sup>6</sup> Before the advent of 3D codes it was customary to take the width into account by scaling the cylindrically symmetrical result by the ratio of antenna-width to circumfence. For the Hera beam-position pick-up this crude method results in  $P_{iot,max} \approx 0.7 W$ .



Table 2: Comparison of computed rf vacuum structures