HIGH ACCURACY ANALYSIS OF ARBITRARY MODES IN TAPERED DISK-LOADED STRUCTURES

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Abstract

In this paper, a high-precision eigenmode-computation analysis of arbitrary azimuthal mode numbers in periodic disc-loaded structure based on variational method will be discussed. It allows for rounding the edge of a disk hole without any approximation in shape treatment and calculates the modes exactly synchronous to the beam. It converges much faster than the mesh-based computer code. Good agreement was observed between the results of variational method and those of other methods.

1 INTRODUCTION

In the past, various numerical methods were developed for the analysis of the higher order modes in a cell[1-4], KN7C[5], TRANSVRS[6], URMEL[7], MAFIA[8], and Ω 2[9], SUPERFISH[10]. The codes KN7C and TRANSVRS can calculate the longitudinal and transverse modes with any phase advance per period, though imposing two approximations: an infinite periodic structure and a flat beam hole surface parallel to the beam axis instead of round edge. Because of the latter assumption, the calculated wake field is not accurate enough[1]. Though these two codes have been very successfully applied to the actual design of structures, they need to be further improved if they are to be used for cells with a round edge beam hole. This is one of the motives for developing the present code.

On the other hand, such codes as SUPERFISH, URMEL, Ω 2 and MAFIA can represent the structure by filling with a mesh so that these can calculate the modes in cells of any shape. However, our method still has its advantages in many aspects. The variational method can the structure geometry without approximation. Therefore it is a better way in the geometry treatment than various mesh-based methods. The fields in the variational method are given as a series expansion, which derives from the Maxwell equations. The calculated fields converge on the true ones very fast when the number of terms increases. Also, it uses less memory than the mesh-based method. The accuracy depends on the terms number used. The frequency accuracy is around 10⁻⁷ for the acceleration mode of our X-band structure with a number of terms more than 55. One can get a sufficiently accurate calculation by increasing the number of terms. Our method can be used to calculate both monopole modes and all azimuthally varying modes with any phase shift per cell, which means that it can calculate all kinds of field in the disk-loaded

structure (any azimuthal mode, any synchrotron phase, any frequency or passpand). The fields are expressed as simple formula. Therefore it can be conveniently used to serve as input for other purpose calculation, such as the base vectors for equivalent circuit model or open-mode expansion model [11].

The present code is named LONGTRANSVRS.

2 THEORY

The theory of variational approach is a well-known method and is basically a mathematical forerunner of the finite element method. In the classical formulation the problem is to find the unknown function or functions that extremize (maximize, minimize) or get stationary under the specified boundary condition. The fields with steady-

state sinusoidal time dependence of $e^{j\omega t}$ in vacuum space can be found by making the following value minimum

$$J = jk \int (Z_0 \mathbf{H})^* \times \mathbf{E} \cdot \mathbf{n} dS. \tag{1}$$

where $k = \omega \sqrt{\varepsilon_0 \mu_0}$ is the propagation constant, **n** is the unit vector outward normal to the surface *S*. The variational form eq.(1) has no further limitation on the trial function on the metal boundary, but the non-metal boundary condition should be satisfied by the trial function.

The accelerator structure with which we are concerned is a conventional disk-loaded cylindrical waveguide as shown in Figure 1. The whole acceleration structure is divided into the inner and outer regions, separated at the common boundary at $r=r_c$. The inner region is characterized by the fact that it has a rounded edge of the disk hole as part of boundary surface.

The Hertz vectors $\Pi_h=\hat{z}\Pi_{h,z}$ and $\Pi_e=\hat{z}\Pi_{e,z}$ are chosen to simplify the solution. In the inner traveling wave region, the Hertz vectors $\Pi_{h,z}$ and $\Pi_{e,z}$ take the following forms

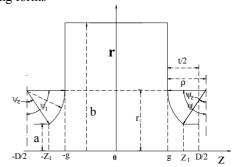


Figure 1 Cross-section of disk-loaded waveguide.

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$$\prod_{e,z}^{in} = -\sum_{n=-\infty}^{+\infty} \frac{A_n}{\chi_n^2} \frac{J_m(\chi_n r)}{J_m(\chi_n r_c)} e^{-j\beta_n z} \cos m\theta , \qquad (2)$$

$$\Pi_{h,z}^{in} = -\frac{1}{Z_0} \sum_{n=-\infty}^{+\infty} \frac{B_n}{\chi_n^2} \frac{J_m(\chi_n r)}{J_m(\chi_n r_c)} e^{-j\beta_n z} \sin m\theta , \qquad (3)$$

where $\chi_n^2 = k^2 - \beta_n^2$, $\beta_n = \beta_0 + 2\pi m/D$, $\beta_0 = \phi_0/D$. J_m is the first kind Bessel function of order m. When χ_n becomes imaginary, the regular Bessel functions can be replaced by modified Bessel functions I_m of a real argument. The above Hertz vectors fulfill the Floquet condition that is required by the trial function in eq. (1).

In the outer standing wave region, we have

$$\prod_{e,z}^{out} = -\sum_{s=0}^{+\infty} \frac{C_s}{\Gamma_s^2} \frac{R_m(\Gamma_s r)}{R_m(\Gamma_s r_c)} \cos \alpha_s (z+g) \cos m\theta \,. \tag{4}$$

$$\prod_{h,z}^{out} = j \frac{1}{Z_0} \sum_{s=0}^{+\infty} \frac{D_s}{\Gamma_s^2} \frac{S_m(\Gamma_s r)}{S_m(\Gamma_s r_c)} \sin \alpha_s (z+g) \sin m\theta, \quad (5)$$

where the transverse propagation constant $\Gamma_s^2 = k^2 - \alpha_s^2$ with $\alpha_s = \pi s/2g$ and

$$R_m(\Gamma r) = J_m(\Gamma r)Y_m(\Gamma b) - J_m(\Gamma b)Y_m(\Gamma r), \qquad (6)$$

$$S_m(\Gamma r) = J_m(\Gamma r) Y_m'(\Gamma b) - J_m'(\Gamma b) Y_m(\Gamma r). \tag{7}$$

Here, Y_m is the second kind regular Bessel function of order m. The regular Bessel functions J_m and Y_m should be replaced by modified Bessel functions I_m and K_m of a real argument when Γ_s is imaginary.

The field components, which are used as the trial functions, can be found from the Hertz vectors (2-5) as follows

$$\mathbf{E} = \nabla \times \nabla \times \mathbf{\Pi}_{e} - j\omega \mu_{0} \nabla \times \mathbf{\Pi}_{h}, \tag{8}$$

$$\mathbf{H} = \nabla \times \nabla \times \mathbf{\Pi}_{h} + j \omega \varepsilon_{0} \nabla \times \mathbf{\Pi}_{a} . \tag{9}$$

Fields given by eqs.(2-3) don't satisfy the metal boundary condition. However, fields given by eqs.(4-5) do. The metal boundary condition on the iris can be satisfied by making function J in eq.(1) minimum.

We use different trial functions in the outer region and inner regions. The fields should be continuous across the interface between the two subregions. We can obtain the fields matching conditions by equating the two tangential components of the magnetic fields at the interface. The procedure to make function J in eq.(1) minimum consists of substituting the trial functions into the functional and thereby expressing the functional in terms of coefficients which are the unknowns, such as A, B, C, D in eqs.(2-5). The functional is then differentiated with respect to each coefficient, and the resulting equation is set to zero.

It is true that the periodicity theorem, as embodied in eqs. (2-3), help us to obtain the waves only for the infinitely long structure. If, however, we know the value of β_0 for each wave, and the proper linear combination of the harmonics composing each wave, we can specify completely the excitation in a finite structure. We merely choose the amplitudes of the various waves so as to match boundary conditions at both ends of the structure.

3 NUMERICAL RESULULT

The field series should be truncated when we make J minimum. Figure 2 shows the value of J around the

resonant frequency for the case of disk-loaded structure with round edge beam hole. All examples used in this paragraph have phase 120 degree and structure geometry t=5.0013 mm, D=3.50088 mm, b=40.9 mm, a=10.0025 mm, δ =0.5076 mm, ρ =2.6007 mm. The number of terms used in the outer region is 53 (s from 0 to 52) and in the inner region 53 (n from –26 to 26). J is normalized by $\int (Z_0\mathbf{H})(Z_0\mathbf{H})^*dV$. It is shown that J takes minimum value at the resonance frequency 2855.9777MHz, which means a true field. The truncation errors decrease when the number of the field terms increases. The convergence of the frequency and field can be estimated from the calculated accuracy of the resonance frequency $\Delta f / f$ and the error of the eigenvector ε :

$$\frac{\Delta f}{f} = \frac{f_n - f_{\infty}}{f_{\infty}} \times 100\% \tag{10}$$

$$\varepsilon = \frac{\int (\mathbf{E}_n - \mathbf{E}_{n-1})(\mathbf{E}_n - \mathbf{E}_{n-1})^* dV}{\int \mathbf{E}_n \mathbf{E}_n^* dV}$$
(11)

where f_n is the calculated frequency using n terms, f_{∞} is the convergence frequency when the number of terms is infinite, and E_n is the eigenfunction obtained from nterms. Figure 3 shows the resonance frequency, accuracy of frequency and estimated error ε of the first longitudinal mode for different numbers of terms in the inner region, where $f_{\infty} \approx f_{55}$ is used for the calculation of $\Delta f / f$. Note that the accuracy in figure 3 is not the exact accuracy. Nevertheless it shows that the convergence rate of the eigenvalue is very rapid. Usually, we choose the series term s ranging from 0 to 52 in the outer region and n from -26 to 26 in the inner region, which correspond to the total number of terms 53 in Figure 3, to reach an eigenvalue convergence $\delta k / k$ better than 1×10^{-6} . The estimated error of the eigenvector also approximately decays exponentially with the number of terms. The error can reach to 10^{-6} when *n* is from -26 to 26. The CPU time is mainly decided by the number of terms in the inner region. The number of terms in the outer region has little effect. We should choose a suitable number of terms in order to get accurate results within a short calculation time. For the above case, The CPU time needed for LONGTRANSVRS to reach within 0.07MHz of its convergence is about 96 sec while the CPU time for SUPERFISH to reach the same precision is about 4196 sec using a 500MHz Pentium-III computer. Figure 4 shows calculated frequency and CPU time used for the SUPERFISH code with different mesh sizes and LONGTRANSVRS code with different numbers of terms number as in Figure 3.

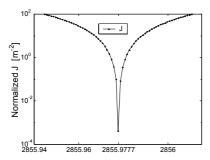


Figure 2 Normalized J vs. frequency

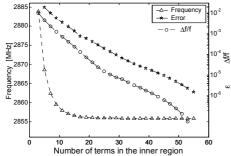


Figure 3 Convergence of frequency and field

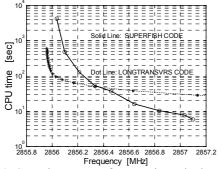


Figure 4 CPU time versus frequencies calculated by the SUPERFISH code and LONGTRANSVRS code.

LONGTRANSVRS code gives practically the same frequency as TRANSVRS code for a structure with right-angled beam hole edges. For a structure with t=2mm, D=8.7474mm, b=10.779mm, a=4.5mm, ϕ_s =0, the first dipole mode frequency is 15.9402820364166 GHz by LONGTRANSVRS, which is the same as that by TRANSVRS within 10^{-3} Hz. Figure 5 shows the convergence of frequency and field for the first dipole mode with phase 120° .

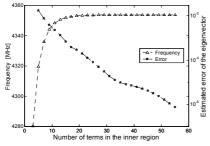


Figure 5 Convergence of frequency and field (m=1)

4 SUMMARY AND CONCLUSIONS

Based on the variational method, a code, LONGTRANSVRS was developed which could calculate a disk-loaded structure with rounded edge beam hole shape. It can calculate all modes in disk-loaded structures with high accuracy, which depends on the number of terms used. Our method has good convergence and it's much faster than SUPERFISH code when the accuracy is high. This code provides a powerful tool to design the disc-loaded structure, such as searching for structure geometry (a or b) with fixed field modes.

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