# DESIGN STUDY OF 325MHz RF POWER COUPLER FOR SUPERCONDUCTING CAVITY

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#### Abstract

We present the design study of the RF input power coupler for 325 MHz superconducting cavities. The power coupler, based on a conventional coaxial transmission line, provides RF powers to the cavity up to 12 kW in CW mode. The thermal interceptors are considered as 4.5 K and 40 K or 4.5 K and 77 K corresponding to the usage of liquid Helium only or both liquid Helium and Nitrogen for cryogenic temperature to reduce the thermal load. The transition box (T-box), which is assembled with power coupler, is designed and applied for impedance matching and inner conductor cooling.



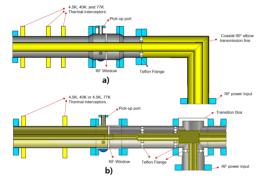


Figure 1: General view of the prototype power coupler

a) 1<sup>st</sup> Version b) 2<sup>nd</sup> Version.

The power coupler is based on conventional coaxial transmission line. Both couplers have three diagnostic ports for vacuum, arc, and electron pick-ups to monitor RF breakdowns. The number of thermal interceptor-'s has been changed for the diversity of cooling gas (liquid Helium only or both liquid Helium and Nitrogen) used for thermal shield in cryo-module of various institutions. Figure 1 shows the prototype design of power couplers and the design requirements are listed in Table 1 [1].

Table 1: Design Requirements of the Input Coupler

Parameters	Values	Unit
Operating frequency	325	MHz
Pass band ( $S_{11} < 0.1$ )	3	MHz
$S_{11}$ at 325MHz	≤-30	dB
Operating Power	12	kW
Q <sub>ext</sub>	$4 \times 10^6$	-

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The main feature of changes from the  $1^{st}$  version to the  $2^{nd}$  one is the reduced diameter of the outer conductor at the RF window section, the presence of the transition box and copper plating on the inside of the outer conductor. The reduced outer diameter decreases the size of coupler port in cryo-module and in accordance with assembling procedure, the RF window must be inside the outer wall of cryo-module, which requires a hole that is larger than the height of the pick-up port. The main purpose of transition box is impedance matching for difference between RF window section and conventional  $50~\Omega$  coaxial transmission line. The cooling gas can be inserted at the end of the T-box through the inner conductor to reduce the temperature increase of the antenna tip and RF window due to RF power. Engineering design for production is in progress.

# **SIMULATION**

Electromagnetic Simulation

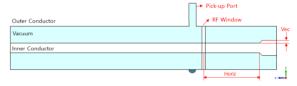


Figure 2: Geometric change of ceramic part (Vec and Horiz).

Figure 2 shows variations of both diameter (Vec) and length (Horiz) of the outer conductor at the RF window to minimize reflected powers at 325 MHz. The antenna diameter is 17.1 mm and was determined by multi-pacting analyses. We performed simulations to check effects of the inner conductor diameter changes from 8 mm to 15 mm. The horizontal length change affects the intensity variance of reflected powers ( $S_{11}$ ) at 325 MHz. We performed the simulation to minimize reflected powers at operating frequency. The optimal value of vertical/horizontal length are 10 mm and 101 mm, respectively, as shown in Figure 3. As a result, the impedance of the antenna section is 90  $\Omega$  and the that of other section is 62.8  $\Omega$  [2].

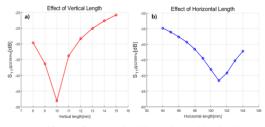
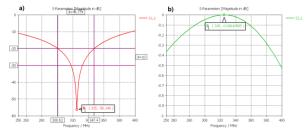
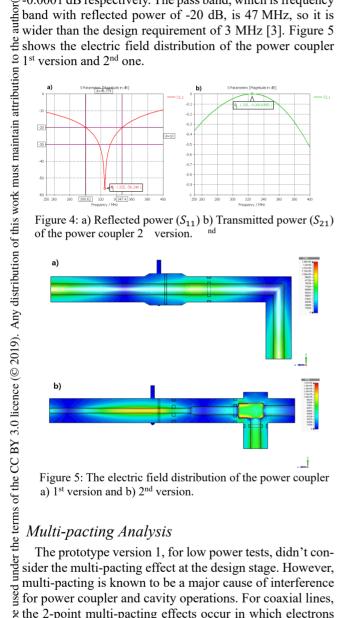


Figure 3: Effect of a) vertical length and b) horizontal length change.

The outer conductor is composed of 0.8-mm-thick stainless steel with 15 µm copper plating at the vacuum section and the inner conductor is made of 1.5-mm-thick OFHC. section have 45 degree angles to avoid wake fields and multi-pacting effects in the coupler. The DE g of the coupler prototype are shown in Figure 4. The geometry of the coupler was tuned to have the minimum reflection and maximum transmission occurred at operating frequency of 225 NAT quency of 325 MHz, and their values are -56.25 dB and quency of 325 MHz, and their values are -56.25 dB and -0.0001 dB respectively. The pass band, which is frequency band with reflected power of -20 dB, is 47 MHz, so it is wider than the design requirement of 3 MHz [3]. Figure 5





for power coupler and cavity operations. For coaxial lines, 8 the 2-point multi-pacting effects occur in which electrons ≈reciprocate between the outer and inner conductors. The change in the diameter of inner conductor of coaxial line makes the impedance difference that can reduce the intensity of multi-pacting [4]. Figure 6 shows the growth rate variation due to the impedance change in the power coupler.

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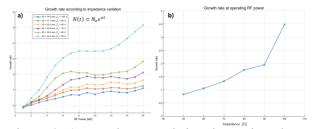


Figure 6: a) Growth rate variation due to impedance change, b) Growth rate at operating RF power.

The growth rate was the largest when the impedance was 50  $\Omega$ . When the impedance is increase, the rate is significantly decreased and the smallest can be seen at  $100 \Omega$  in Figure 6 a). However, the smaller the inner conductor, the harder it becomes to design and circulate the refrigerant for cooling the antenna tip and RF window. Thus, we chose the 17.1 mm for inner conductor diameter to minimize the multi-pacing and difficulty in engineering design.

# Thermal Analysis

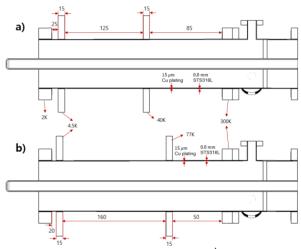


Figure 7: Schematic drawing of the 2<sup>nd</sup> version power couplers with a) 4.5K and 40K thermal interceptors. b) 4.5K and 77K thermal interceptors.

Figure 7 shows the optimized locations of each interceptors. The power coupler bridge between room temperature (300 K) and cryogenic temperature (2 K) and acts as a continuous heat source to the superconducting cavities. The optimal locations of the thermal interceptors connected to the outer conductor were chosen to minimize the static and dynamic heat loads. The thermal interceptors of 4.5 K and 40K are available for cavities use only liquid helium as cooling gas in cryo-module, and those of 4.5 K and 77 K can be used for cavities with liquid helium and nitrogen. Table 2 presents analytical thermal loads on each interceptor location.

Table 2: Static Heat Load to Thermal Interceptors

Interceptor Temp. (K)	4.5K, 40K	4.5K, 77K
	Values (W)	Values (W)
2K	0.3070	0.3838
4.5K	1.5459	1.3602
40K	0.5189	-
77K	=	0.3445

# Mechanical Analysis

Figure 8 shows the stresses, strains and deformations at the ceramic window when three times of the gravity force are applied to the inner conductor and ceramic window [5].

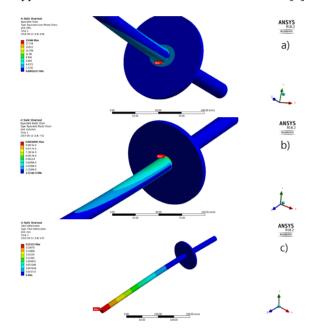


Figure 8: a) Stress, b) Strain, and c) Deformation of inner conductor and ceramic window.

The estimated stresses on the antenna (OFHC,  $\sigma_{vield} = 33.3 MPa$ ) and the window (99.7%-alumina, t = 6 mm,  $\sigma_{vield} = 2350 MPa$ ) are below the yield strength of the materials [6].

The modal analysis was also performed to avoid vibration modes due to ground vibrations and harmonics of the AC electric power frequency of 60 Hz. The fixed point is same as previous analysis. Figure 9 shows the four vibrational modes of the inner conductor connected with the ceramic window. The resonance frequency modes of the power coupler antenna are listed in Table 3 [7].

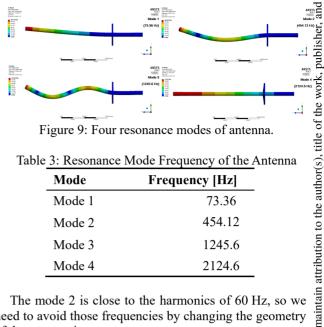


Figure 9: Four resonance modes of antenna.

Table 3: Resonance Mode Frequency of the Antenna

Mode	Frequency [Hz]
Mode 1	73.36
Mode 2	454.12
Mode 3	1245.6
Mode 4	2124.6

The mode 2 is close to the harmonics of 60 Hz, so we need to avoid those frequencies by changing the geometry of the antenna inner part.

# **CONCLUSION**

We studied prototype designs of the 325-MHz input coupler for proton/heavy-ion superconducting cavities. The 1st version has already been fabricated and is in the test preparation stage. It has the advantages of minimize the multipacting effects, cooling system for antenna tip and RF window.

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