



BPM electrode and high power feedthrough –Special topics in wideband feedthrough–

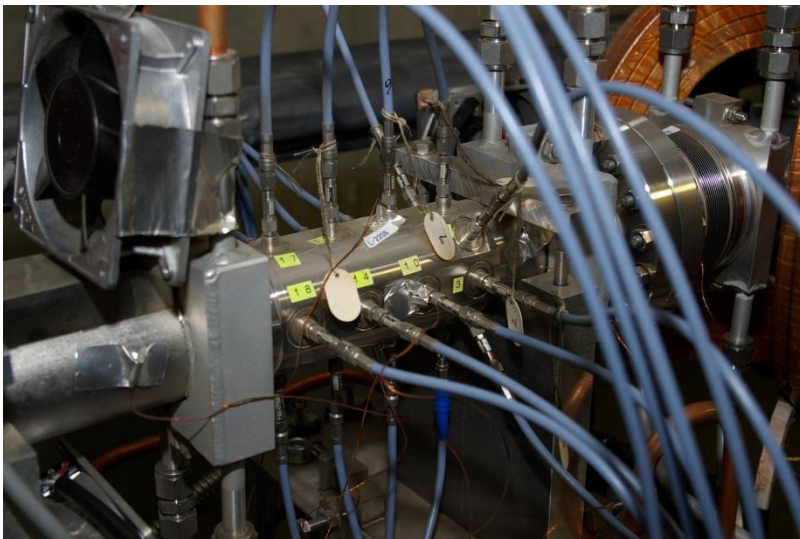
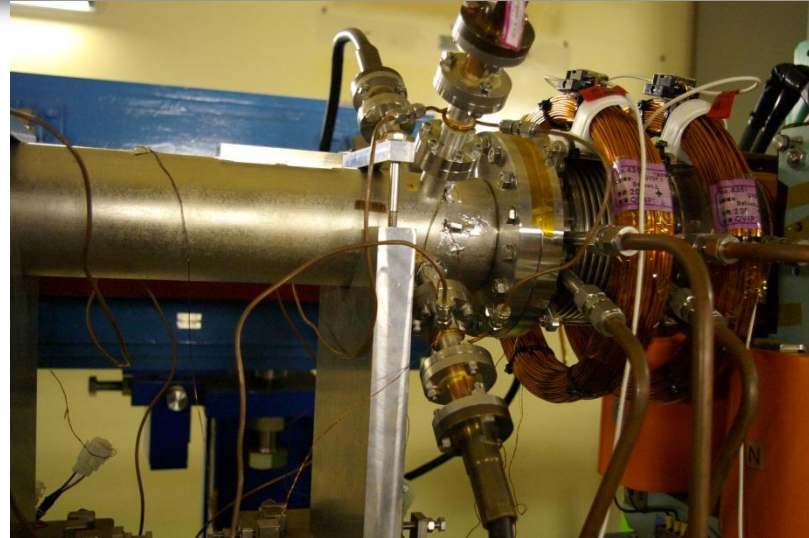
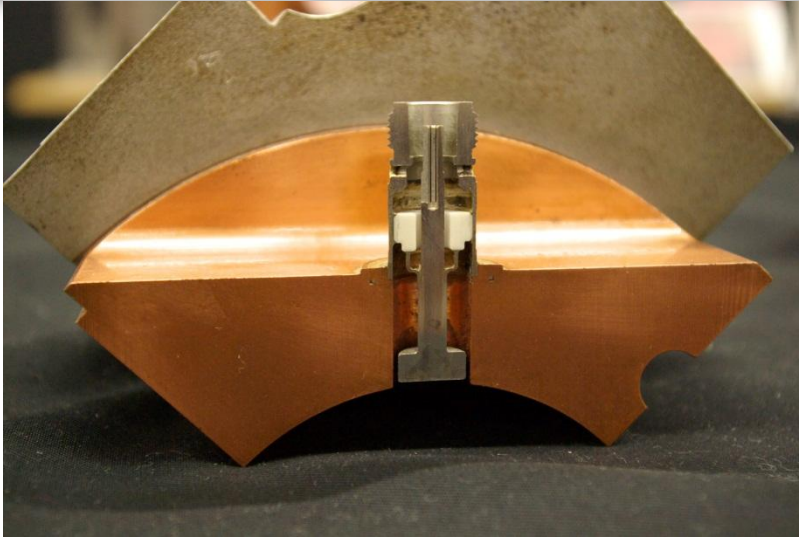
Makoto Tobiyama

KEK Accelerator Laboratory

Introduction

- **Beam runs in vacuum chambers (mostly) made of good-conducting metal.**
 - Vacuum chambers shield (fast) electro-magnetic wave from / to the beam.
 - Need some kind of “Feedthrough” to get / feed the electro-magnetic signal from / to beam.
- **Wideband coaxial feedthrough**
 - Button-type electrode
 - Stripline kicker / monitor
 - Plate-type position monitor
 - Ion (or electron cloud) clearing electrode

Feedthrough



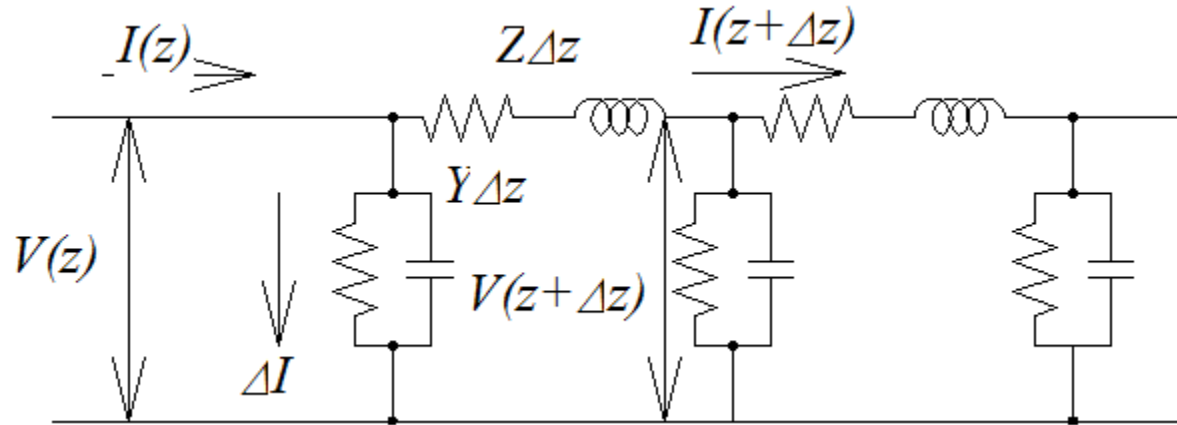
Need to consider..

- **Specification of the feedthrough**
 - Structure
 - Total size
 - Kind of RF connector, vacuum structure...
 - Mechanical (and heating) toughness
 - Frequency range, allowed SWR, reflection..
 - Power range
 - Beam induced power
 - Supplied power from outside
 - allowed (V)SWR

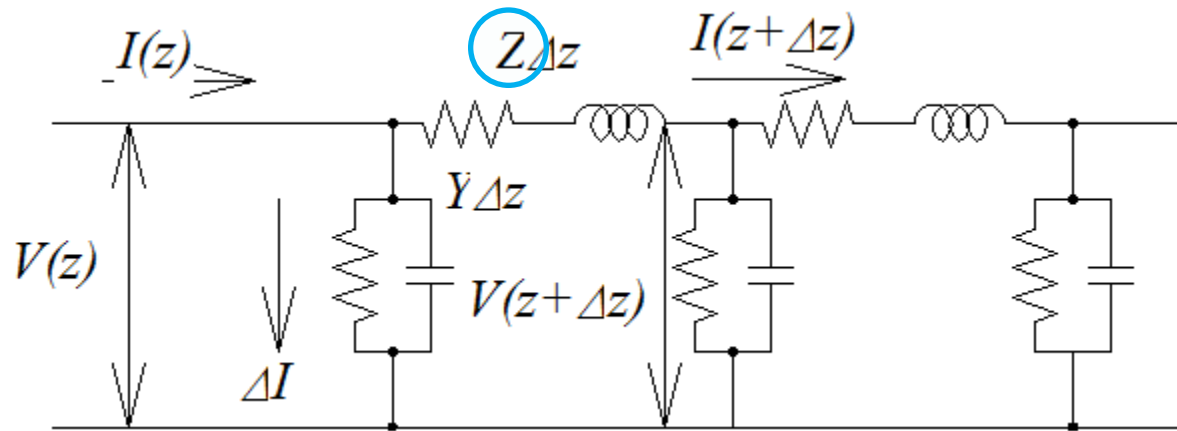
In this tutorial

- **Review of transmission line theory**
 - S-Parameter, (V)SWR
 - Time domain behavior (TDR)
- **Frequency-domain simulation**
 - Ansys HFSS
- **Time-domain simulation**
 - GdfidL
- **Design, evaluation of the button electrodes**
- **Design, evaluation of the high-power feedthroughs**
- **Summary**

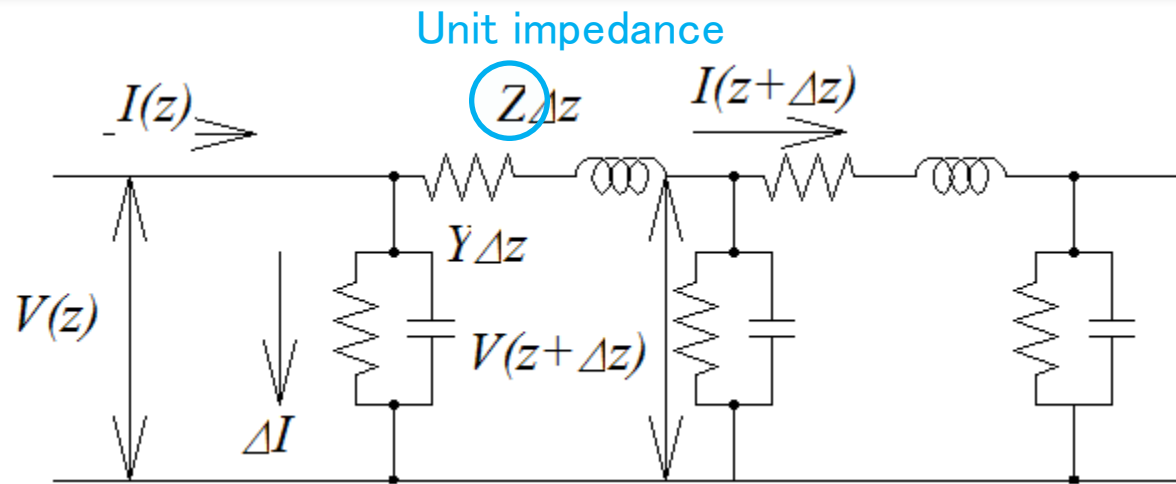
Circuit representation of a uniform transmission line



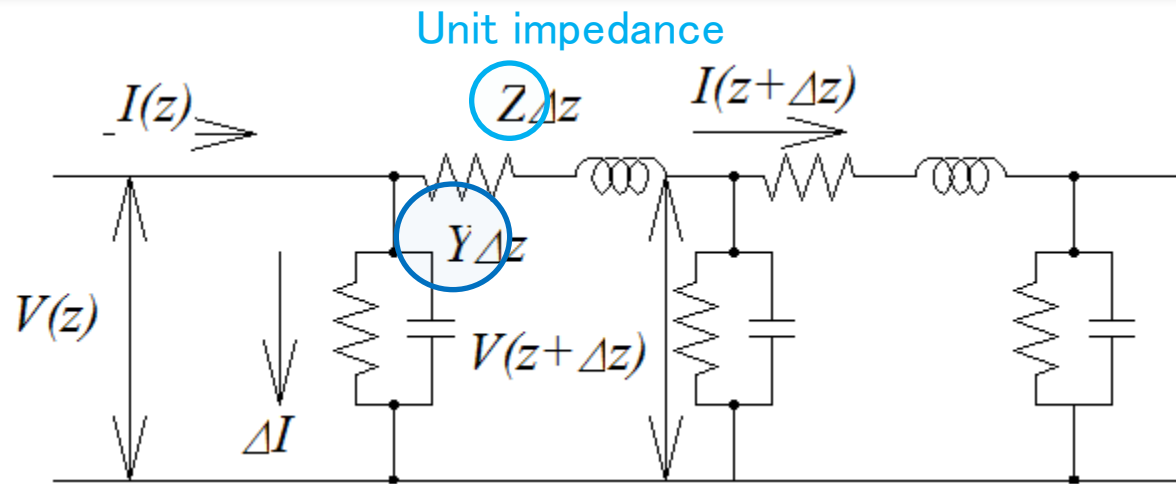
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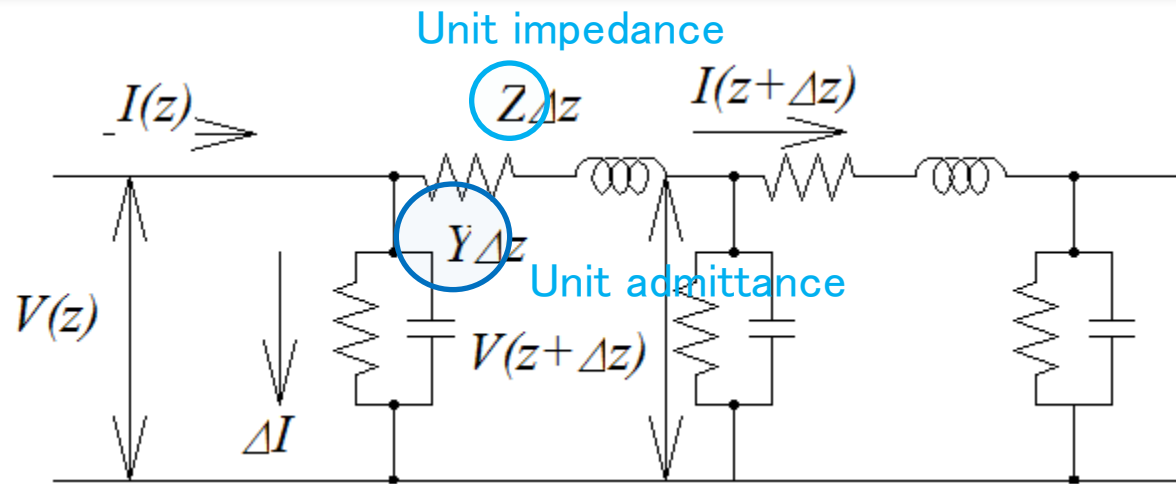
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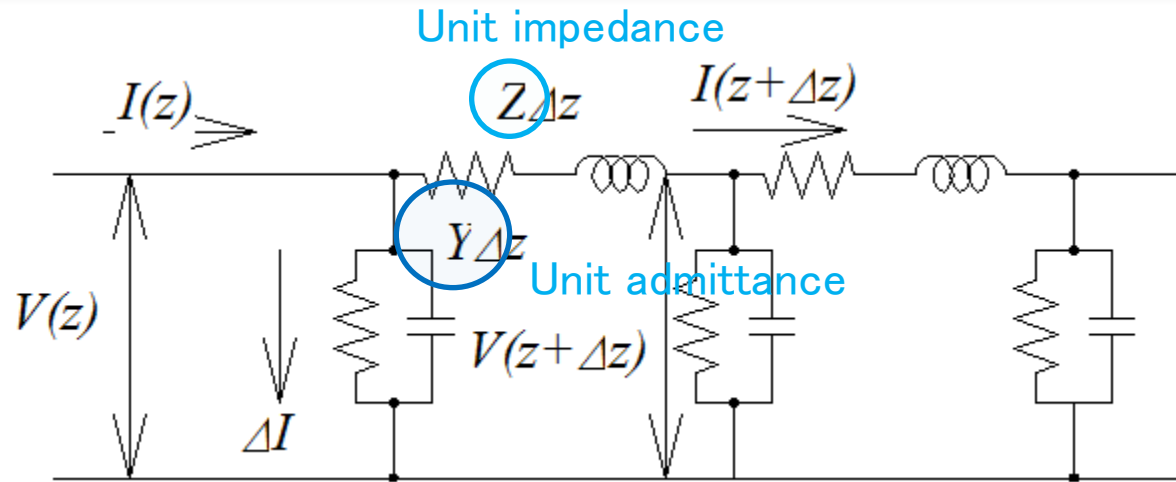
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Circuit representation of a uniform transmission line



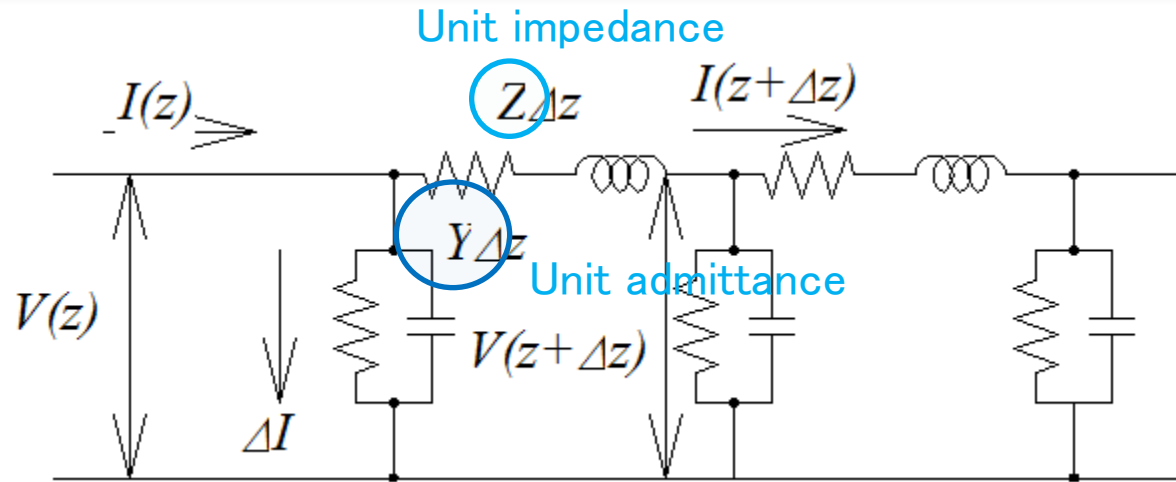
Circuit representation of a uniform transmission line



$$\Delta V = -IZ\Delta z$$

$$\Delta I = -VY\Delta z$$

Circuit representation of a uniform transmission line



$$\Delta V = -IZ\Delta z$$

$$\Delta I = -VY\Delta z$$

$$\frac{dV}{dz} = -IZ$$

$$\frac{dI}{dz} = -VY$$

$$\frac{d^2V}{dz^2} = \gamma^2 V, \frac{d^2I}{dz^2} = \gamma^2 I$$

$$\gamma^2 = ZY$$

Transmission line (cont.)

■ Solution

$$\begin{aligned} V &= V_1 e^{-\gamma z} + V_2 e^{\gamma z} \\ &= V_1 e^{(j\omega t - \gamma z)} + V_2 e^{(j\omega t + \gamma z)} \end{aligned}$$

■ Normal transmission line

$$Z = R + jL\omega$$

$$Y = G + jC\omega$$

$$\gamma = (LC\omega^2)^{\frac{1}{2}} \left[1 - \frac{RG}{LC\omega^2} - j \left(\frac{G}{C\omega} + \frac{R}{L\omega} \right) \right]^{\frac{1}{2}}$$

$$\approx j\omega\sqrt{LC} \left[1 - j \left(\frac{G}{C\omega} + \frac{R}{L\omega} \right) \right]^{\frac{1}{2}}$$

$$\approx j\omega\sqrt{LC} \left(1 - j \frac{R}{2L\omega} \right)$$

Transmission line (cont.)

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■ Normal transmission line

Resistance

$$Z = \textcircled{R} + jL\omega$$

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■ Normal transmission line

Resistance Inductance

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Parallel
conductance

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Parallel

conductance

Capacitance

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Parallel conductance Capacitance

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$$R / L\omega \ll 1$$

$$G / C\omega \ll 1$$

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Propagation constant

$$\gamma = \alpha + j\beta$$

$$\alpha = \frac{R}{2} \sqrt{\frac{C}{L}}, \beta = \omega \sqrt{LC}$$

$$V = V_1 e^{j(\omega t - \beta z)} e^{-\alpha z} + V_2 e^{j(\omega t + \beta z)} e^{\alpha z}$$

■ Phase velocity

$$v_p = \frac{dz}{dt} = \frac{\omega}{\beta} = \frac{1}{\sqrt{LC}}$$

$$\beta = \frac{2\pi f}{f\lambda} = \frac{2\pi}{\lambda}$$

Propagation constant

Attenuation constant

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Attenuation constant

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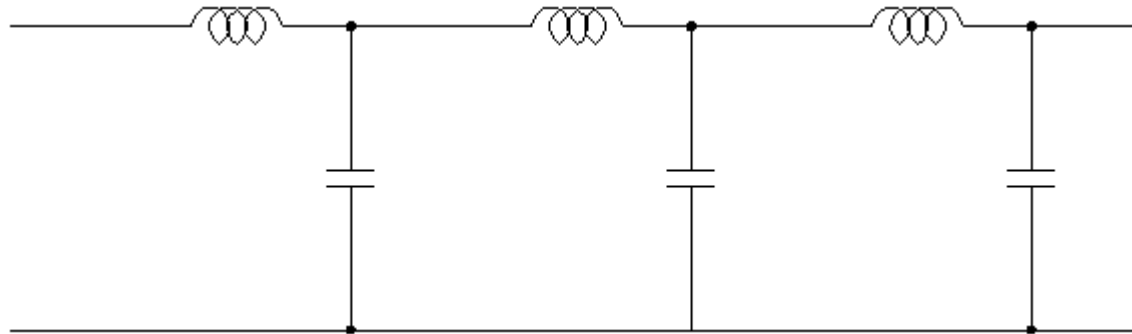
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Loss less line



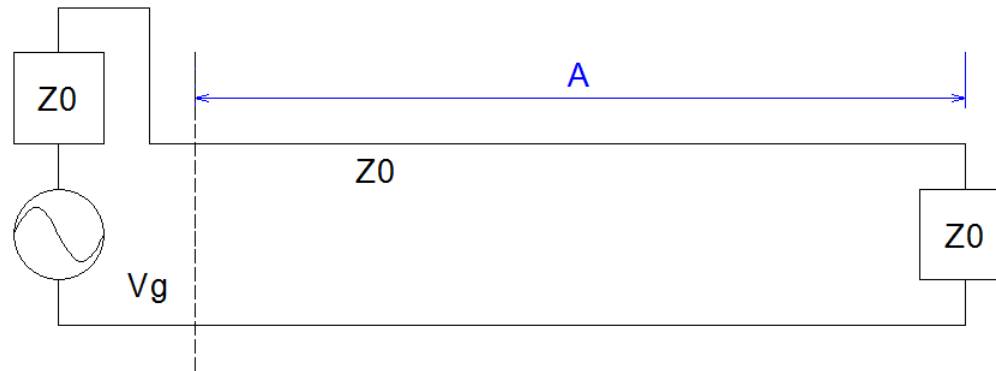
$$V = V_1 e^{-j\beta z} + V_2 e^{j\beta z}$$

$$I = \frac{1}{Z_0} (V_1 e^{-j\beta z} - V_2 e^{j\beta z})$$

$$Z_0 = \frac{Z}{\gamma} = \sqrt{\frac{L}{C}}$$

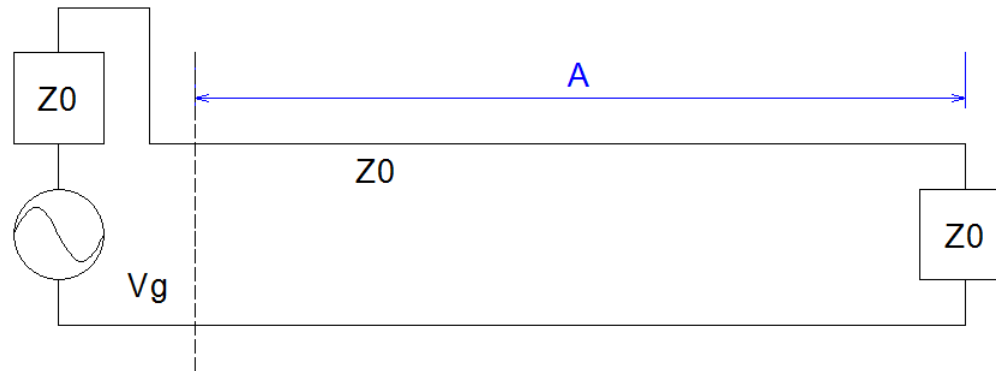
$$\beta = \omega \sqrt{LC}$$

Impedance matched line



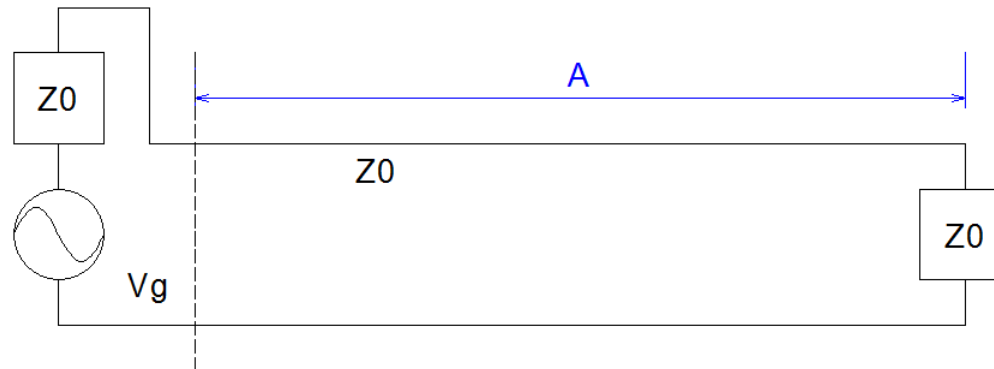
$$\frac{V(A)}{I(A)} = Z_0 = Z_0 \frac{V_1 e^{-j\beta A} + V_2 e^{j\beta A}}{V_1 e^{-j\beta A} - V_2 e^{j\beta A}}$$

Impedance matched line



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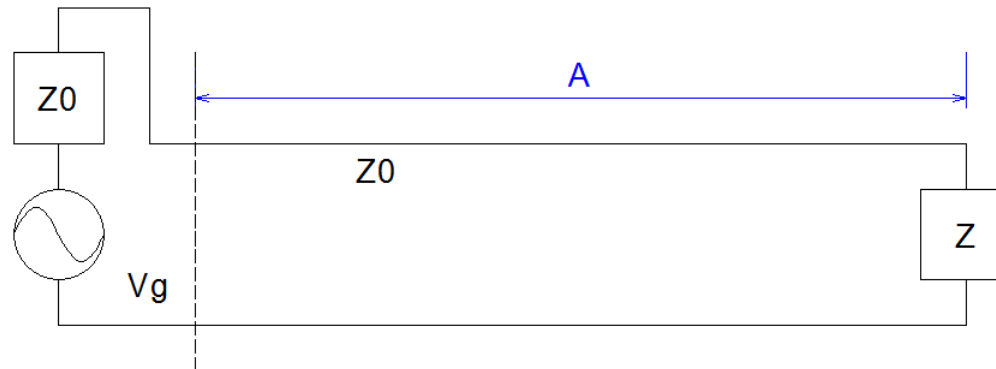
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- No reflection! (Only forwarding wave exist)

Arbitrary termination

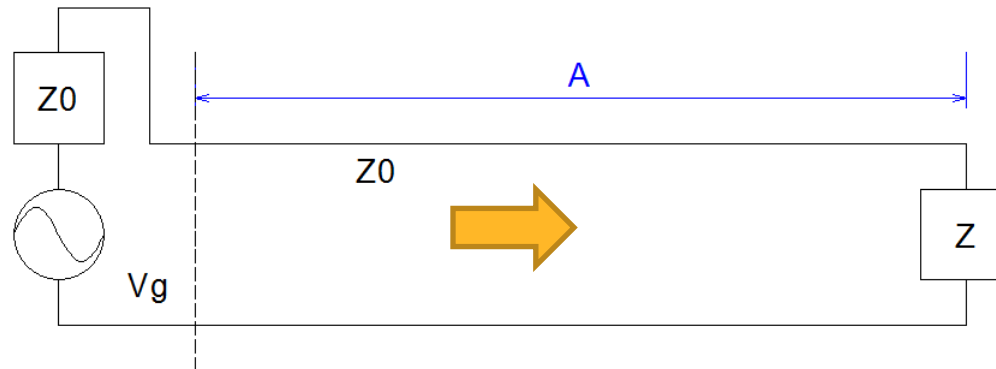


- To satisfy $V(A)/I(A)=Z$, we need both forwarding and reflecting wave.

$$V = V_1 e^{-j\beta z} \left(1 + 2 \frac{|V_2|}{|V_1|} \cos \chi + \frac{|V_2|^2}{|V_1|^2} \right)^{\frac{1}{2}} \exp \left(j \tan^{-1} \frac{\frac{|V_2|}{|V_1|} \sin \chi}{1 + \frac{|V_2|}{|V_1|} \cos \chi} \right)$$

$$\chi = 2\beta z + \theta_2 - \theta_1$$

Arbitrary termination

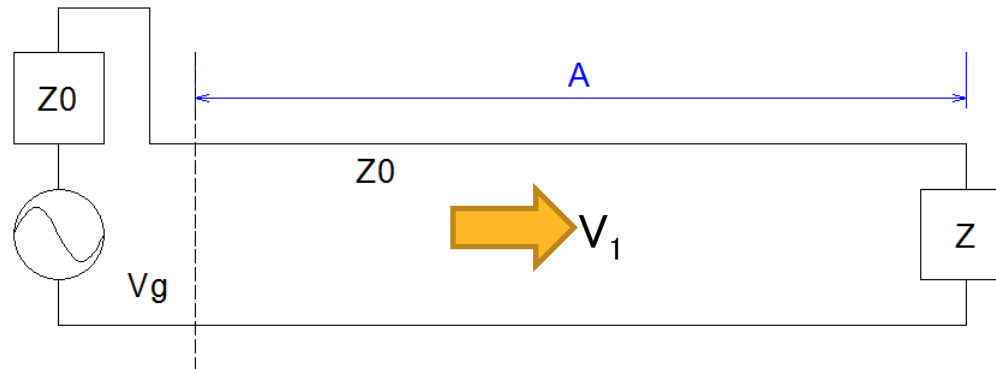


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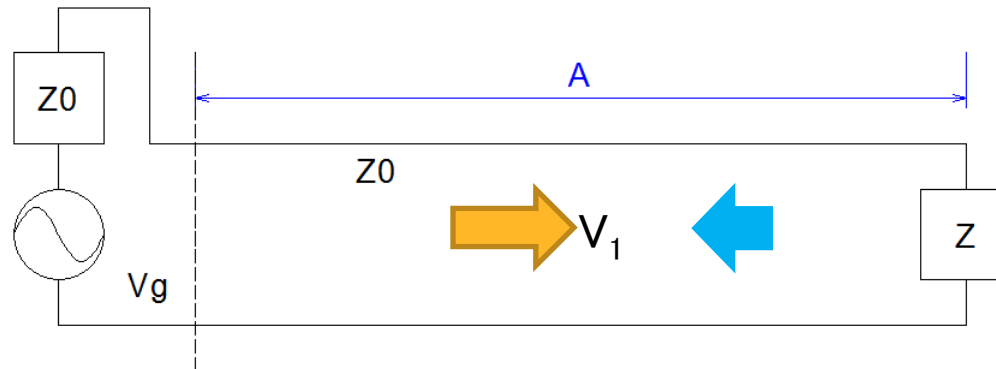


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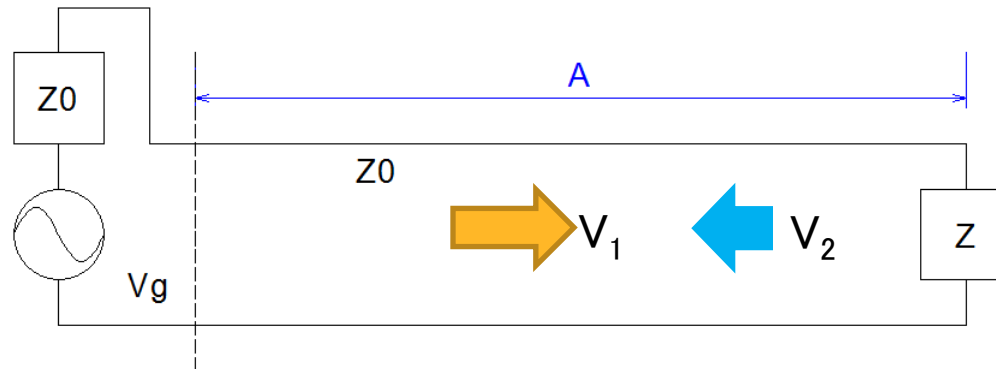


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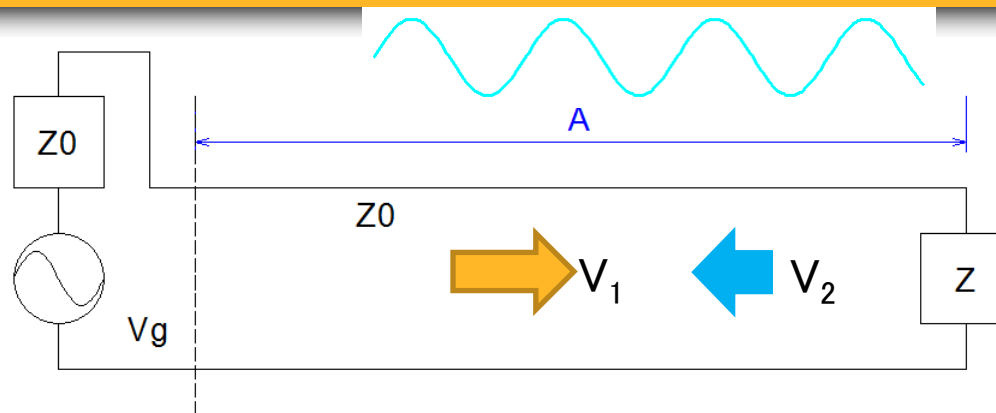


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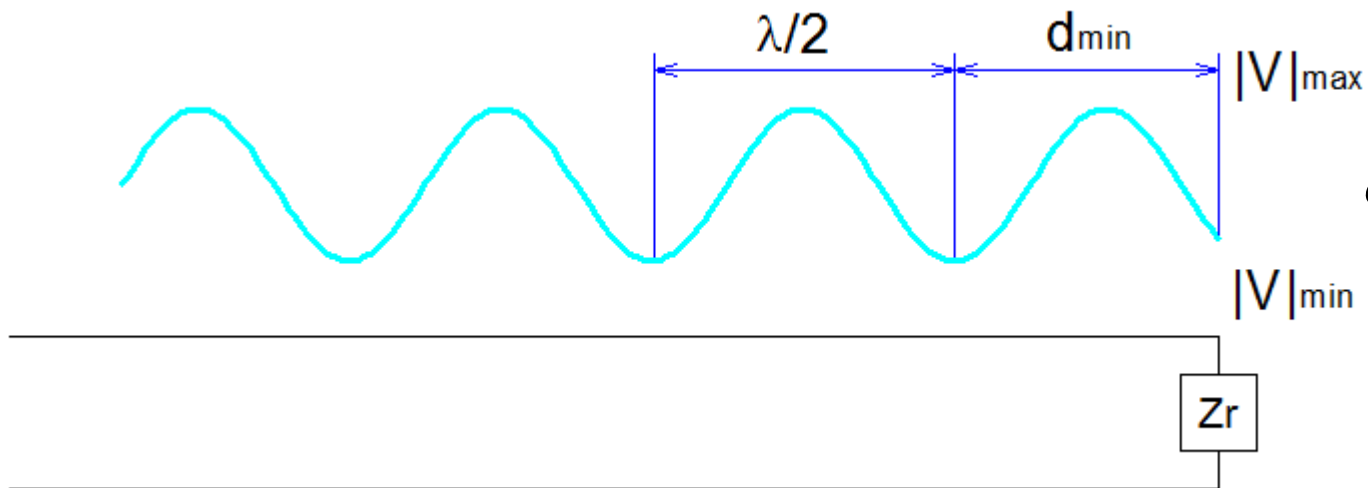


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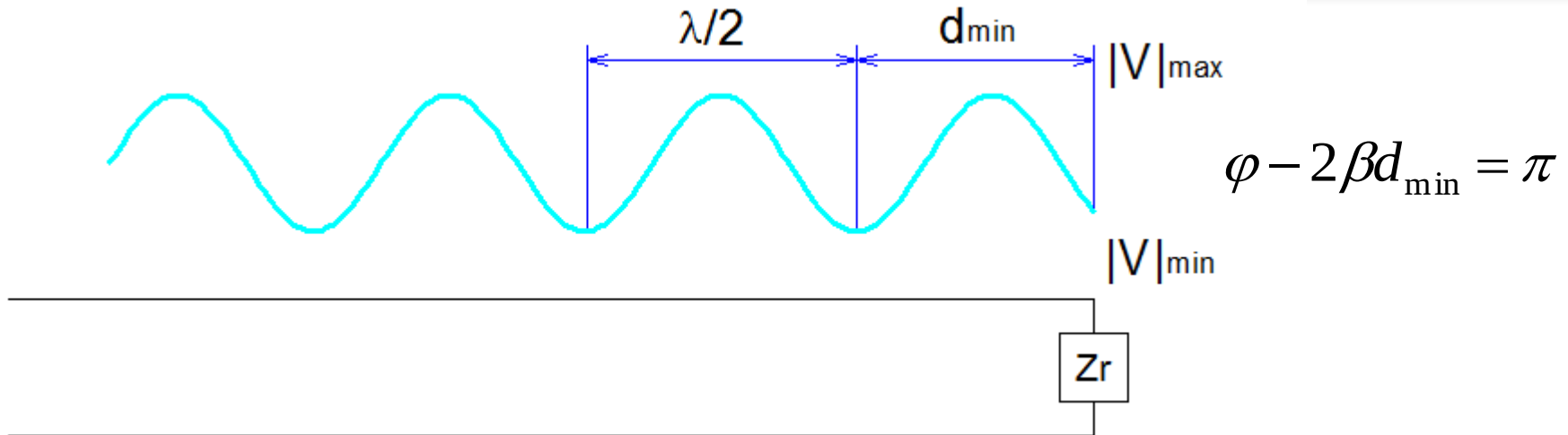
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Standing wave ratio



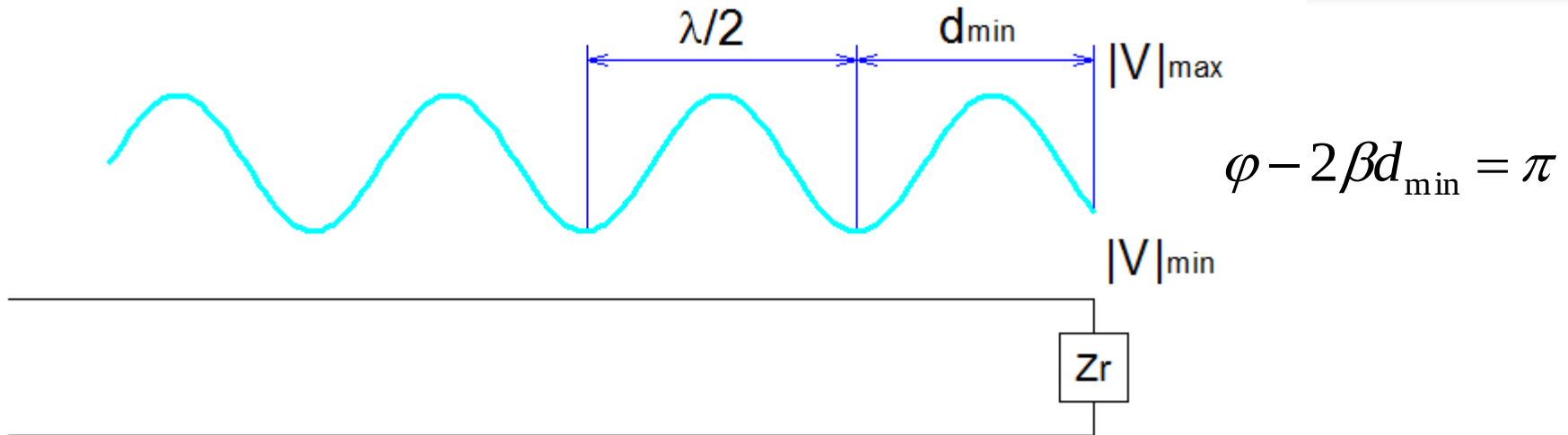
$$\varphi - 2\beta d_{\min} = \pi$$

Standing wave ratio



- Standing wave ratio (SWR)

Standing wave ratio

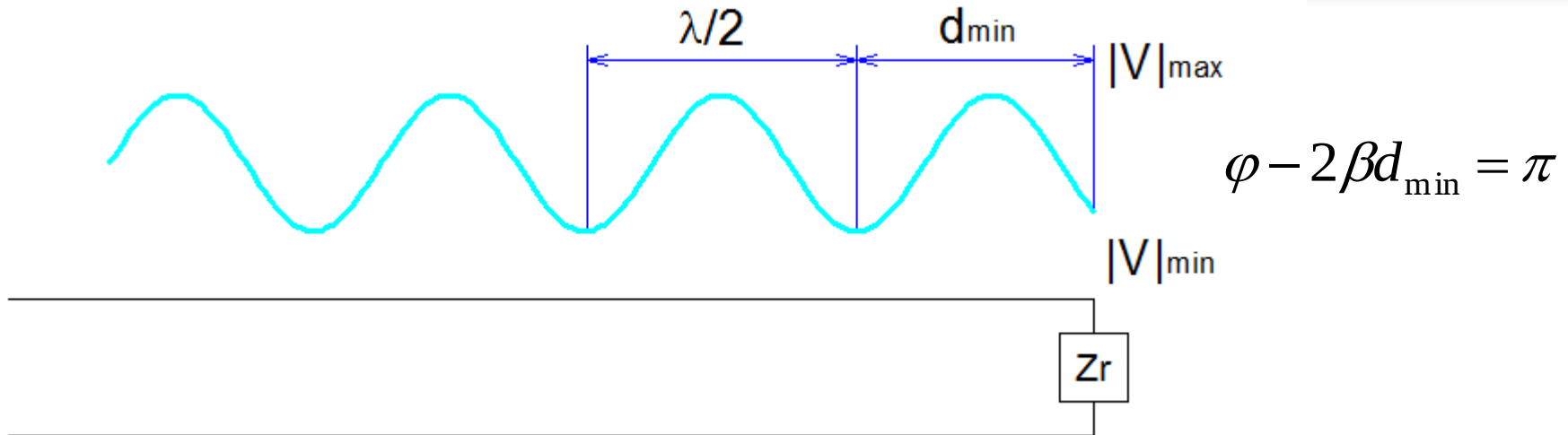


■ Standing wave ratio (SWR)

$$S = \frac{|V|_{\max}}{|V|_{\min}} = \frac{1 + |\rho|}{1 - |\rho|}$$

$$|\rho| = \frac{S - 1}{S + 1}$$

Standing wave ratio



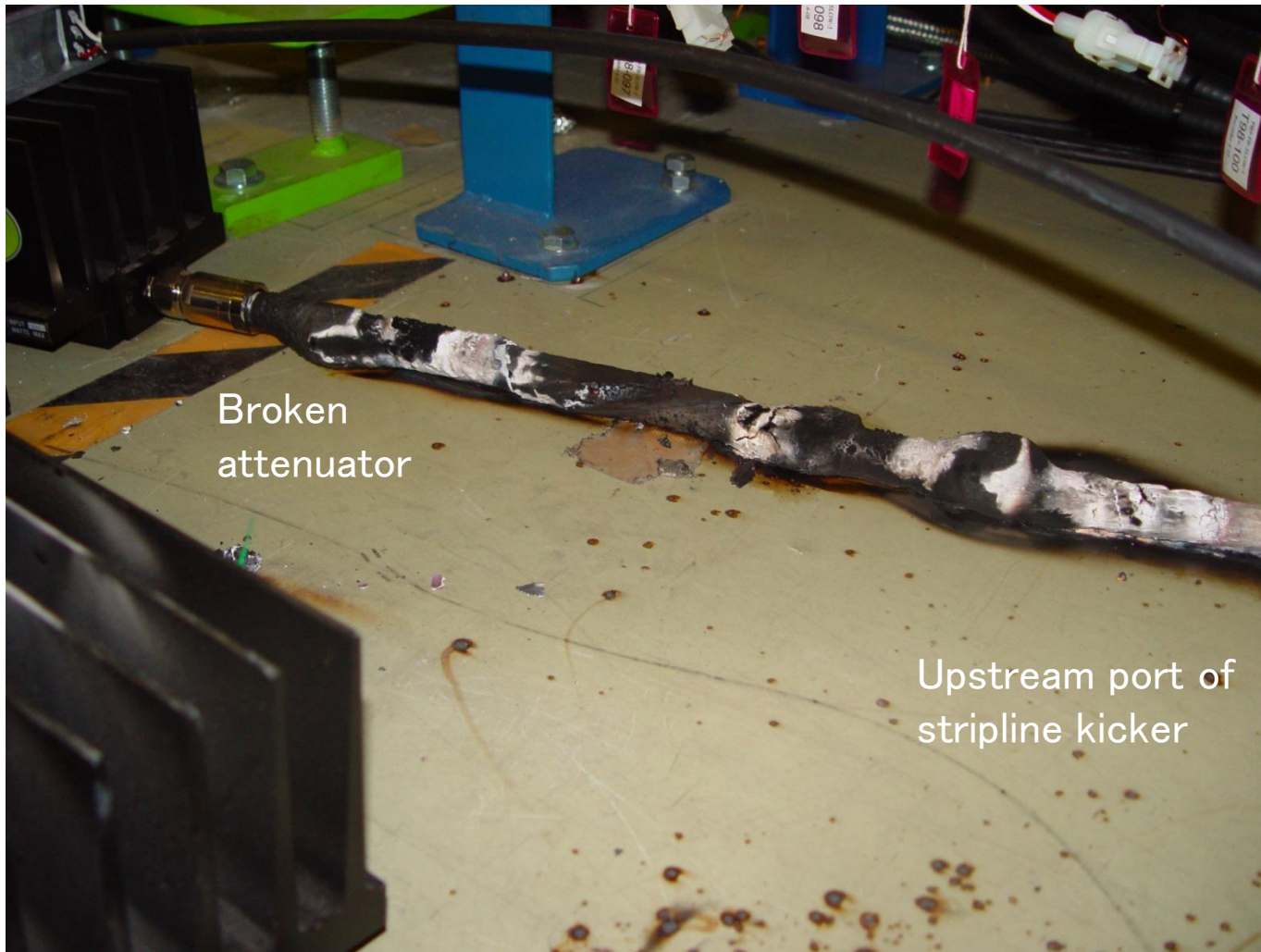
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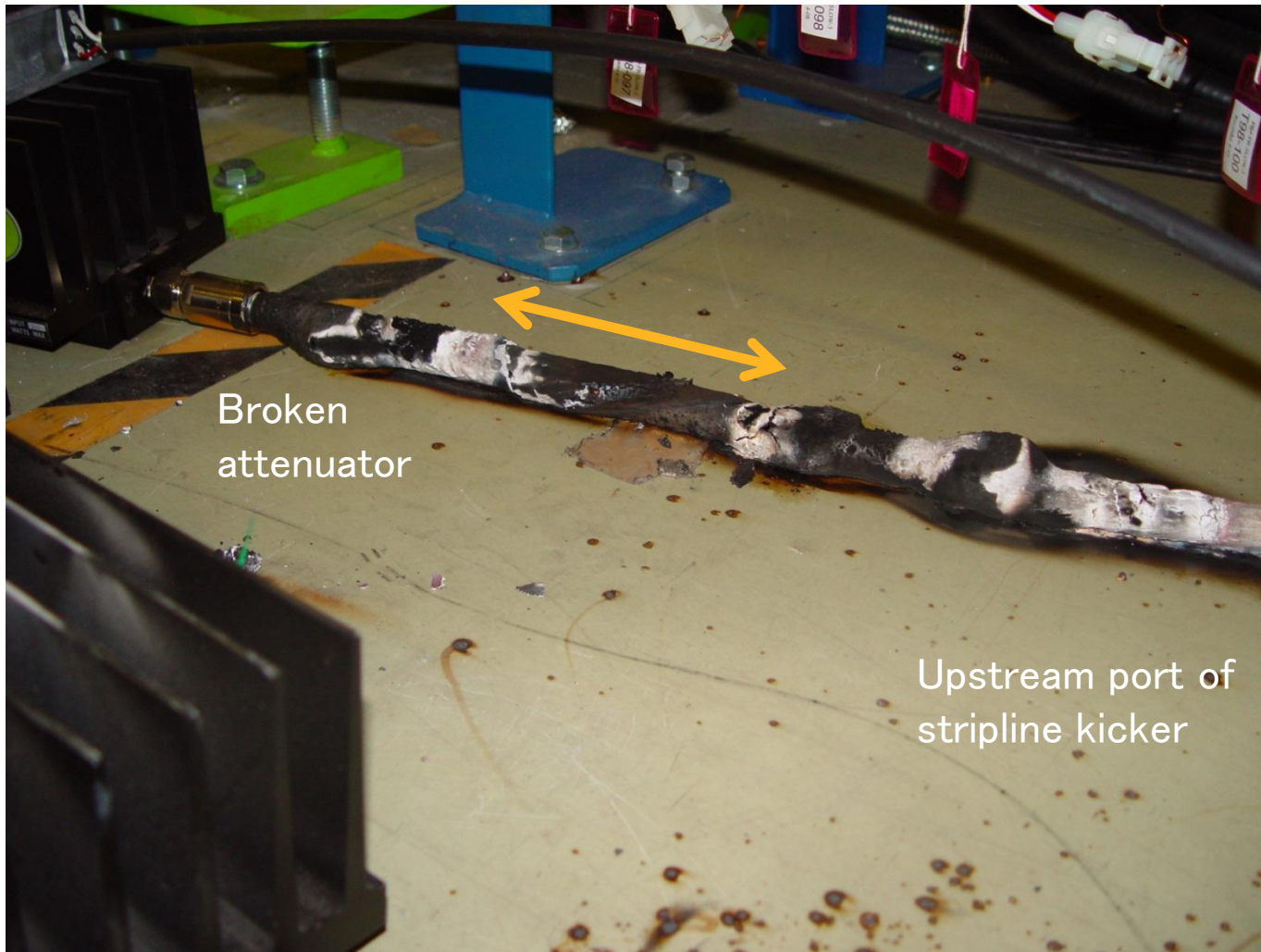
ρ : reflection constant

$$|\rho| = \frac{S - 1}{S + 1}$$

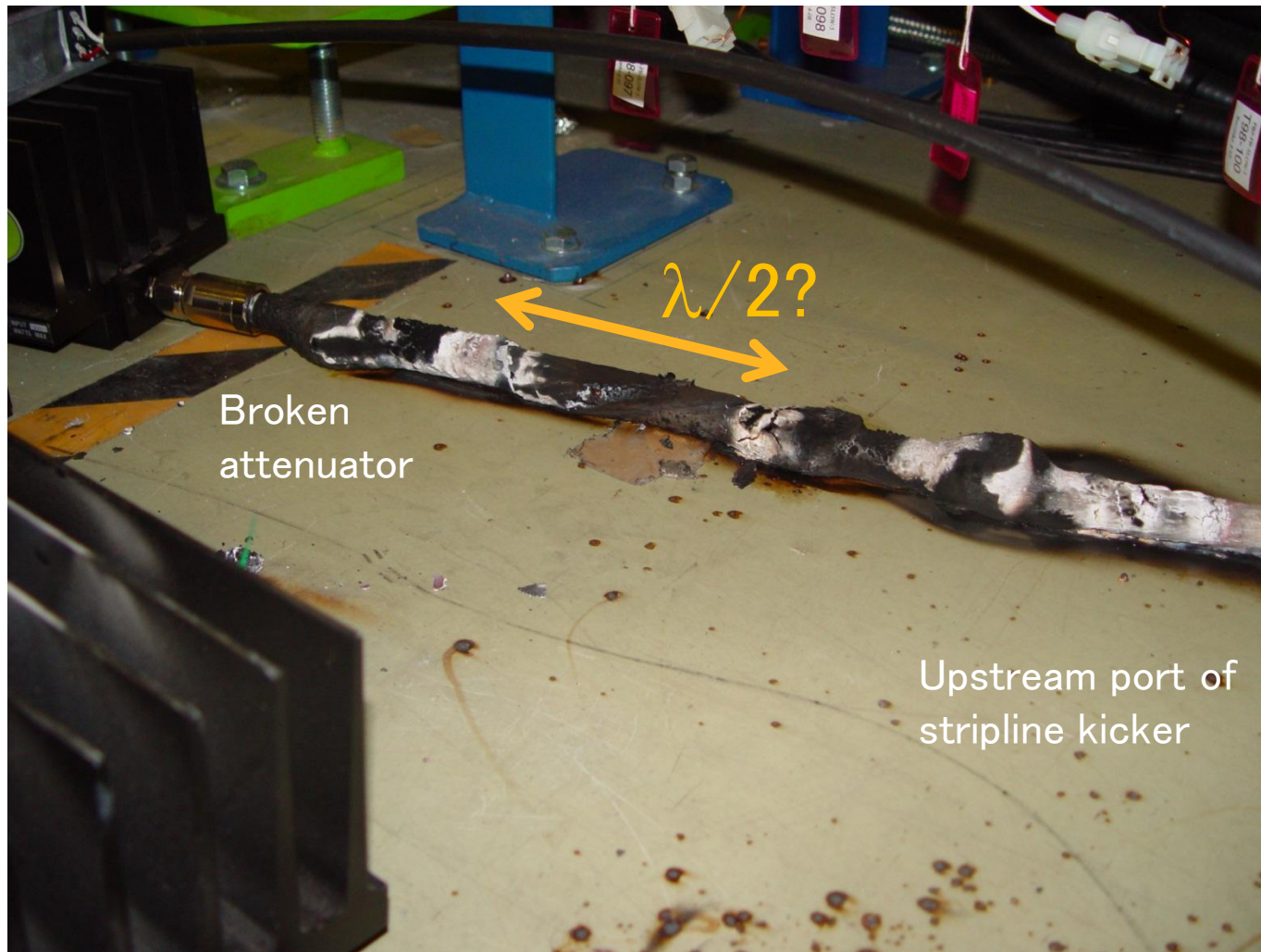
Example of standing wave!



Example of standing wave!



Example of standing wave!



Impedance of the load

- We can estimate the load impedance using

$$Z_r = Z_0 \frac{1 + \rho}{1 - \rho}$$

$$\begin{aligned} \frac{Z_r}{Z_0} &= \frac{1 + |\rho|e^{j\varphi}}{1 - |\rho|e^{j\varphi}} = \frac{(S + 1) + (S - 1)e^{j(2\beta d_{\min} + \pi)}}{(S + 1) - (S - 1)e^{j(2\beta d_{\min} + \pi)}} \\ &= \frac{1 - jS \tan \beta d_{\min}}{S - j \tan \beta d_{\min}} \end{aligned}$$

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$$Z_r = Z_0 \frac{1 + \rho}{1 - \rho}$$

Reflection constant

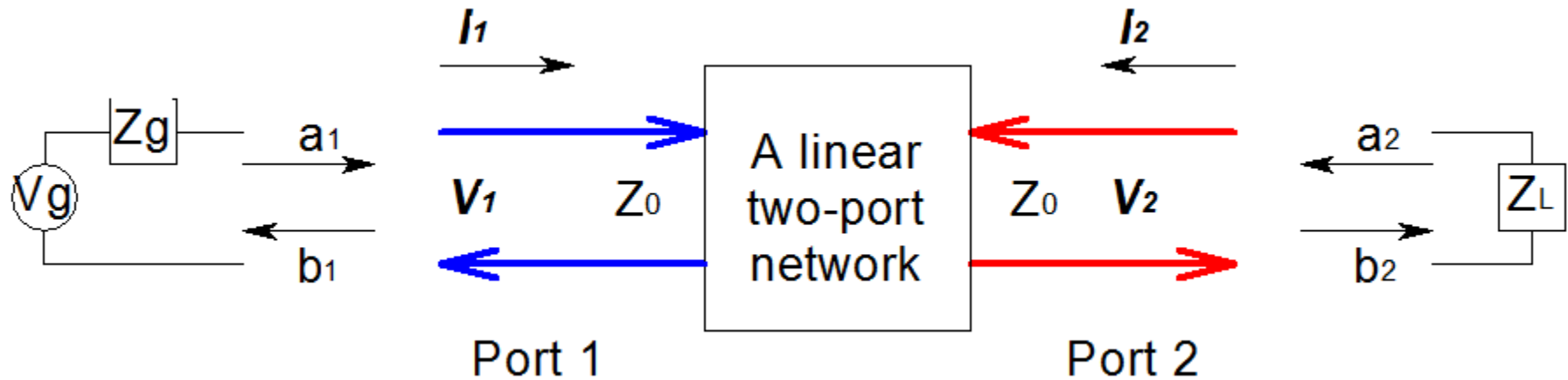
$$\begin{aligned} \frac{Z_r}{Z_0} &= \frac{1 + |\rho|e^{j\varphi}}{1 - |\rho|e^{j\varphi}} = \frac{(S + 1) + (S - 1)e^{j(2\beta d_{\min} + \pi)}}{(S + 1) - (S - 1)e^{j(2\beta d_{\min} + \pi)}} \\ &= \frac{1 - jS \tan \beta d_{\min}}{S - j \tan \beta d_{\min}} \end{aligned}$$

Impedance of the load

- We can estimate the load impedance using

$$\begin{aligned}
 Z_r &= Z_0 \frac{1 + \rho}{1 - \rho} \quad \begin{array}{l} \text{Reflection} \\ \text{constant} \end{array} \\
 \frac{Z_r}{Z_0} &= \frac{1 + |\rho| e^{j\varphi}}{1 - |\rho| e^{j\varphi}} = \frac{(S + 1) + (S - 1)e^{j(2\beta d_{\min} + \pi)}}{(S + 1) - (S - 1)e^{j(2\beta d_{\min} + \pi)}} \quad \text{SWR} \\
 &= \frac{1 - jS \tan \beta d_{\min}}{S - j \tan \beta d_{\min}}
 \end{aligned}$$

Scattering Matrix(S-Parameters)



$$\begin{pmatrix} \mathbf{b}_1 \\ \mathbf{b}_2 \end{pmatrix} = \begin{pmatrix} S_{11} & S_{12} \\ S_{21} & S_{22} \end{pmatrix} \begin{pmatrix} \mathbf{a}_1 \\ \mathbf{a}_2 \end{pmatrix}$$

S-parameters

- Input Reflection Coefficient with $Z_L=Z_0$

$$S_{11} = \left. \frac{b_1}{a_1} \right|_{a_2=0}$$

- Output Reflection Coefficient with $Z_G=Z_0$ and $V_G=0$

$$S_{22} = \left. \frac{b_2}{a_2} \right|_{a_1=0}$$

- Forward Transmission Coefficient with $Z_L=Z_0$

$$S_{21} = \left. \frac{b_2}{a_1} \right|_{a_2=0}$$

- Reverse Transmission Coefficient with $Z_G=Z_0$ and $V_G=0$

$$S_{12} = \left. \frac{b_1}{a_2} \right|_{a_1=0}$$

S-parameters

- Input Reflection Coefficient with $Z_L=Z_0$

$$S_{11} = \left. \frac{b_1}{a_1} \right|_{a_2=0}$$

Input match

- Output Reflection Coefficient with $Z_G=Z_0$ and $V_G=0$

$$S_{22} = \left. \frac{b_2}{a_2} \right|_{a_1=0}$$

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$$S_{21} = \left. \frac{b_2}{a_1} \right|_{a_2=0}$$

Gain or loss

- Reverse Transmission Coefficient with $Z_G=Z_0$ and $V_G=0$

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S-parameters

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Output match

- Forward Transmission Coefficient with $Z_L=Z_0$

$$S_{21} = \left. \frac{b_2}{a_1} \right|_{a_2=0}$$

Gain or loss

- Reverse Transmission Coefficient with $Z_G=Z_0$ and $V_G=0$

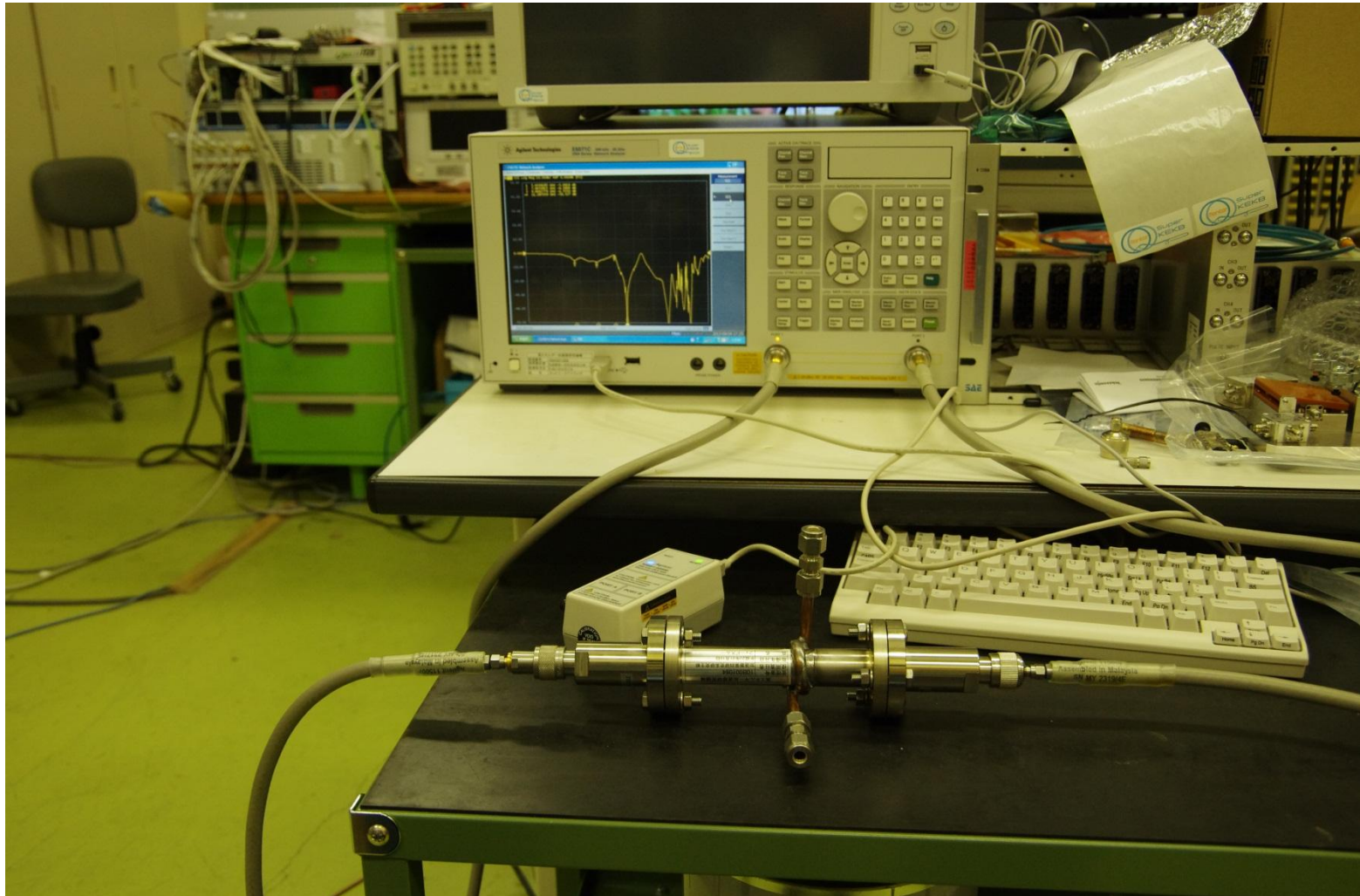
$$S_{12} = \left. \frac{b_1}{a_2} \right|_{a_1=0}$$

Isolation

Why Use S-Parameters?

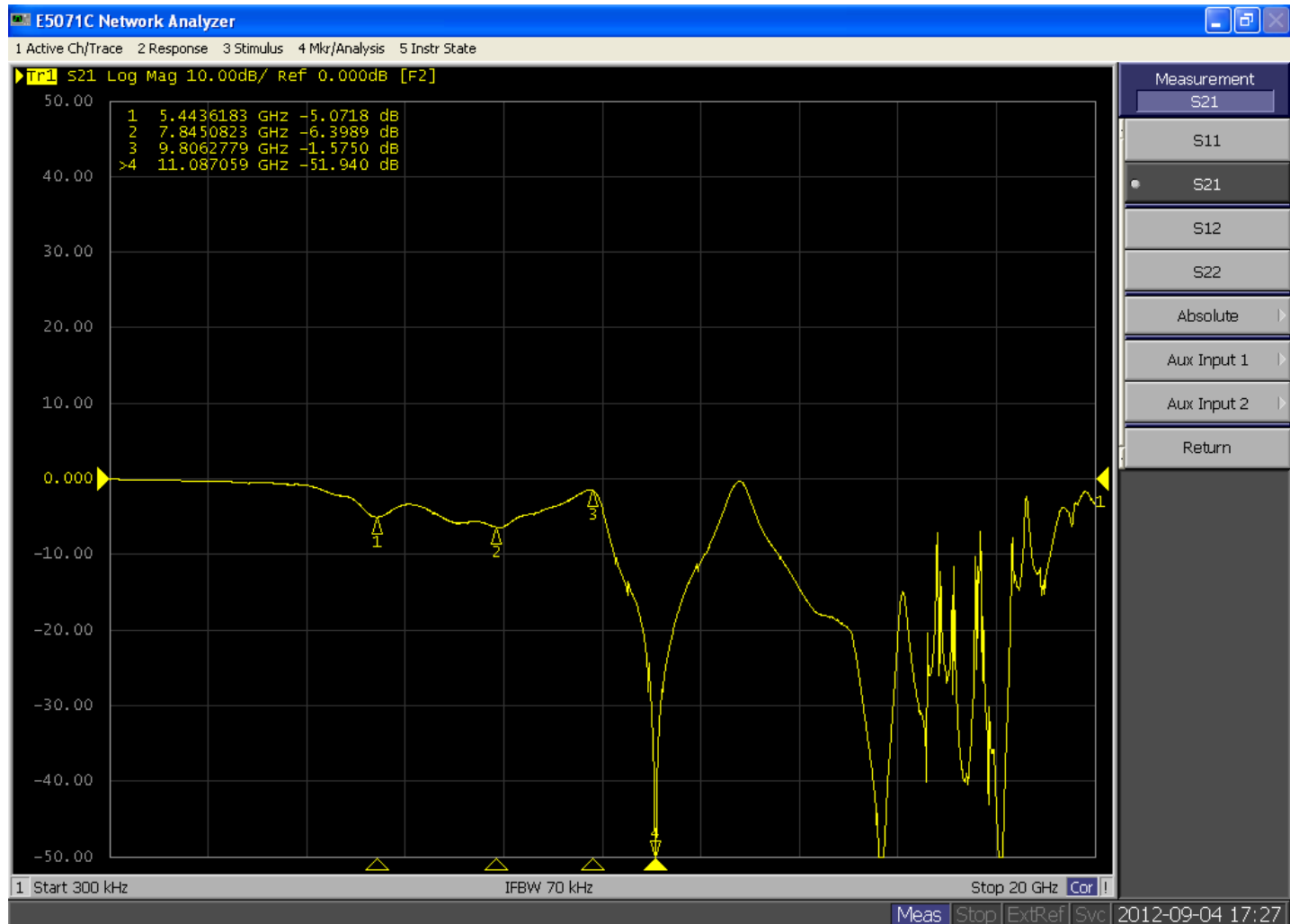
- **relatively easy to obtain at high frequencies**
 - measure voltage traveling waves with a vector network analyzer
 - don't need shorts/opens which can cause active devices to oscillate or self-destruct
- **relate to familiar measurements (gain, loss, reflection coefficients...)**
- **can cascade S-parameters of multiple devices to predict system performance**
- **can compute H, Y or Z parameters from S-parameters if desired**
- **can easily import and use S-parameter files in electronic-simulation tools**

RF Network Analyzer



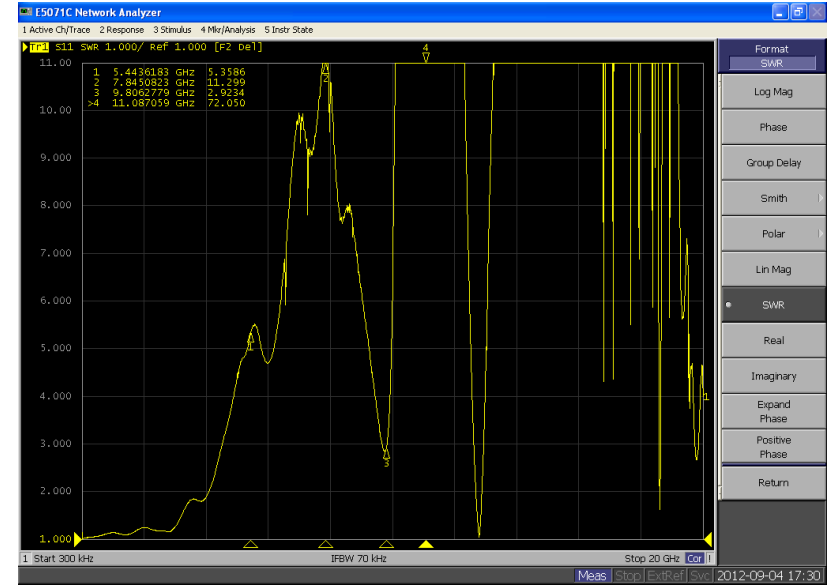
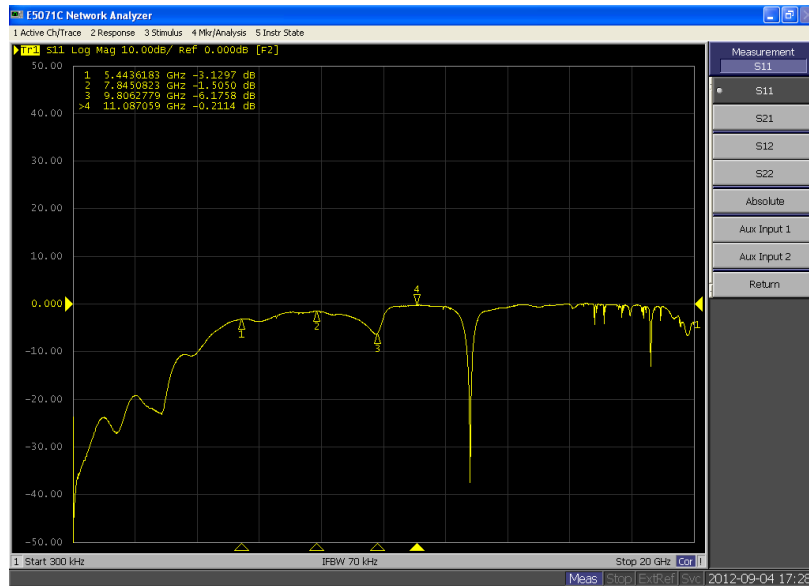
S21

S21(in dB)

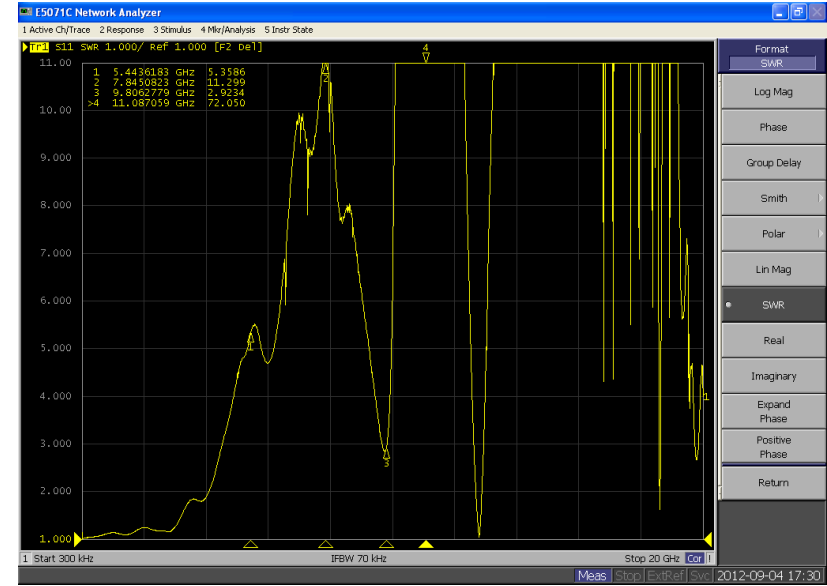
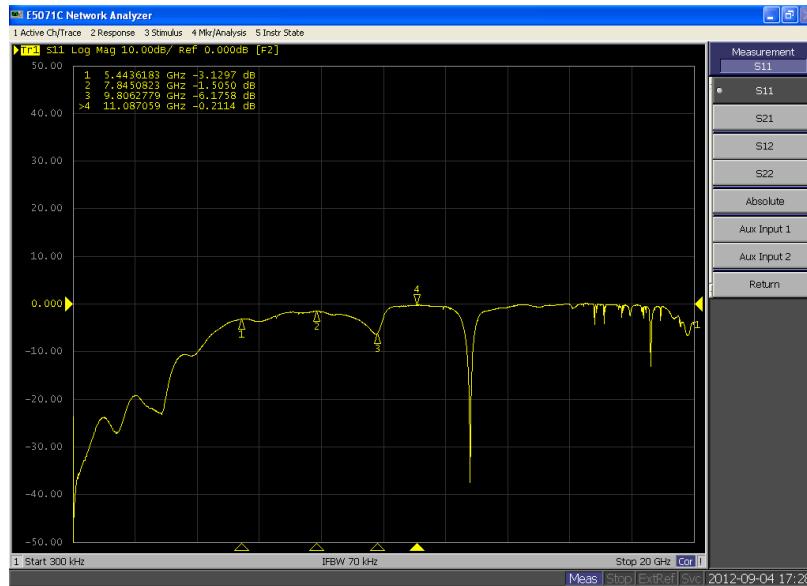


Frequency

S11 and SWR(VSWR)

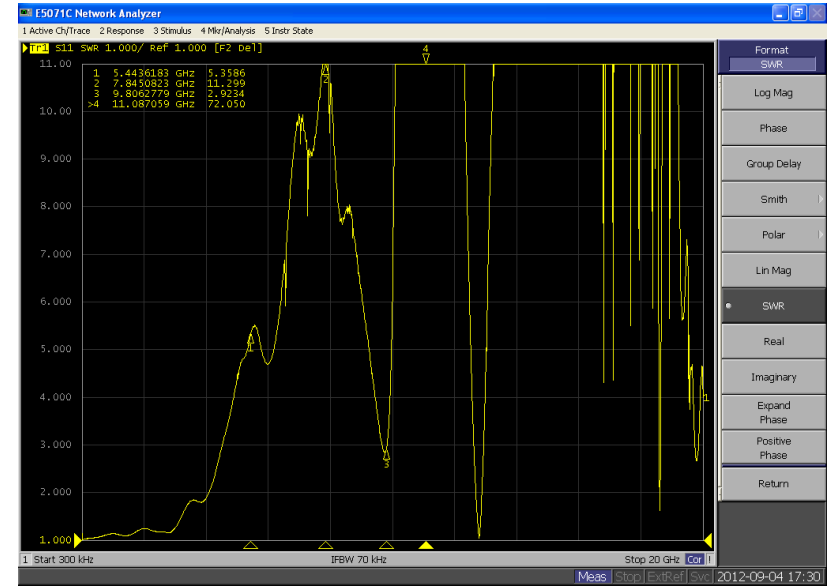
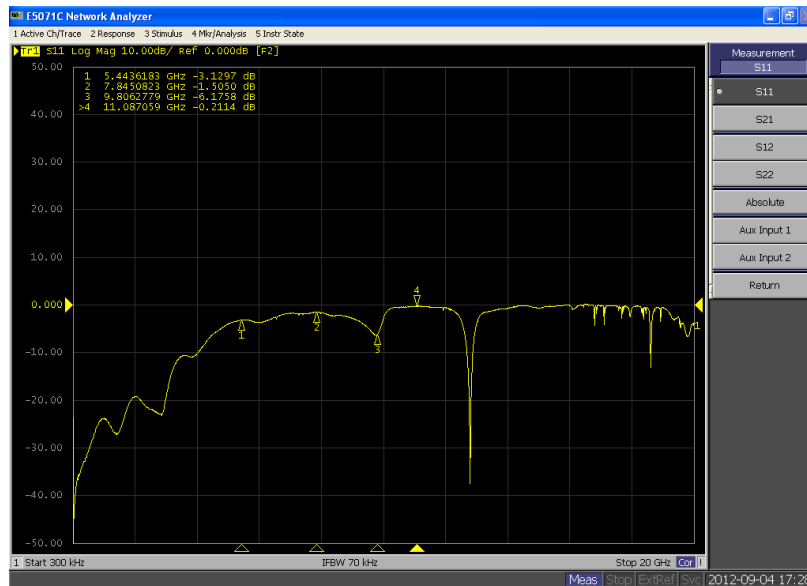


S11 and SWR(VSWR)



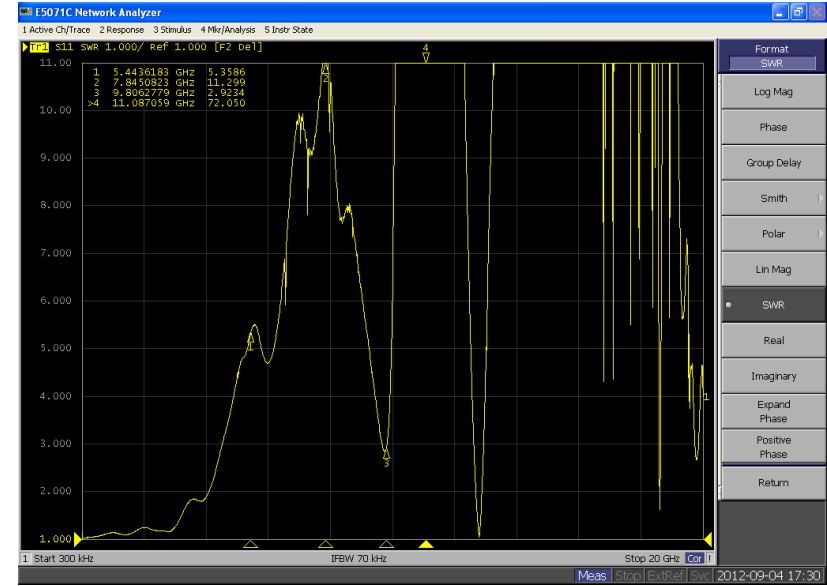
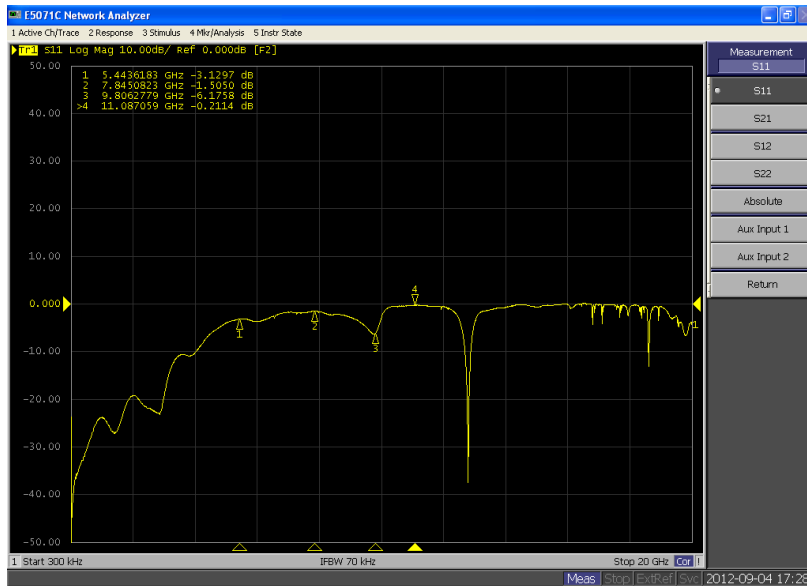
$$20 \log_{10} |S_{11}| \text{ dB}$$

S11 and SWR(VSWR)



- Return loss
- $$20 \log_{10} |S_{11}| \text{ dB}$$

S11 and SWR(VSWR)

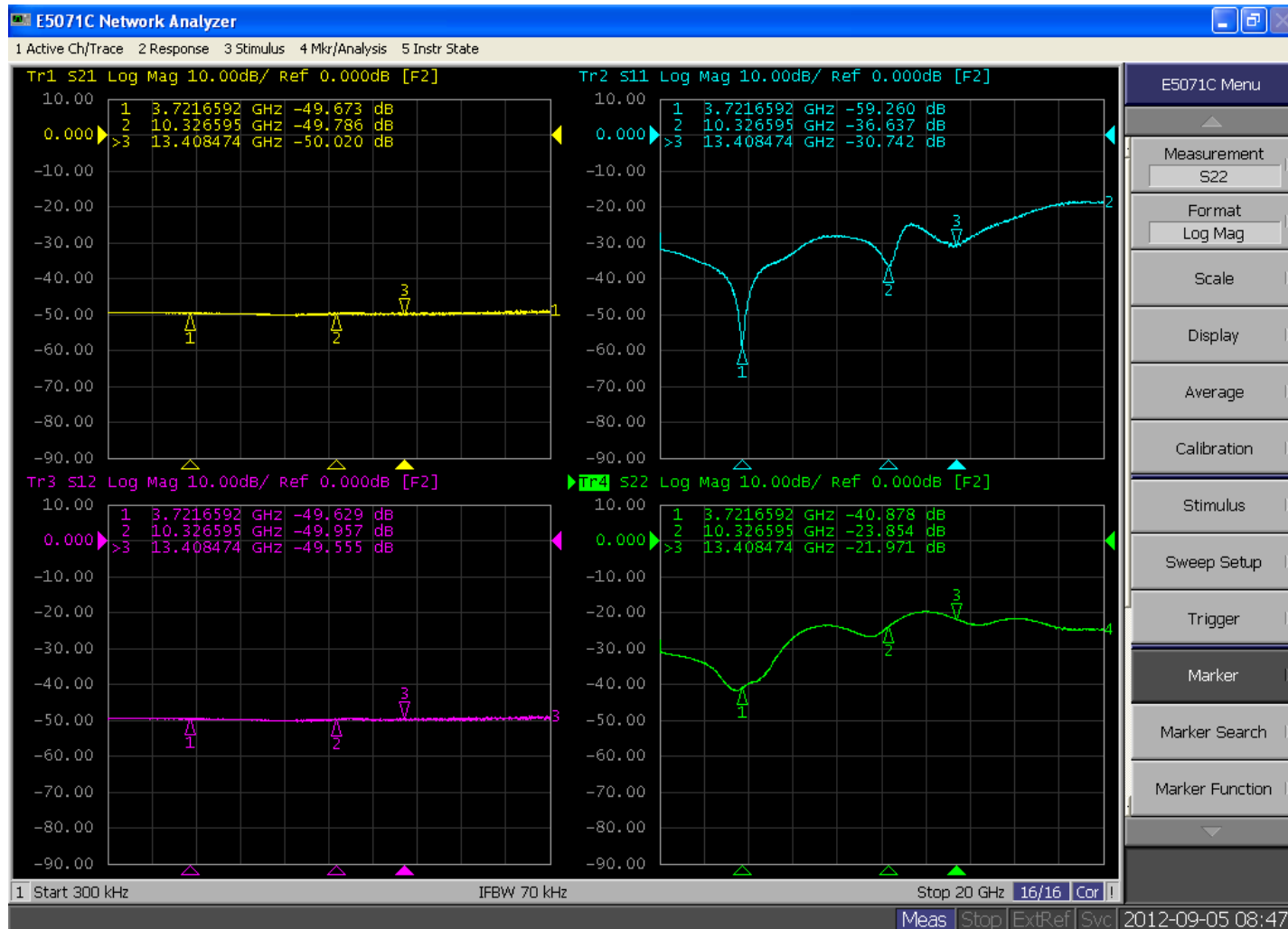


- Return loss

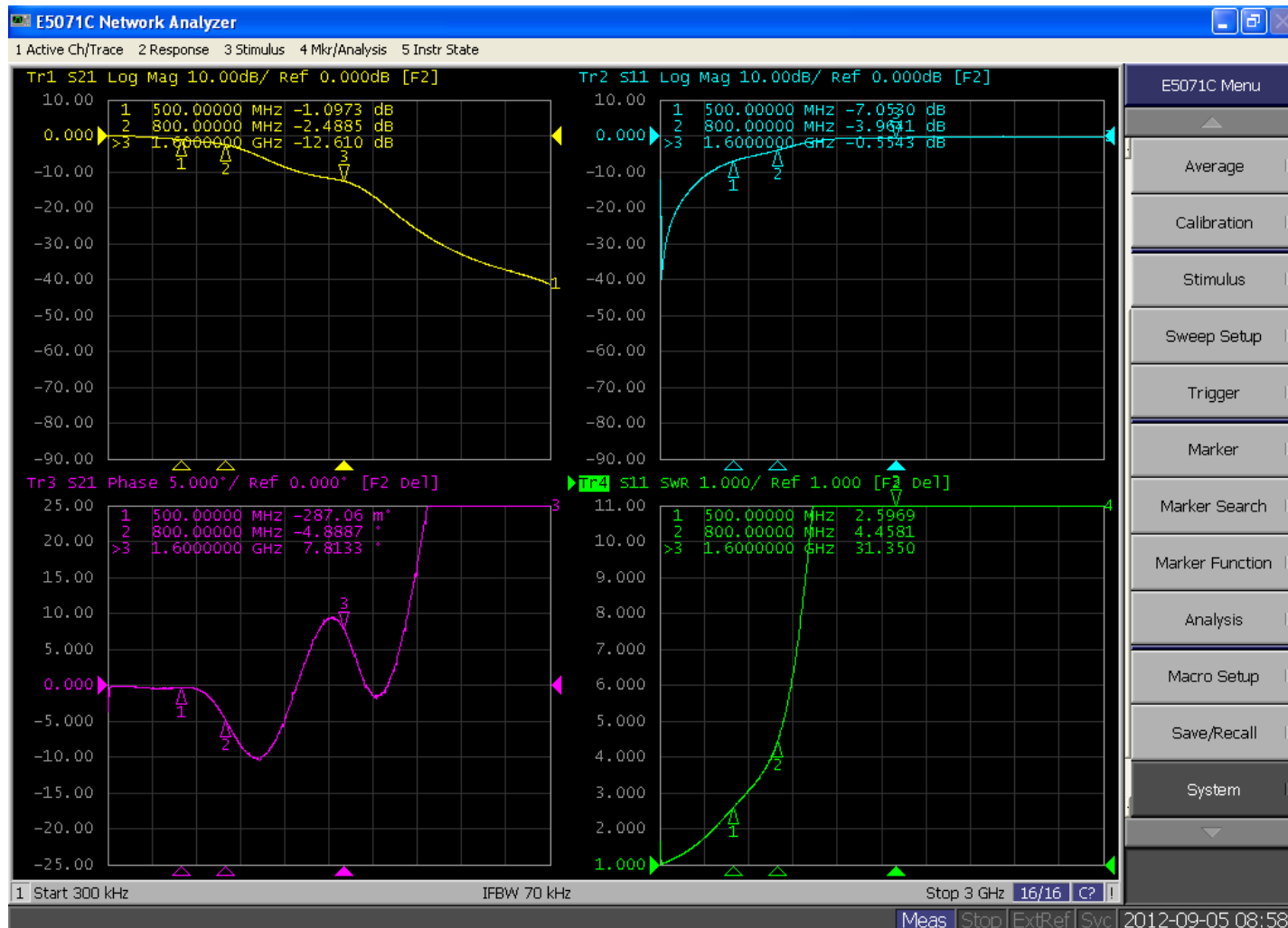
$$20 \log_{10} |S_{11}| \text{ dB}$$

$$SWR_{input} = \frac{1 + |S_{11}|}{1 - |S_{11}|}$$

Example (50dB attenuator)



Example (800MHz Bessel LPF)



Time domain behavior--TDR

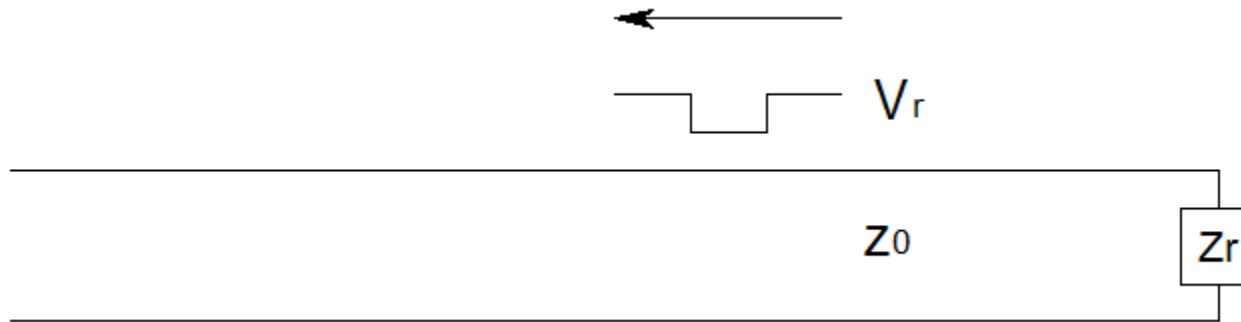


- Using voltage ratio between forward and reflecting wave

$$Z_r = \frac{1 + \rho}{1 - \rho} Z_0$$

$$\rho = \frac{V_r}{V_i}$$

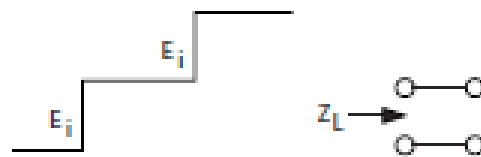
Time domain behavior--TDR



- Using voltage ratio between forward and reflecting wave

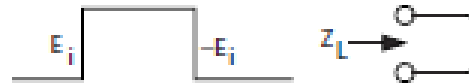
$$Z_r = \frac{1 + \rho}{1 - \rho} Z_0$$

$$\rho = \frac{V_r}{V_i}$$

(A) Open Circuit Termination ($Z_L = \infty$)

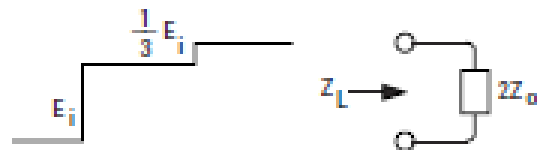
$$(A) E_r = E_i$$

Therefore $\frac{Z_L - Z_0}{Z_L + Z_0} = +1$
 Which is true as $Z_L \rightarrow \infty$
 $\therefore Z = \text{Open Circuit}$

(B) Short Circuit Termination ($Z_L = 0$)

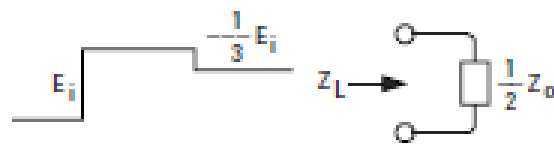
$$(B) E_r = -E_i$$

Therefore $\frac{Z_L - Z_0}{Z_L + Z_0} = -1$
 Which is only true for finite Z
 When $Z_L = 0$
 $\therefore Z = \text{Short Circuit}$

(C) Line Terminated in $Z_L = 2Z_0$

$$(C) E_r = +\frac{1}{3} E_i$$

Therefore $\frac{Z_L - Z_0}{Z_L + Z_0} = +\frac{1}{3}$
 and $Z_L = 2Z_0$

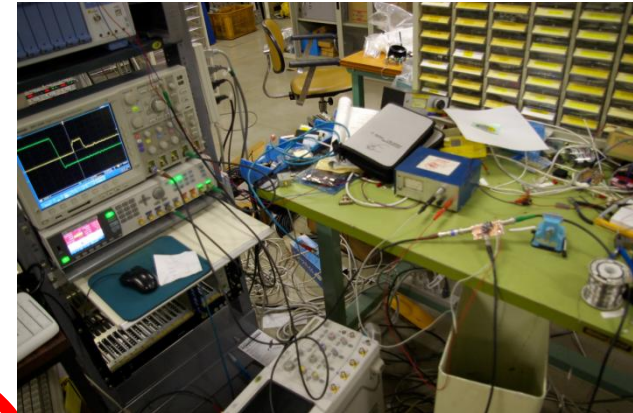
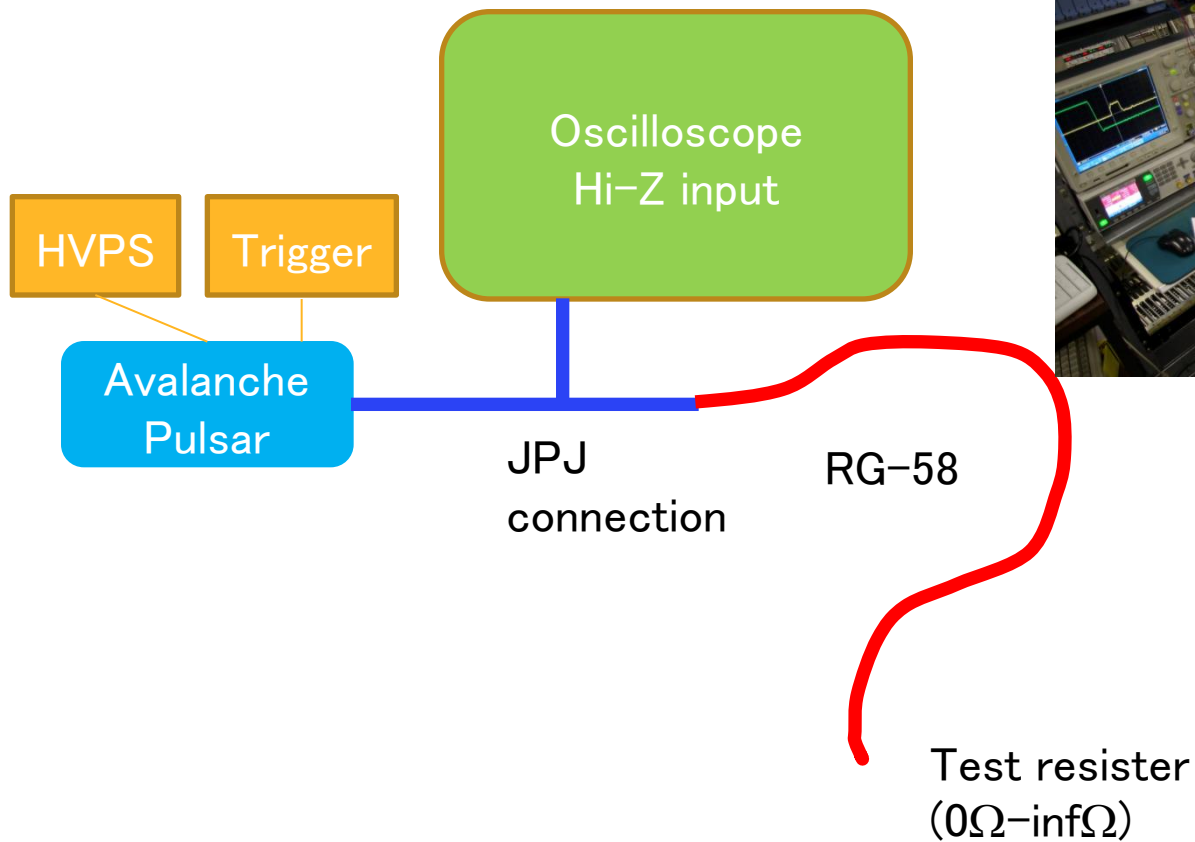
(D) Line Terminated in $Z_L = \frac{1}{2} Z_0$

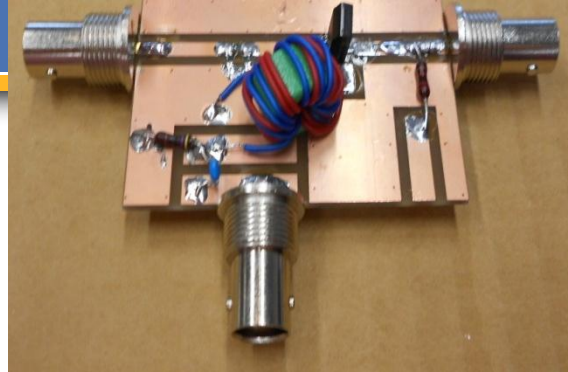
$$(D) E_r = -\frac{1}{3} E_i$$

Therefore $\frac{Z_L - Z_0}{Z_L + Z_0} = -\frac{1}{3}$
 and $Z_L = \frac{1}{2} Z_0$

Science Camp 2011 (for High School students)



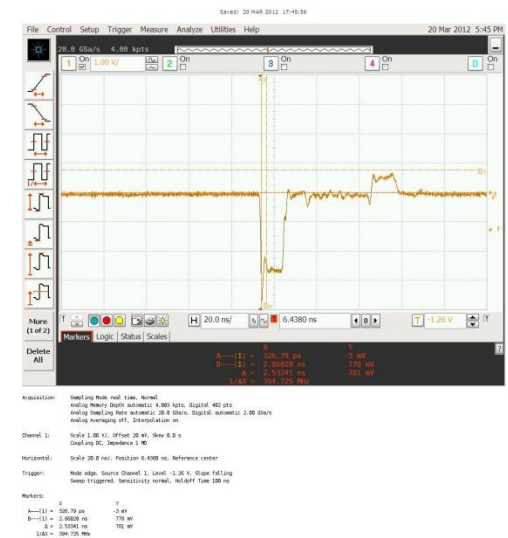




51Ω



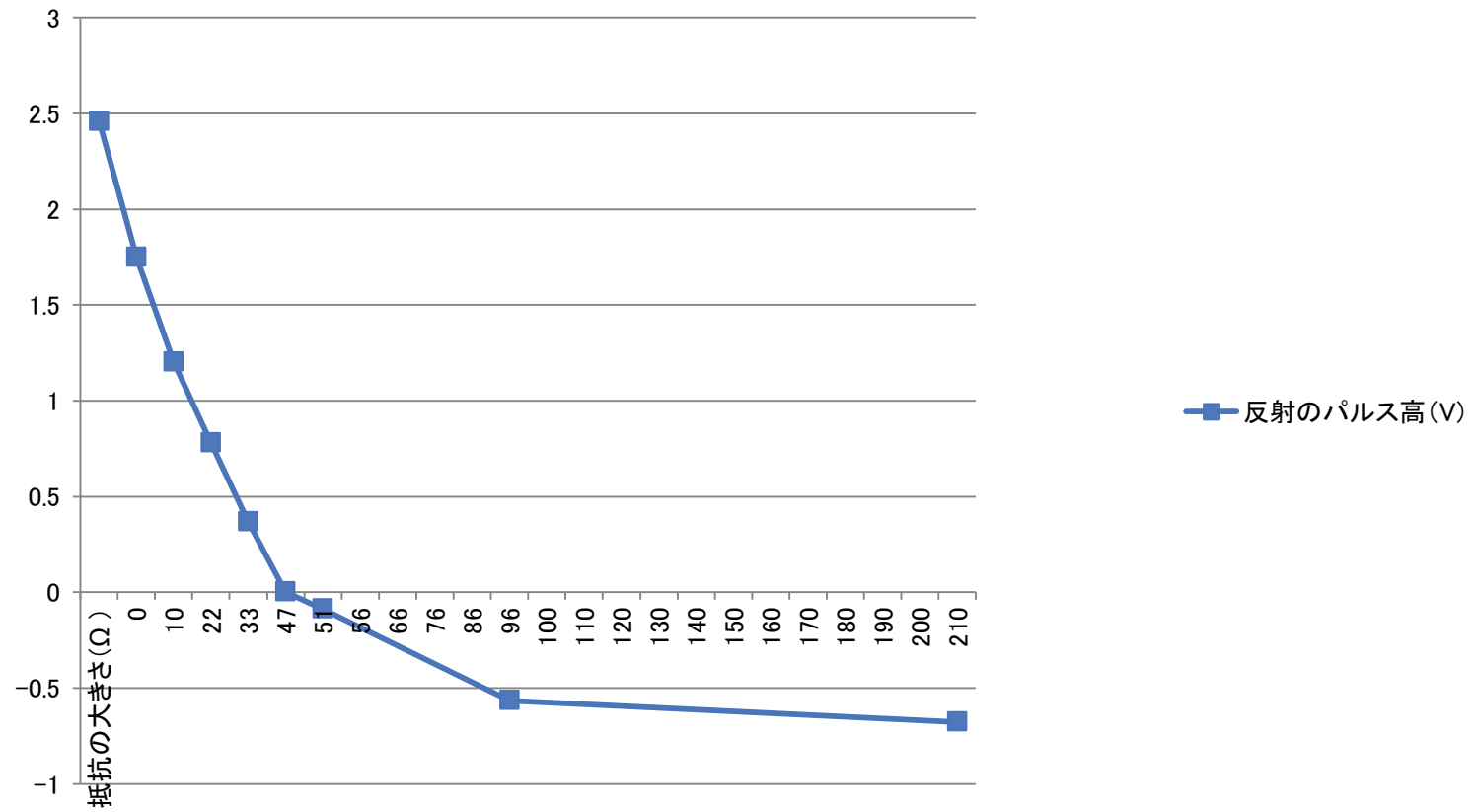
470Ω



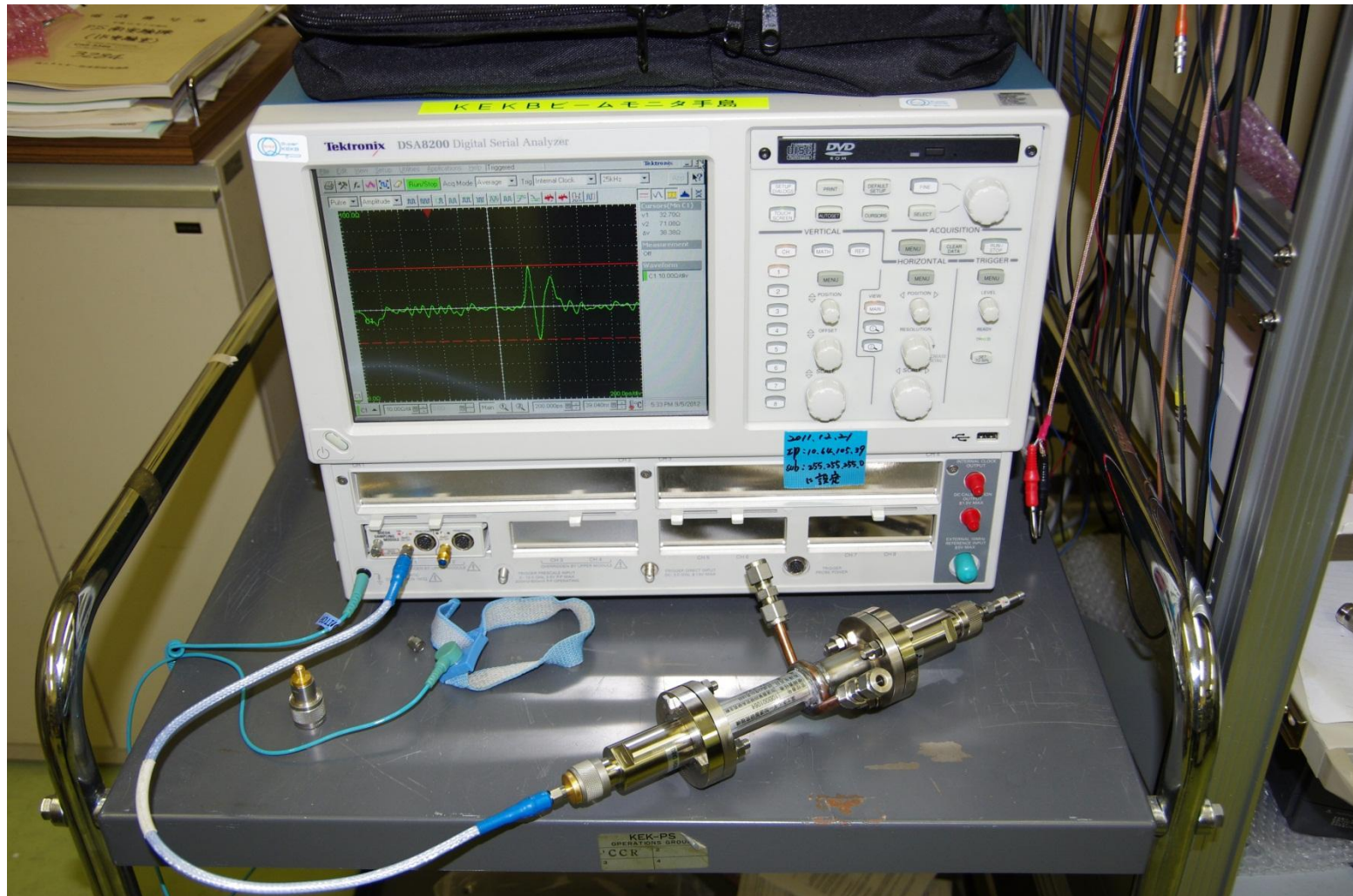
33Ω

Matching of 50-ohm line

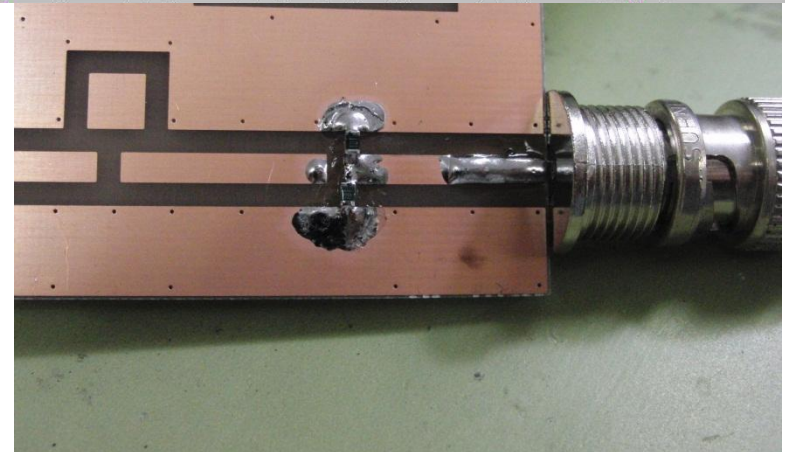
Reflection (V)



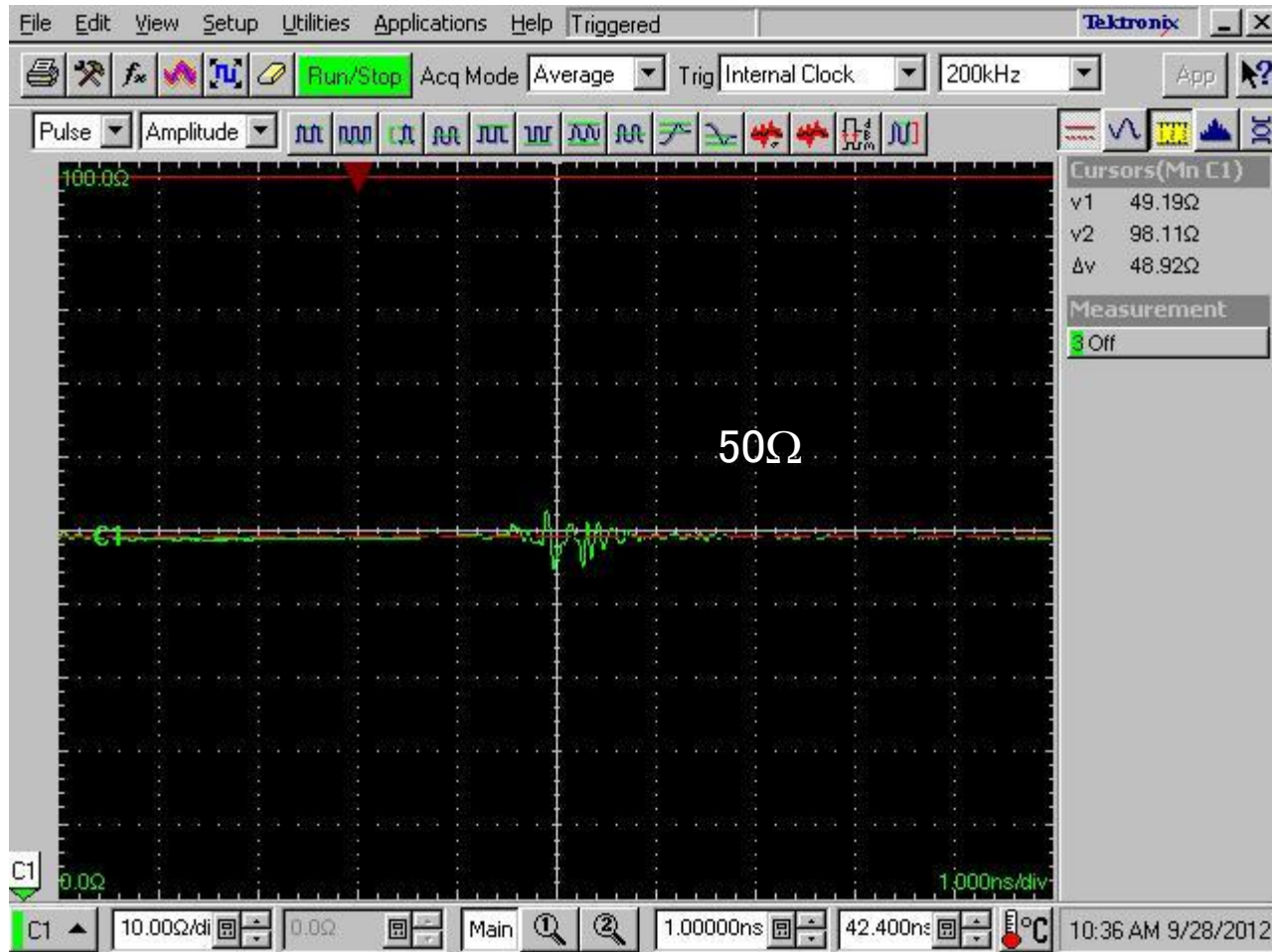
Real TDR



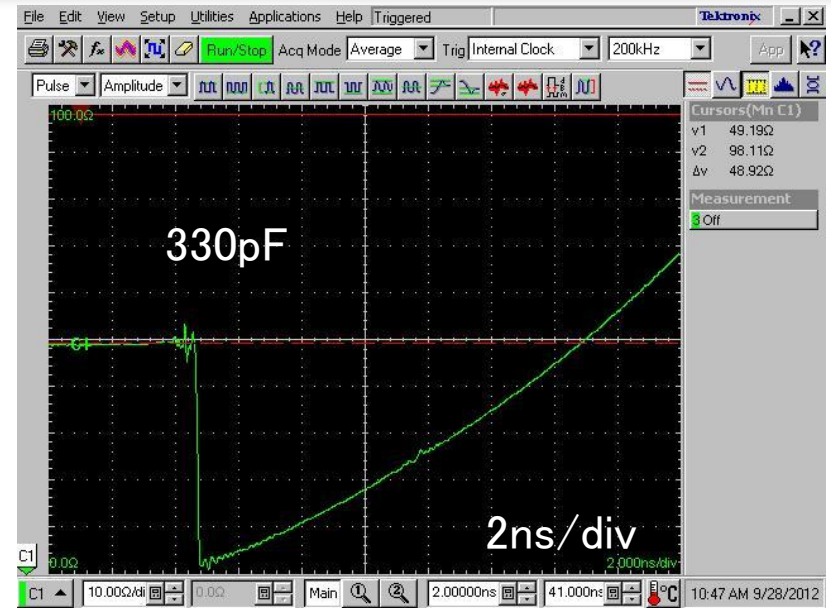
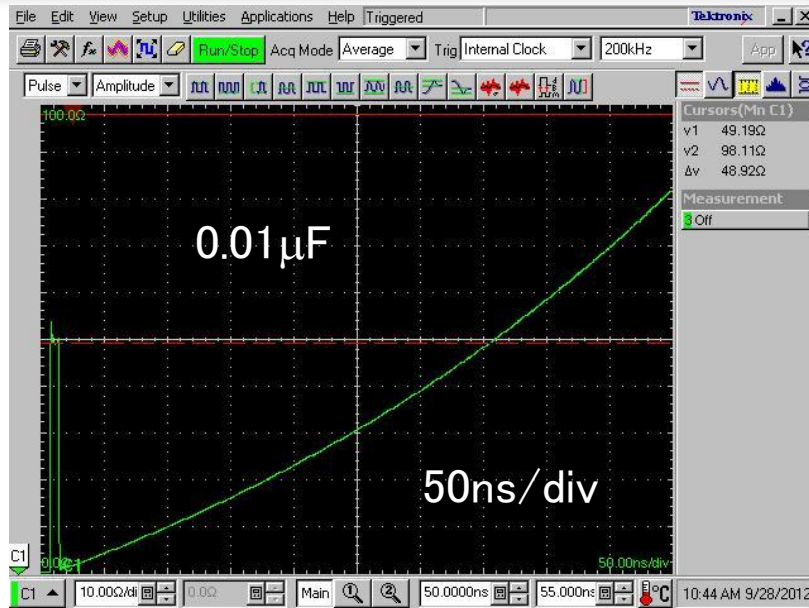
TDR (chip resister)



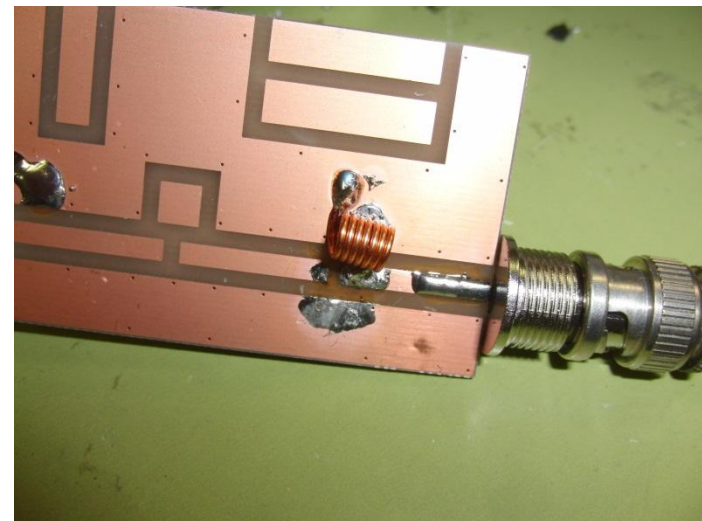
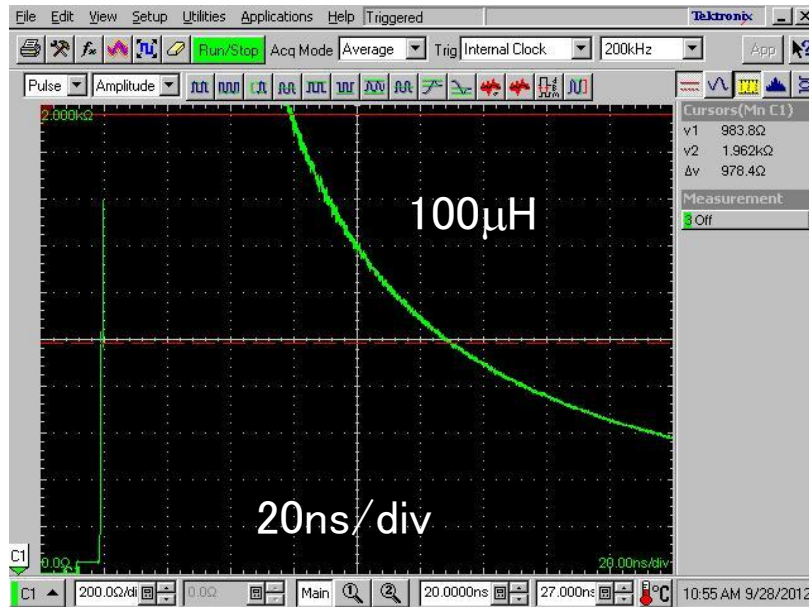
TDR



Capacitor



Inductance



Tips

- **Good matched two-ports (such as J-J structure) is needed to measure the S-parameters of the feedthrough structure.**
 - For the TDR measurements, open structure (button or rod on vacuum side) might usually be OK



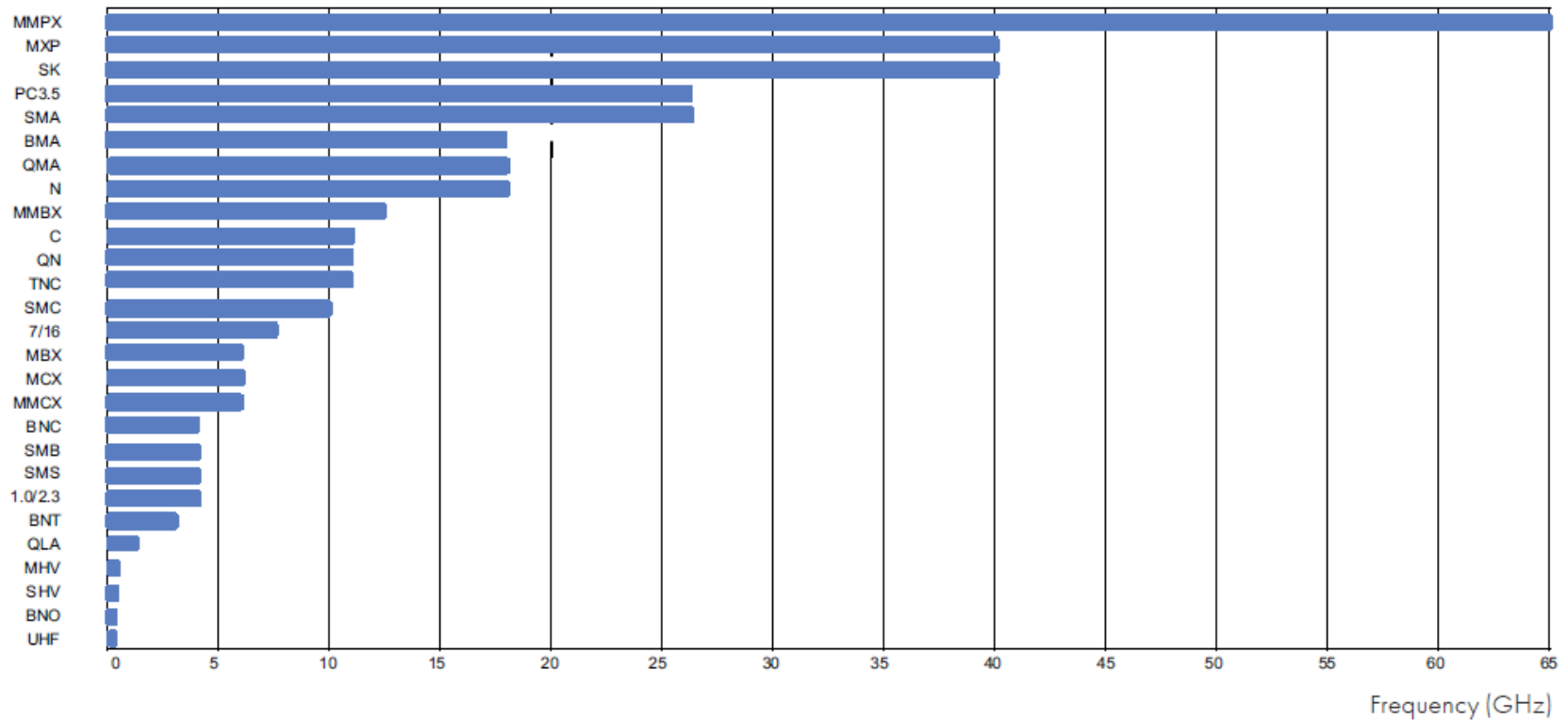
Ideal feedthrough?

- **Mechanical**
 - Completely satisfy required mechanical toughness
 - Satisfy vacuum requirements

- **Frequency domain**
 - S_{21} (and S_{12}) ~ 1 (0dB)
 - S_{11} (and S_{22}) $\sim -\infty$ dB

- **Time domain**
 - Impedance=50 Ohm, completely matched structure

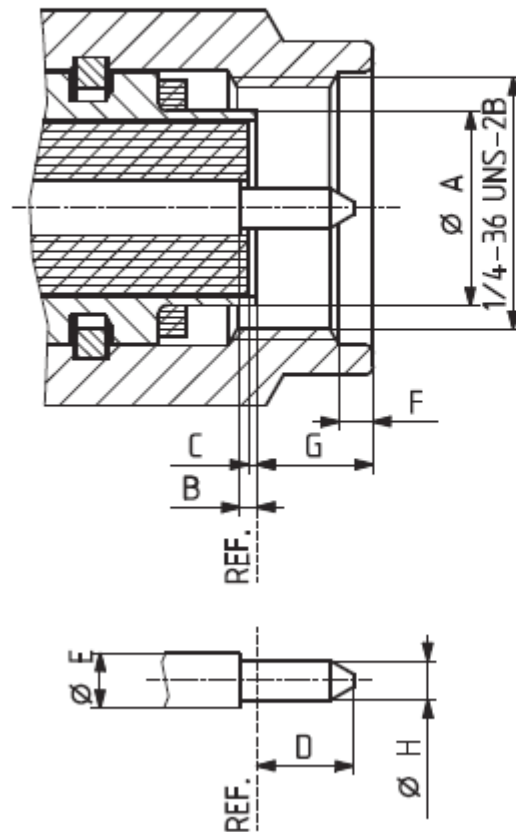
RF connector



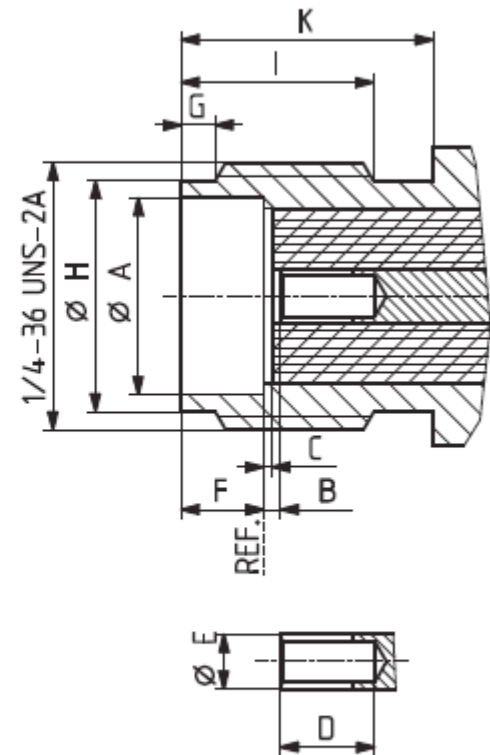
Frequency range of connector series.

Plug(male) and Jack(female)

Plug (male)

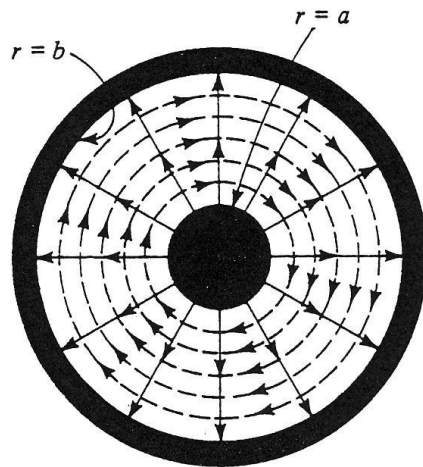


Jack (female)



Size of the feedthrough

- Which connectors do you want to use?
 - N?, SMA?, BNC?, 7/16-DIN?, 7/8-EIA?



—→ *E* lines
 - - -→ *H* lines

$$Z_0 = 60 \sqrt{\frac{\mu_R}{\epsilon_R}} \ln \frac{b}{a}$$

Attenuation Length (dB/m)

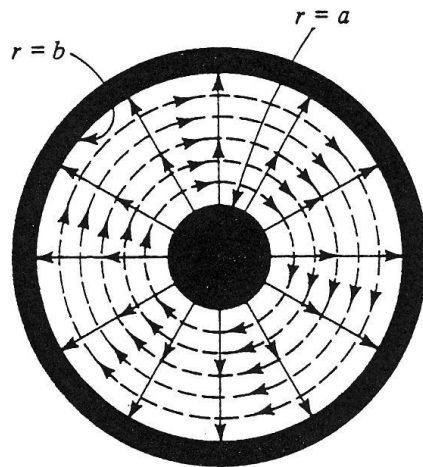
$$\alpha \approx \alpha_c + \alpha_d$$

$$\alpha_c = 13.6 \frac{\delta_s \sqrt{\epsilon_R} \{1 + (b/a)\}}{\lambda_0 b \ln(b/a)},$$

$$\alpha_d = 27.3 \frac{\sqrt{\epsilon_R}}{\lambda_0} \tan \delta$$

Size of the feedthrough

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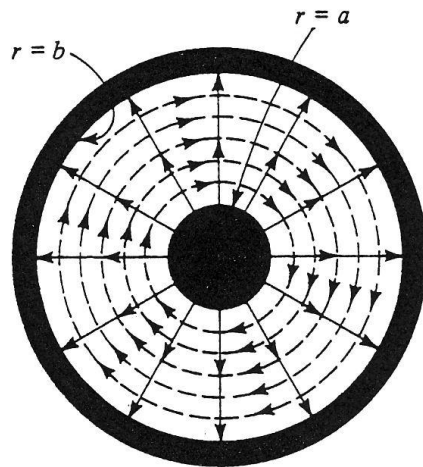
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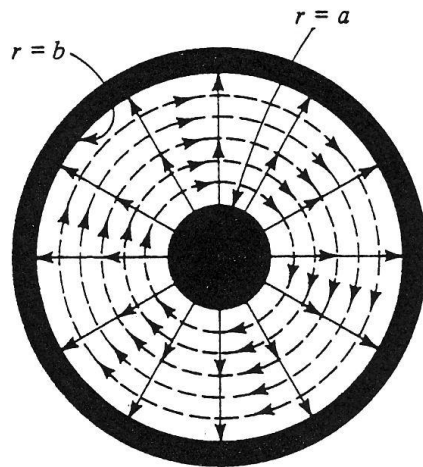
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Size of the feedthrough

- Which connectors do you want to use?
 - N?, SMA?, BNC?, 7/16-DIN?, 7/8-EIA?



—→ *E* lines
 - - -→ *H* lines

$$Z_0 = 60 \sqrt{\frac{\mu_R}{\epsilon_R}} \ln \frac{b}{a}$$

Attenuation Length (dB/m)

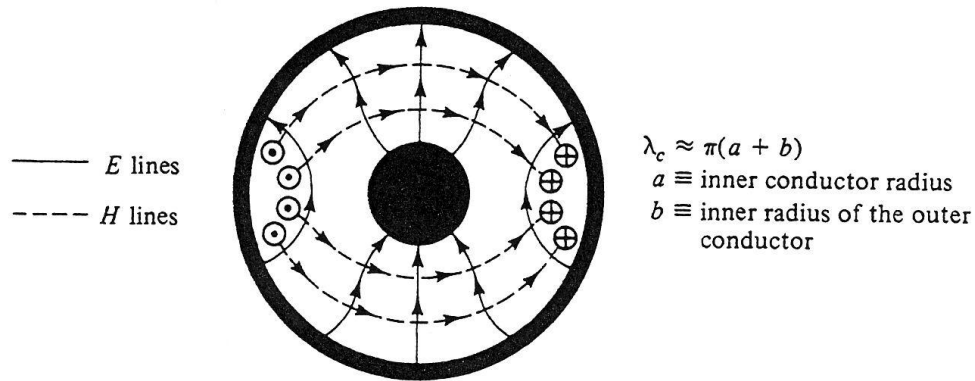
$$\alpha \approx \alpha_c + \alpha_d$$

$$\alpha_c = 13.6 \frac{\delta_s \sqrt{\epsilon_R} \{1 + (b/a)\}}{\lambda_0 b \ln(b/a)},$$

$$\alpha_d = 27.3 \frac{\sqrt{\epsilon_R}}{\lambda_0} \tan \delta$$

Higher-mode propagation in coaxial lines.

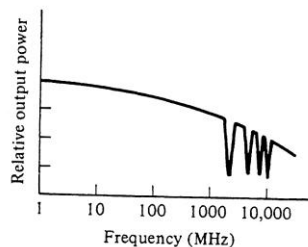
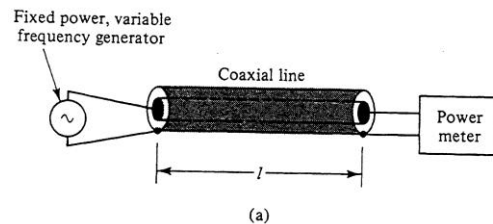
■ TE₁₁mode



$$\lambda_c \approx \pi(a + b)$$

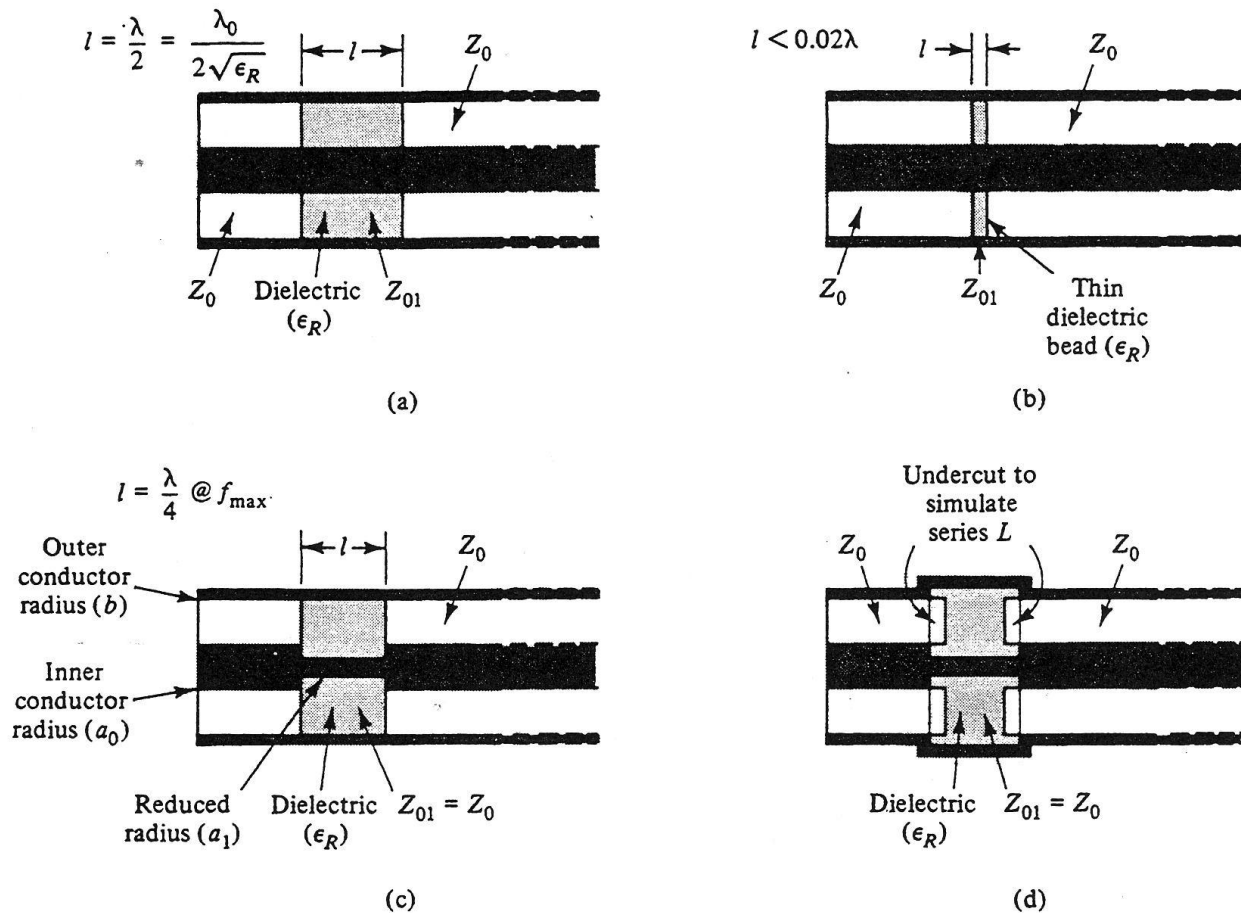
$$f_c \approx \frac{c}{\pi(a + b)\sqrt{\epsilon_R}}$$

Figure 5-3 The electromagnetic field pattern for the TE_{11} mode in a coaxial line.



- Type-N: 18GHz
- SMA: 34GHz
- RG-223: ~30GHz
- EIA 7/8": ~6.8GHz

Connectors in the air



$$Z_{01} = \frac{Z_0}{\sqrt{\epsilon_R}}$$

Figure 6-8 Four types of dielectric bead supports for coaxial lines.

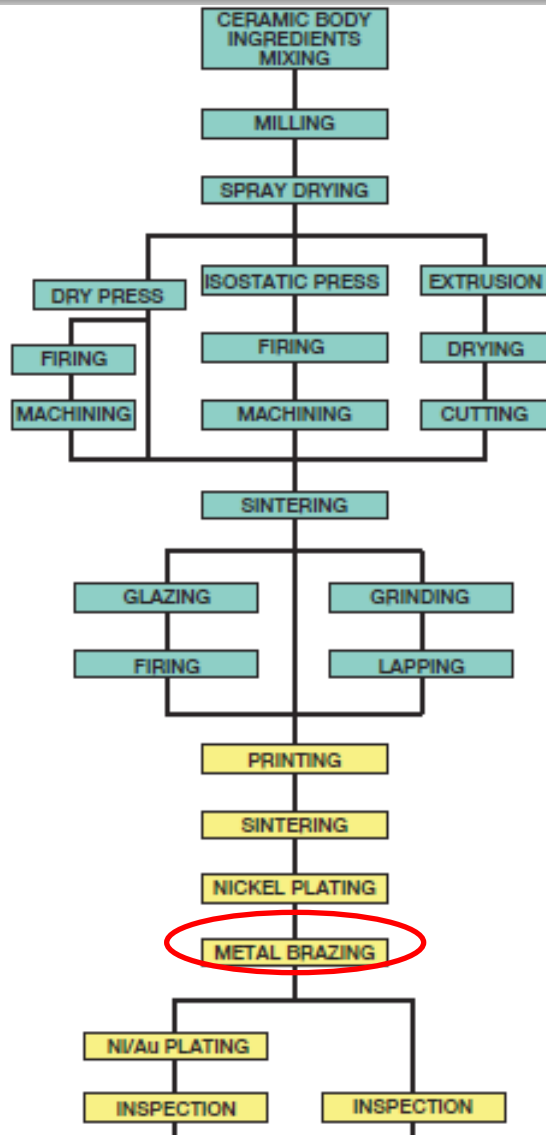
Feedthrough for vacuum

- **Vacuum seal (and support for inner conductor)**
 - Alumina ceramic ($\epsilon_R \sim 10$), Si_3N_4 (~ 8)
 - Glass ($\epsilon_R \sim 4-5$)
 - Much larger than the support used in the air (~ 2)
- **Need to stand mechanical stress coming from**
 - Vacuum pressure
 - (huge) pressure when connecting the cable (or attenuator) to the feedthrough
 - Thermal stress
 - Baking
 - Heating by the beam power
- **Need to keep good RF contact under severe condition**
 - Heat cycle
 - Radiation, active gas

Outer/inner conductor

- **Kovar (Nickel–cobalt ferrous alloy)**
 - Trademark of Carpenter Technology Corporation
 - Designed to be compatible with the thermal expansion characteristics of borosilicate glass
 - Ferromagnetics: Not so good for use near strong magnet.
 - Thermal conductivity : Low
- **Ti**
 - Non-magnetic material
- **Cu, Al, Stainless steel..**
 - Also be usable. Consult your ceramic company.

Feedthrough process



Brazing temperature

- ~800 degC
- After brazing process, metal parts will be well annealed – original characteristics might be changed.
 - No spring action at all!

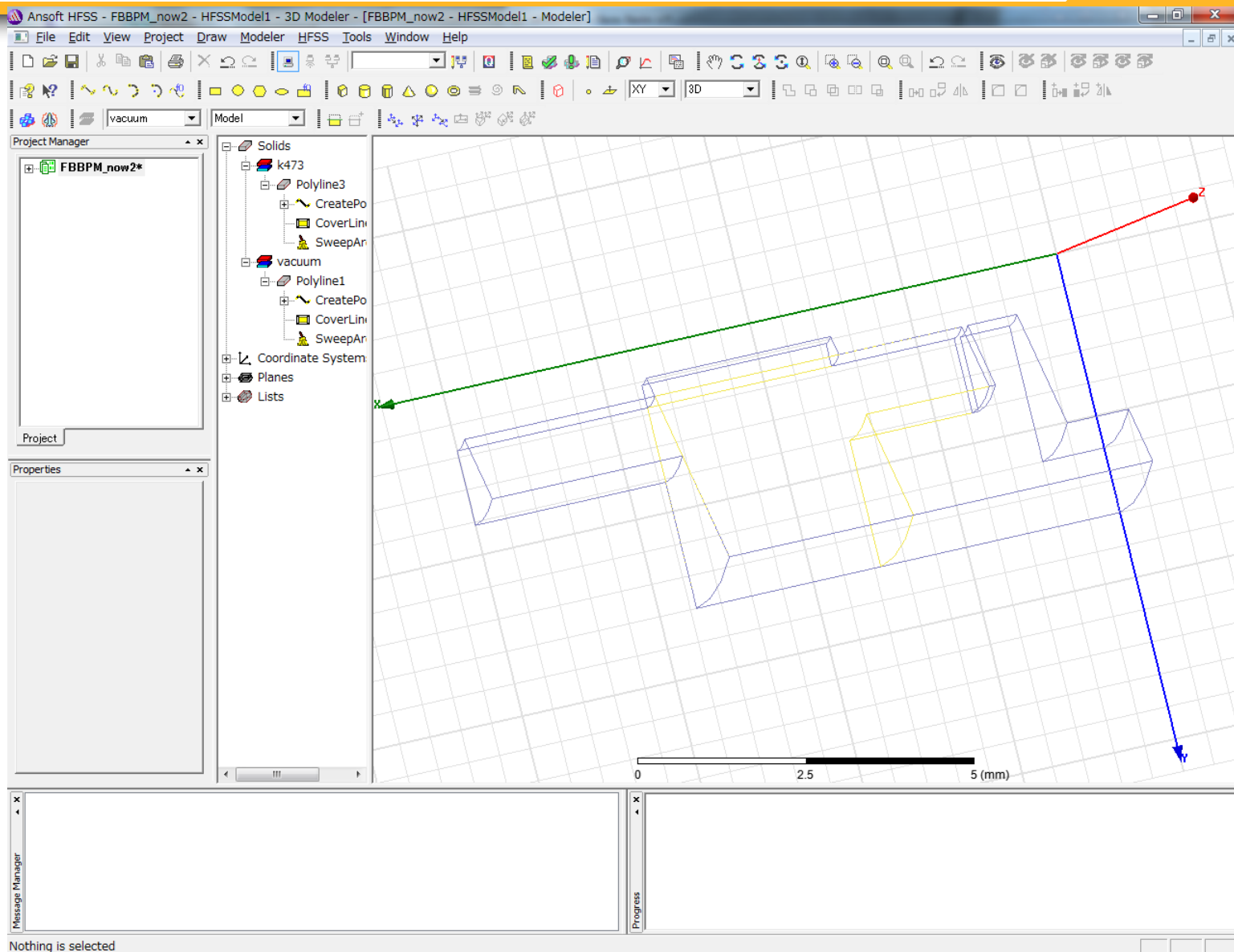
Braze to vacuum chamber OK?

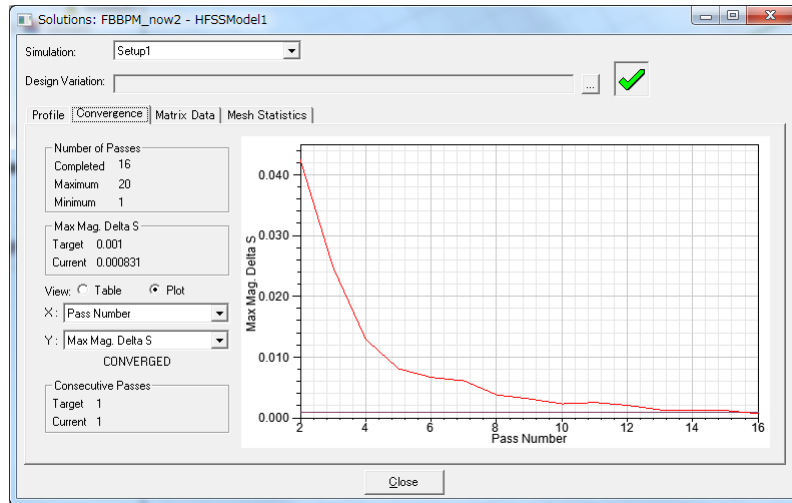
- Second (third?) brazing process
- There exist brazing materials with lower temperature
 - Not so easy to control the temperature

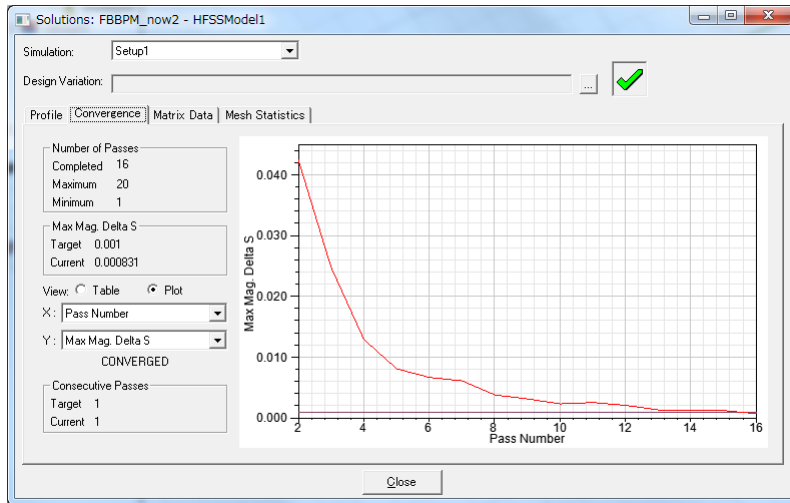
Frequency domain design

- **ANSYS (Ansoft) HFSS**
 - 3D full-wave electromagnetic field simulator
 - Multiple state-of-the-art solver (finite element method or integral equation method)
 - calculate S-parameters, field patterns, eigen values, impedances..
 - Good user interface, Good interface to other tools such as ANSYS, etc.
 - Automatic mesh handling
 - (Very) Fast simulation speed

Example







Solutions: FBBPM_now2 - HFSSModel1

Simulation: Setup1

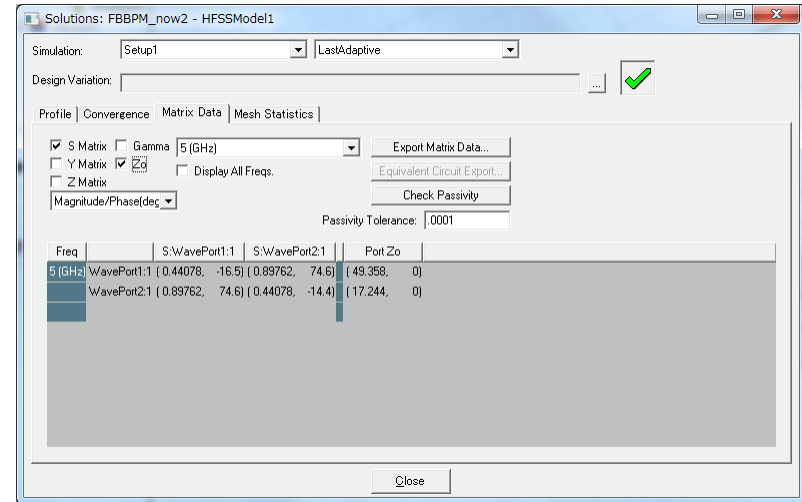
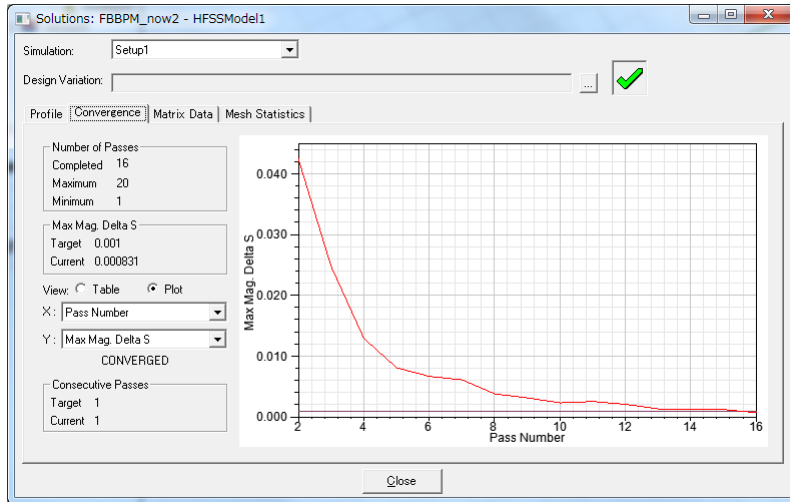
Design Variation:

Profile | Convergence | Matrix Data | Mesh Statistics

Task	Real Time	CPU Time	Memory	Information
Frequency: 21.25 GHz				Full Solution # 19
Simulation Setup	00:00:00	00:00:00	35.4 M	Disk = 0 KBytes
Matrix Assembly	00:00:00	00:00:00	61 M	Disk = 0 KBytes, 9601 tetrahedra, WavePort1: 84 triangles, WavePort2: 90 triangles
Solver MRS4	00:00:01	00:00:02	130 M	Disk = 0 KBytes, matrix size 61092, matrix bandwidth 20.7
Field Recovery	00:00:00	00:00:00	130 M	Disk = 0 KBytes, 2 excitations
				Interpolation Error: S Matrix error 0 %
				Interpolating sweep converged
Solution Process				Elapsed time : 00:00:39, Hfss ConEngine Memory : 46.7 M
Total	00:00:21	00:00:40		Time: 09/10/2012 11:16:56, Status: Normal Completion

Export...

Close



Solutions: FBBPM_now2 - HFSSModel1

Simulation: Setup1

Design Variation: [] [✓]

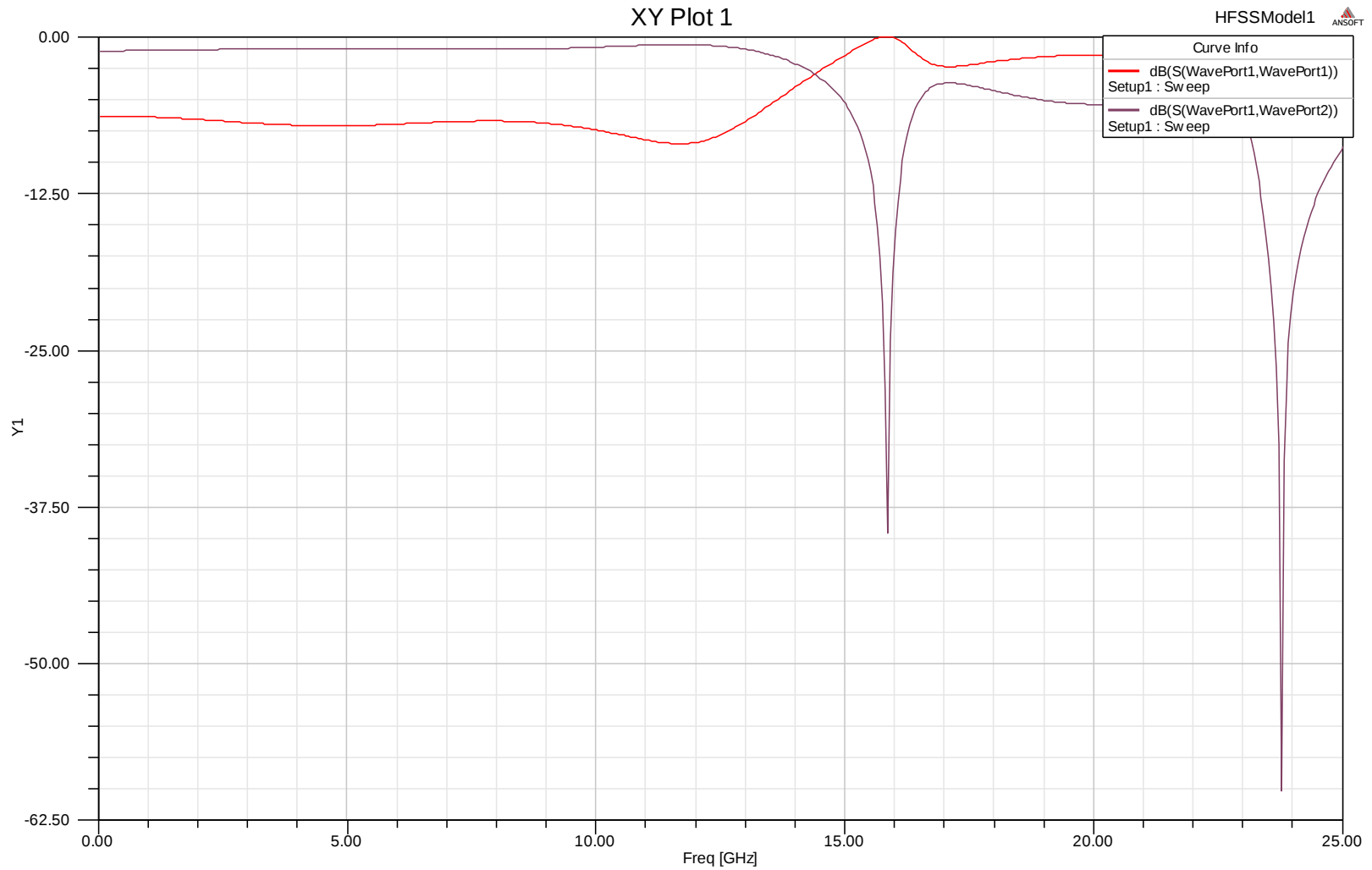
Profile | Convergence | Matrix Data | Mesh Statistics

Task	Real Time	CPU Time	Memory	Information
Frequency: 21.25 GHz				Full Solution # 19
Simulation Setup	00:00:00	00:00:00	35.4 M	Disk = 0 KBytes
Matrix Assembly	00:00:00	00:00:00	61 M	Disk = 0 KBytes, 9601 tetrahedra, WavePort1: 84 triangles, WavePort2: 90 tri
Solver MRS4	00:00:01	00:00:02	130 M	Disk = 0 KBytes, matrix size 61092, matrix bandwidth 20.7
Field Recovery	00:00:00	00:00:00	130 M	Disk = 0 KBytes, 2 excitations
				Interpolation Error: S Matrix error 0 %
				Interpolating sweep converged
Solution Process				Elapsed time : 00:00:39, Hfss ConEngine Memory : 46.7 M
Total	00:00:21	00:00:40		Time: 09/10/2012 11:16:56, Status: Normal Completion

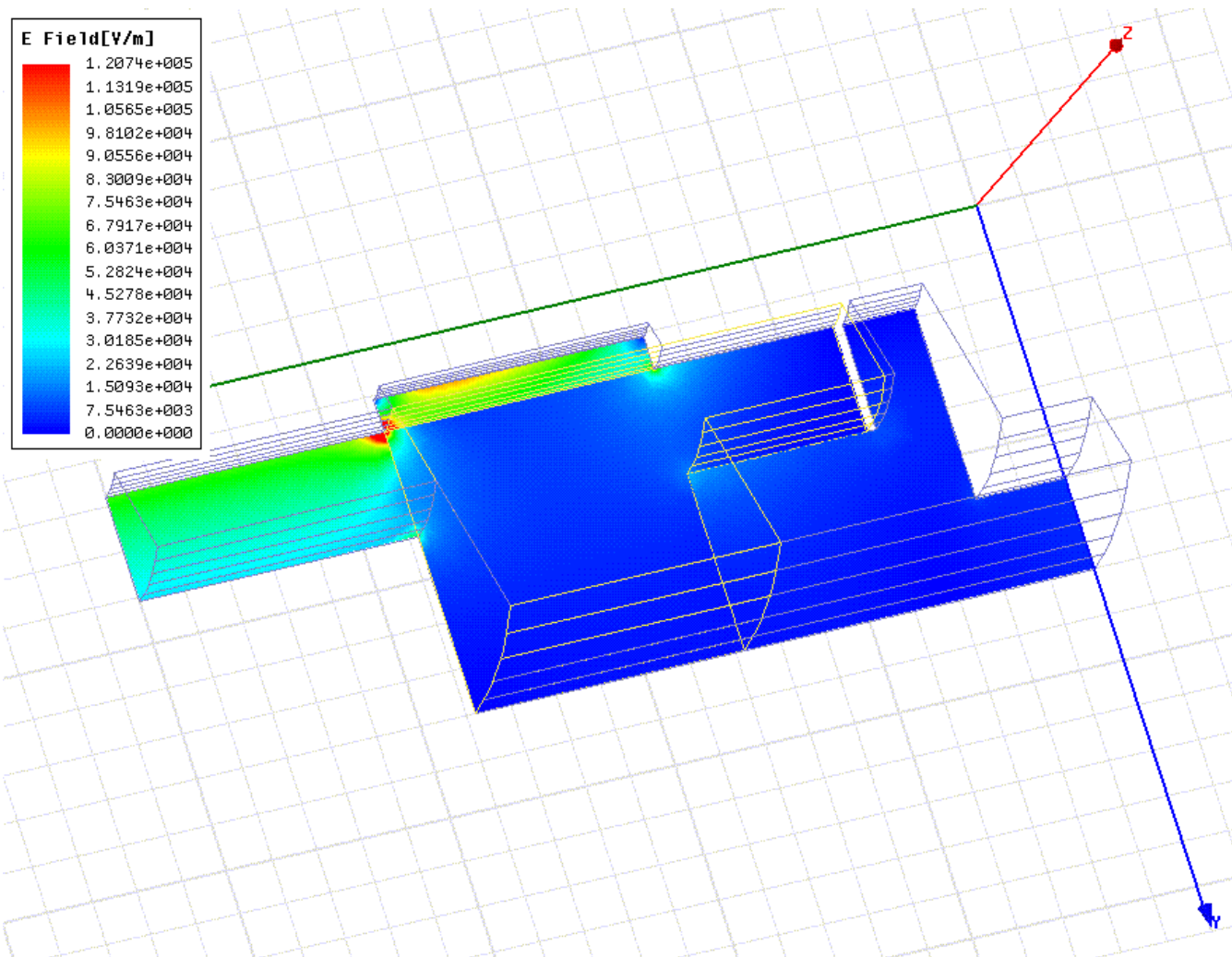
Export...

Close

(Fast) Frequency sweep



E-Field pattern



Time domain simulation

- **GdfidL Electromagnetic Field simulator**
 - Time dependent Fields in loss-free or lossy structure
 - Fields may be excited by port mode or relativistic line charge
 - Resonant fields in loss-free or lossy structure
- **Calculated result**
 - S-parameters, including power (voltage) value
 - Wake potentials including loss factor, impedance etc.
 - Q values and Shunt impedance
- **Need (some)large-scale parallel computing resources**
 - KEKB: 256 cores, 512GB Linux cluster
 - Text-base user interface, longer simulation time

Example of GdfidL input(1)

```

define(EL,1) define(MAG,2)
define(INF, 1000)

define(PipeRadius, 25.40e-3/2)
define(PipeStart, 100.0e-3)
define(PipeEnd,100.0e-3)
define(BpmAdd,13.5e-3)
define(STPSIZE,0.1e-3)

-general
  outfile= /users/tobiyama/GD/bpm-CL22/temp/
  scratchbase =/users/tobiyama/GD/scratch/
-material
  material=10
  type=electric
  material=15
  type=normal, epsr=9.7, muer=1, kappa=0, mkappa=0

```

```

-mesh
  spacing=0.1e-3
  pxlow= 0
  pxhigh=(PipeRadius+BpmAdd-5*STPSIZE)
  pylow= 0
  pyhigh=(PipeRadius+BpmAdd-5*STPSIZE)
  pzlow=-PipeStart+5*STPSIZE
  pzhigh=PipeEnd-5*STPSIZE

  cxlow=magnetic, cxhigh=electric
  cylow=magnetic, cyhigh=electric
  czlow=electric, czhigh=electric

-brick
  material=EL
  xlow= -INF, xhigh=INF
  ylow= -INF, yhigh=INF
  zlow= -INF, zhigh=INF
  doit

```

Example of GdfidL input(1)

```

define(EL,1) define(MAG,2)
define(INF, 1000)

define(PipeRadius, 25.40e-3/2)
define(PipeStart, 100.0e-3)
define(PipeEnd,100.0e-3)
define(BpmAdd,13.5e-3)
define(STPSIZE,0.1e-3)

-general
  outfile= /users/tobiyama/GD/bpm-CL22/temp/
  scratchbase =/users/tobiyama/GD/scratch/
-material
  material=10
  type=electric
  material=15
  type=normal, epsr=9.7, muer=1, kappa=0, mkappa=0

```

```

-mesh
  spacing=0.1e-3
  pxlow= 0
  pxhigh=(PipeRadius+BpmAdd-5*STPSIZE)
  pylow= 0
  pyhigh=(PipeRadius+BpmAdd-5*STPSIZE)
  pzlow=-PipeStart+5*STPSIZE
  pzhigh=PipeEnd-5*STPSIZE

  cxlow=magnetic, cxhigh=electric
  cylow=magnetic, cyhigh=electric
  czlow=electric, czhigh=electric

-brick
  material=EL
  xlow= -INF, xhigh=INF
  ylow= -INF, yhigh=INF
  zlow= -INF, zhigh=INF
  doit

```

Example of GdfidL input(1)

```

define(EL,1) define(MAG,2)
define(INF, 1000)

define(PipeRadius, 25.40e-3/2)
define(PipeStart, 100.0e-3)
define(PipeEnd,100.0e-3)
define(BpmAdd,13.5e-3)
define(STPSIZE,0.1e-3)

-general
  outfile= /users/tobiyama/GD/bpm-CL22/temp/
  scratchbase =/users/tobiyama/GD/scratch/
-material
  material=10
  type=electric
  material=15
  type=normal, epsr=9.7, muer=1, kappa=0, mkappa=0

```

```

-mesh
  spacing=0.1e-3
  pxlow= 0
  pxhigh=(PipeRadius+BpmAdd-5*STPSIZE)
  pylow= 0
  pyhigh=(PipeRadius+BpmAdd-5*STPSIZE)
  pzlow=-PipeStart+5*STPSIZE
  pzhigh=PipeEnd-5*STPSIZE

  cxlow=magnetic, cxhigh=electric
  cylow=magnetic, cyhigh=electric
  czlow=electric, czhigh=electric

-brick
  material=EL
  xlow= -INF, xhigh=INF
  ylow= -INF, yhigh=INF
  zlow= -INF, zhigh=INF
  doit

```

Example of GdfidL input(1)

```

define(EL,1) define(MAG,2)
define(INF, 1000)

define(PipeRadius, 25.40e-3/2)
define(PipeStart, 100.0e-3)
define(PipeEnd,100.0e-3)
define(BpmAdd,13.5e-3)
define(STPSIZE,0.1e-3)

-general
  outfile= /users/tobiyama/GD/bpm-CL22/temp/
  scratchbase =/users/tobiyama/GD/scratch/
-material
  material=10
  type=electric
  material=15
  type=normal, epsr=9.7, muer=1, kappa=0, mkappa=0

```

```

-mesh
  spacing=0.1e-3
  pxlow= 0
  pxhigh=(PipeRadius+BpmAdd-5*STPSIZE)
  pylow= 0
  pyhigh=(PipeRadius+BpmAdd-5*STPSIZE)
  pzlow=-PipeStart+5*STPSIZE
  pzhigh=PipeEnd-5*STPSIZE

  cxlow=magnetic, cxhigh=electric
  cylow=magnetic, cyhigh=electric
  czlow=electric, czhigh=electric

-brick
  material=EL
  xlow= -INF, xhigh=INF
  ylow= -INF, yhigh=INF
  zlow= -INF, zhigh=INF
  doit

```

Example of GdfidL input(2)

```
#
# cave beam pipe
#
-gbor
material=0
origin=(0,0,0)
zprimedirection=(0,0,1)
rprimedirection=(1,0,0)
range=(0,360)

clear

point(-PipeStart,0)
point(-PipeStart,PipeRadius)
point(PipeEnd,PipeRadius)
point(PipeEnd,0)
point(-PipeStart,0)
show=all
doit
```

```
#
# bpm block
#
macro BpmBlock
define(Dist, @arg1)
define(Angle, (@arg2)*@pi/180)

define(BpmDiskRadius, 6e-3/2)
define(BpmRodRadius, 1.8e-3/2)
define(BpmairRadius, 1.1e-3/2)
define(BpmOutRadius1, 8e-3/2)
define(BpmOutRadius2, 4.1e-3/2)
define(BpmCera1Radius, 4.1e-3/2)
define(BpmCera2Radius, 8e-3/2)
define(BpmDiskLength, 1e-3)
define(BpmRod1Length, 2.9e-3)
define(BpmRod2Length, 3e-3)
define(BpmRod3Length, 6.1e-3)

define(Cerastart, 1.9e-3)
define(Cera1Length, 2e-3)
define(Cera2Length, 3e-3)

define(DP1, (5e-
3+BpmDiskLength+BpmRod1Length+BpmRod2Length))
define(DP2, (DP1+Dist - 5e-3))
```

Example of GdfidL input(2)

```
#
# cave beam pipe
#
-gbor
material=0
origin=(0,0,0)
zprimedirection=(0,0,1)
rprimedirection=(1,0,0)
range=(0,360)
```

```
clear
```

```
point(-PipeStart,0)
point(-PipeStart,PipeRadius)
point(PipeEnd,PipeRadius)
point(PipeEnd,0)
point(-PipeStart,0)
show=all
doit
```

```
#
# bpm block
#
macro BpmBlock
define(Dist, @arg1)
define(Angle, (@arg2)*@pi/180)

define(BpmDiskRadius, 6e-3/2)
define(BpmRodRadius, 1.8e-3/2)
define(BpmairRadius, 1.1e-3/2)
define(BpmOutRadius1, 8e-3/2)
define(BpmOutRadius2, 4.1e-3/2)
define(BpmCera1Radius, 4.1e-3/2)
define(BpmCera2Radius, 8e-3/2)
define(BpmDiskLength, 1e-3)
define(BpmRod1Length, 2.9e-3)
define(BpmRod2Length, 3e-3)
define(BpmRod3Length, 6.1e-3)

define(Cerastart, 1.9e-3)
define(Cera1Length, 2e-3)
define(Cera2Length, 3e-3)

define(DP1, (5e-
3+BpmDiskLength+BpmRod1Length+BpmRod2Length))
define(DP2, (DP1+Dist - 5e-3))
```

Example of GdfidL input(2)

```
#
# cave beam pipe
#
-gbor
material=0
origin=(0,0,0)
zprimedirection=(0,0,1)
rprimedirection=(1,0,0)
range=(0,360)
```

```
clear
```

```
point(-PipeStart,0)
point(-PipeStart,PipeRadius)
point(PipeEnd,PipeRadius)
point(PipeEnd,0)
point(-PipeStart,0)
show=all
doit
```

```
#
# bpm block
#
macro BpmBlock
define(Dist, @arg1)
define(Angle, (@arg2)*@pi/180)

define(BpmDiskRadius, 6e-3/2)
define(BpmRodRadius, 1.8e-3/2)
define(BpmairRadius, 1.1e-3/2)
define(BpmOutRadius1, 8e-3/2)
define(BpmOutRadius2, 4.1e-3/2)
define(BpmCera1Radius, 4.1e-3/2)
define(BpmCera2Radius, 8e-3/2)
define(BpmDiskLength, 1e-3)
define(BpmRod1Length, 2.9e-3)
define(BpmRod2Length, 3e-3)
define(BpmRod3Length, 6.1e-3)

define(Cerastart, 1.9e-3)
define(Cera1Length, 2e-3)
define(Cera2Length, 3e-3)

define(DP1, (5e-
3+BpmDiskLength+BpmRod1Length+BpmRod2Length))
define(DP2, (DP1+Dist - 5e-3))
```

Example of GdfidL input(3)

```
#
-gccylinder
material=0, radius=BpmOutRadius1,length = DP1
origin=(cos(Angle)*(Dist - 5e-3), sin(Angle)*(Dist - 5e-3), 0)
direction=(cos(Angle),sin(Angle),0)
doit
-gccylinder
material=0, radius=BpmOutRadius2
length= BpmRod3Length
origin=(cos(Angle)*(DP2),sin(Angle)*(DP2), 0)
direction=(cos(Angle),sin(Angle),0)
doit
#
# ceramic
#
-gccylinder
material=15, radius=BpmCera1Radius,length=Cera1Length
origin=(cos(Angle)*(Dist+Cerastart), sin(Angle)*(Dist+Cerastart), 0)
direction=(cos(Angle),sin(Angle),0)
doit
-gccylinder
material=15, radius=BpmCera2Radius,length=Cera2Length
origin=(cos(Angle)*(Dist+Cerastart+Cera1Length),¥
      sin(Angle)*(Dist+Cerastart+Cera1Length), 0)
direction=(cos(Angle),sin(Angle),0)
doit
```

```
# air again
-gccylinder
material=0, radius=BpmRodRadius,length =
(Cera1Length+Cera2Length)
origin=(cos(Angle)*(Dist+Cerastart),sin(Angle)*(Dist-Cerastart) 0)
direction=(cos(Angle),sin(Angle),0)
doit
# bpm disk
-gccylinder
material=10, radius=BpmDiskRadius,length=BpmDiskLength
origin=(cos(Angle)*(Dist), sin(Angle)*(Dist), 0)
direction=(cos(Angle),sin(Angle),0)
doit
-gccylinder
material=10, radius=BpmRodRadius, length=BpmRod1Length
origin=(cos(Angle)*(Dist+BpmDiskLength),¥
      sin(Angle)*(Dist+BpmDiskLength), 0)
direction=(cos(Angle),sin(Angle),0)
doit

....still many definitions...

endmacro #BpmBlock
```

Example of GdfidL input(4)

```

call BpmBlock(13.2e-3,90)
call BpmBlock(13.2e-3,0)
call BpmBlock(13.2e-3,-90)
call BpmBlock(13.2e-3,180)

-fdtd
-lcharge
  charge=1e-12
  sigma=6e-3
  xposition=0, yposition=0
  shigh=5
  showdata=no
-ports
  name=beamlow, plane=zlow, modes=0, npml=40, doit
  name=beamhigh, plane=zhigh, modes=0, npml=40, doit
  name=bpmyp, plane=yhigh, modes=1, npml=15, doit
  name=bpmxp, plane=xhigh, modes=1, npml=15, doit

-fdtd
doit

```

- **mesh 0.1mm**
- **bunch length 6mm**
- **beam=(0,0) [center]**
- **wake up to 5m**

Example of GdfidL input(4)

```
call BpmBlock(13.2e-3,90)
call BpmBlock(13.2e-3,0)
call BpmBlock(13.2e-3,-90)
call BpmBlock(13.2e-3,180)
```

```
-fddd
-lcharge
  charge=1e-12
  sigma=6e-3
  xposition=0, yposition=0
  shigh=5
  showdata=no
-ports
  name=beamlow, plane=zlow, modes=0, npml=40, doit
  name=beamhigh, plane=zhigh, modes=0, npml=40, doit
  name=bpmyp, plane=yhigh, modes=1, npml=15, doit
  name=bpmxp, plane=xhigh, modes=1, npml=15, doit

-fddd
doit
```

- **mesh 0.1mm**
- **bunch length 6mm**
- **beam=(0,0) [center]**
- **wake up to 5m**

Example of GdfidL input(4)

```
call BpmBlock(13.2e-3,90)
call BpmBlock(13.2e-3,0)
call BpmBlock(13.2e-3,-90)
call BpmBlock(13.2e-3,180)
```

```
-fddd
```

```
-lcharge
```

```
charge=1e-12
```

```
sigma=6e-3
```

```
xposition=0, yposition=0
```

```
shigh=5
```

```
showdata=no
```

```
-ports
```

```
name=beamlow, plane=zlow, modes=0, npml=40, doit
```

```
name=beamhigh, plane=zhigh, modes=0, npml=40, doit
```

```
name=bpmyp, plane=yhigh, modes=1, npml=15, doit
```

```
name=bpmxp, plane=xhigh, modes=1, npml=15, doit
```

```
-fddd
```

```
doit
```

- **mesh 0.1mm**
- **bunch length 6mm**
- **beam=(0,0) [center]**
- **wake up to 5m**

Example of GdfidL input(4)

```
call BpmBlock(13.2e-3,90)
call BpmBlock(13.2e-3,0)
call BpmBlock(13.2e-3,-90)
call BpmBlock(13.2e-3,180)
```

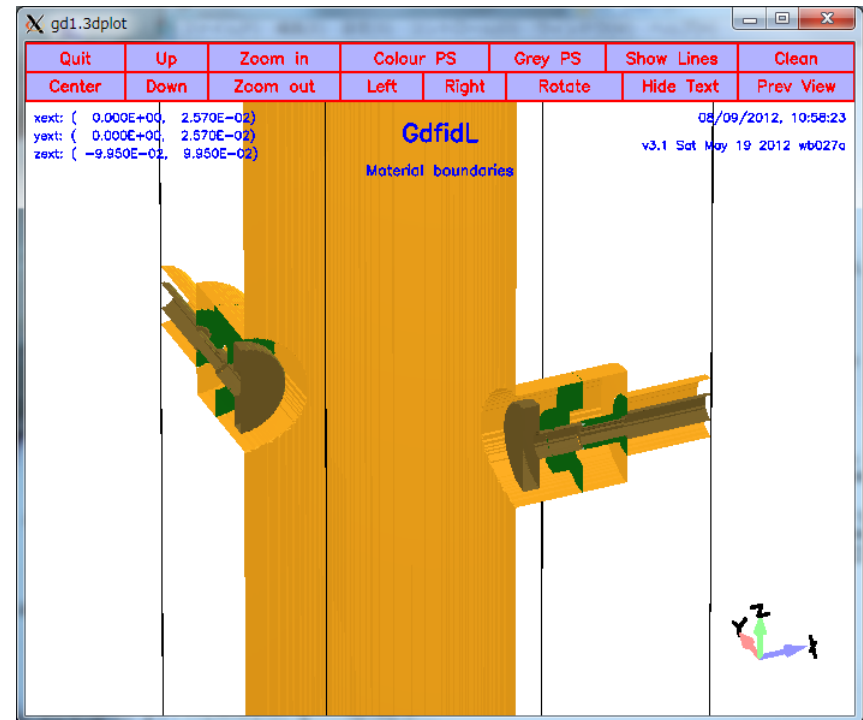
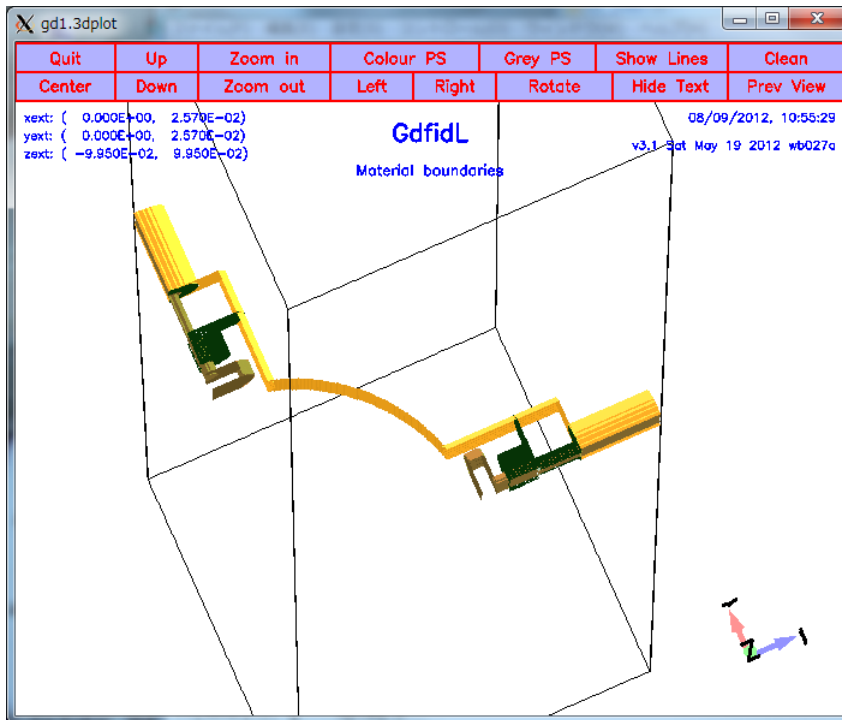
```
-fddd
-lcharge
  charge=1e-12
  sigma=6e-3
  xposition=0, yposition=0
  shigh=5
  showdata=no
```

```
-ports
  name=beamlow, plane=zlow, modes=0, npml=40, doit
  name=beamhigh, plane=zhigh, modes=0, npml=40, doit
  name=bpmyp, plane=yhigh, modes=1, npml=15, doit
  name=bpmxp, plane=xhigh, modes=1, npml=15, doit
```

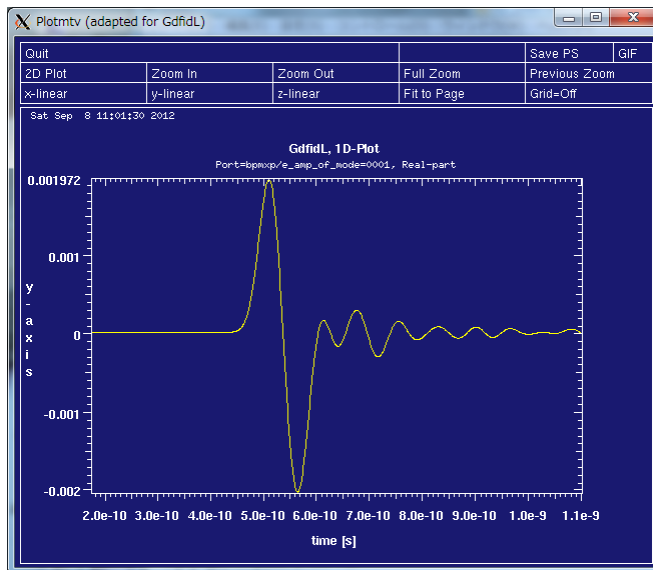
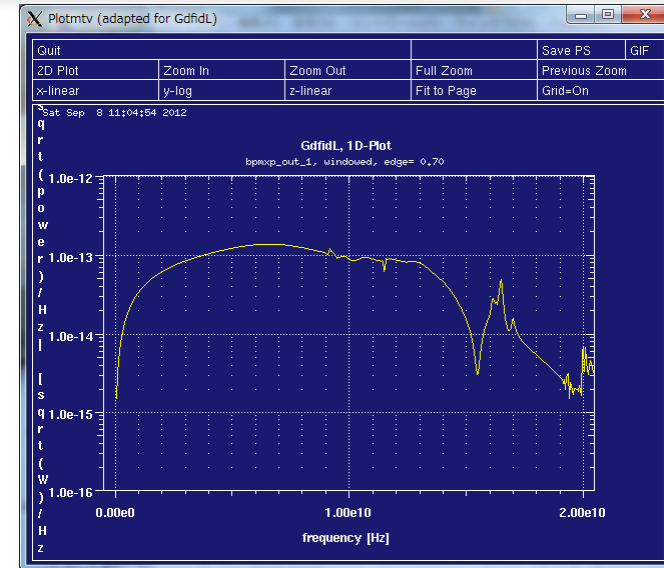
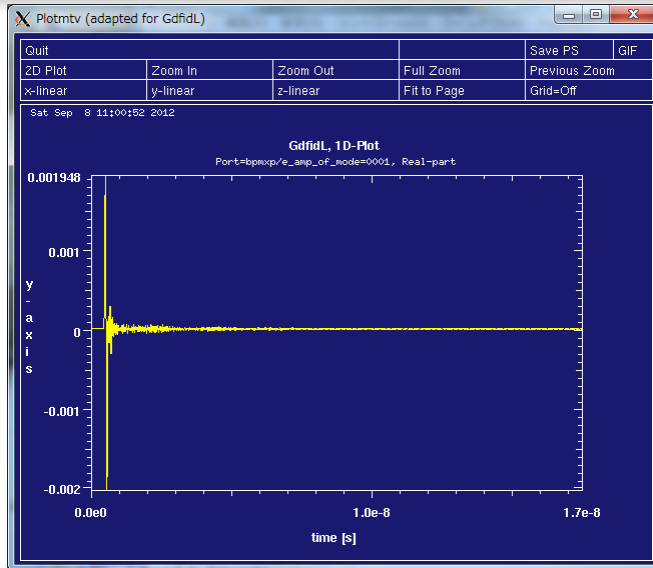
```
-fddd
doit
```

- **mesh 0.1mm**
- **bunch length 6mm**
- **beam=(0,0) [center]**
- **wake up to 5m**

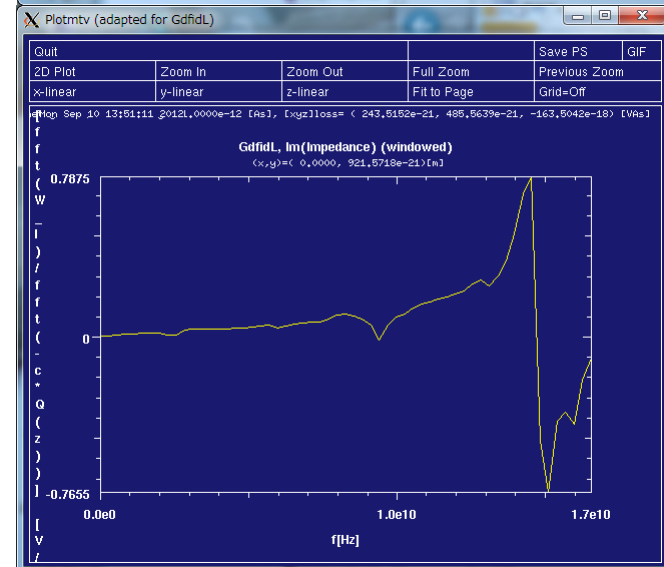
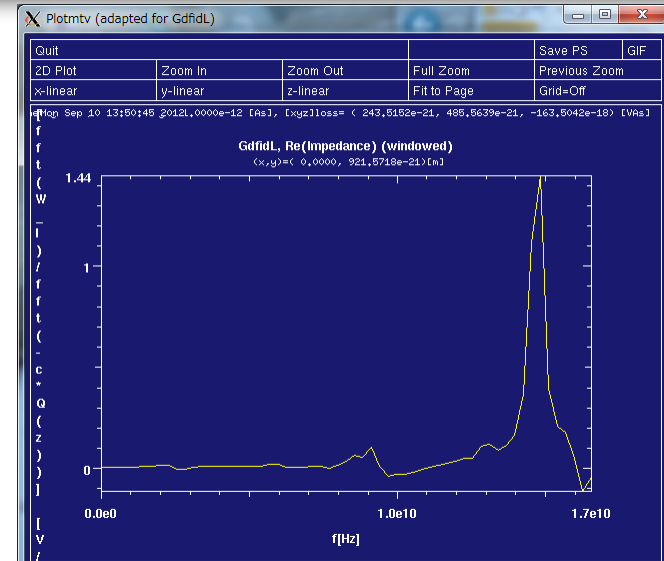
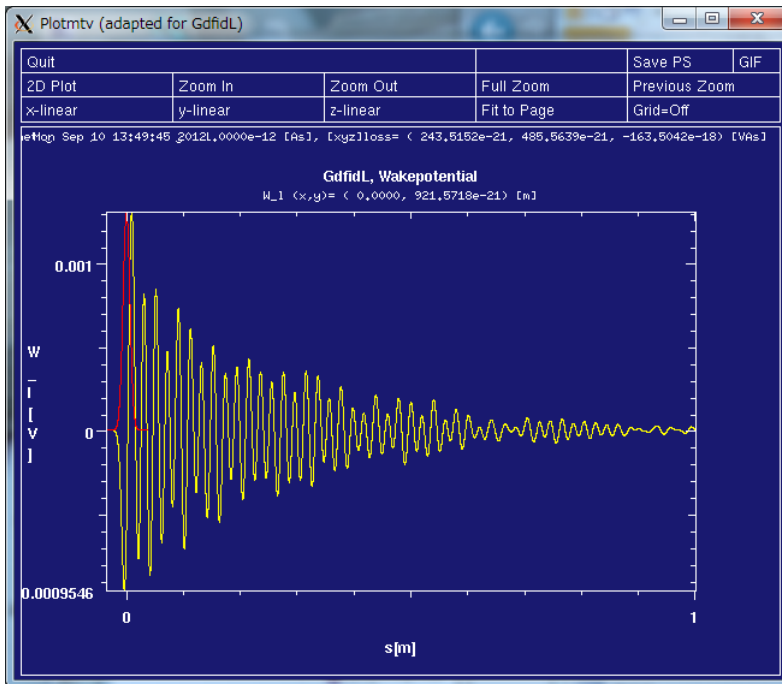
Model view



Post processing (using gd1.pp)

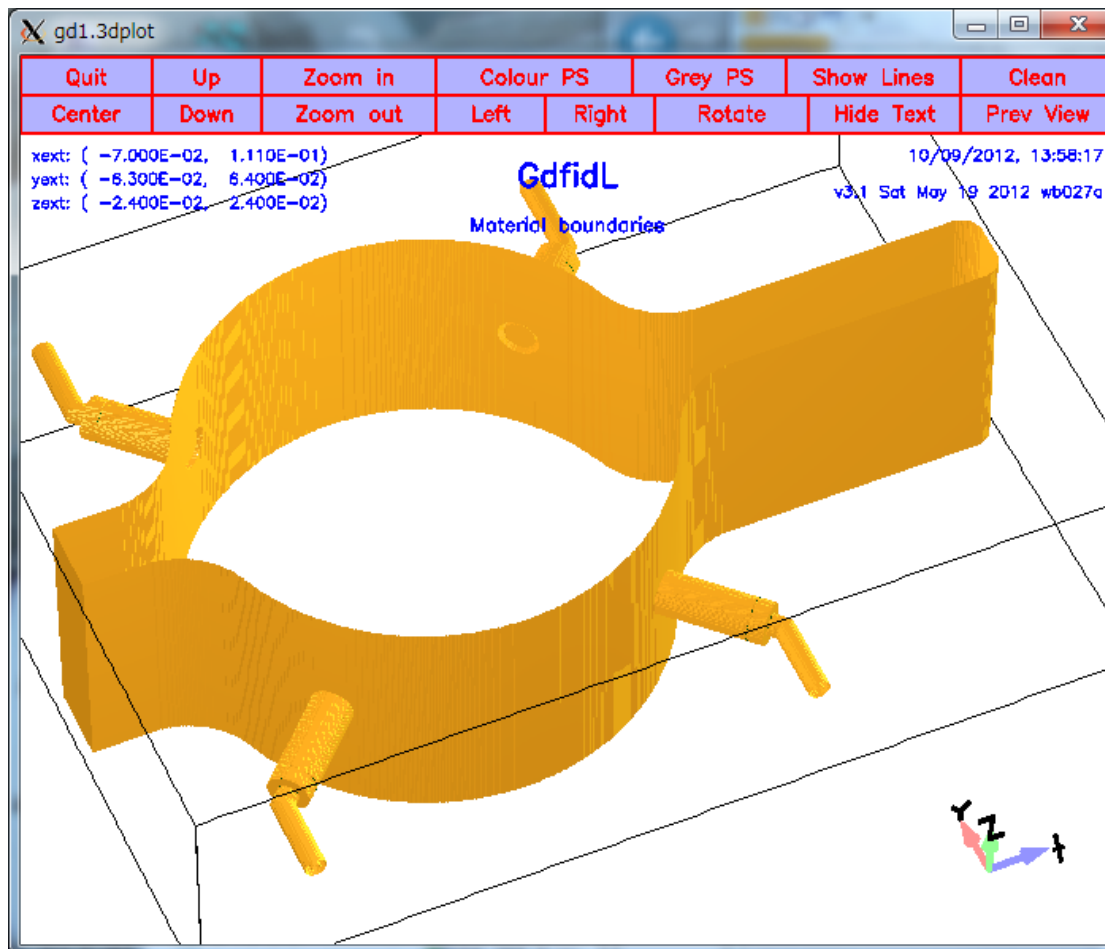


wakes, impedances

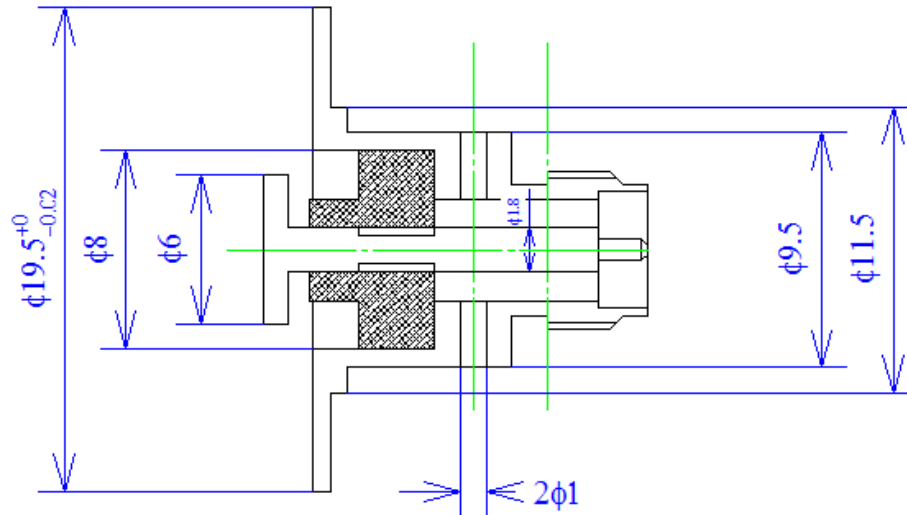


Tips

- Ports must be placed orthogonally to the boundaries.



Button electrode (KEKB-FB)



Feedback用SMAフィードスルーC型概略図

作図: 飛山真理 19/Mar/2004

縮尺4:1



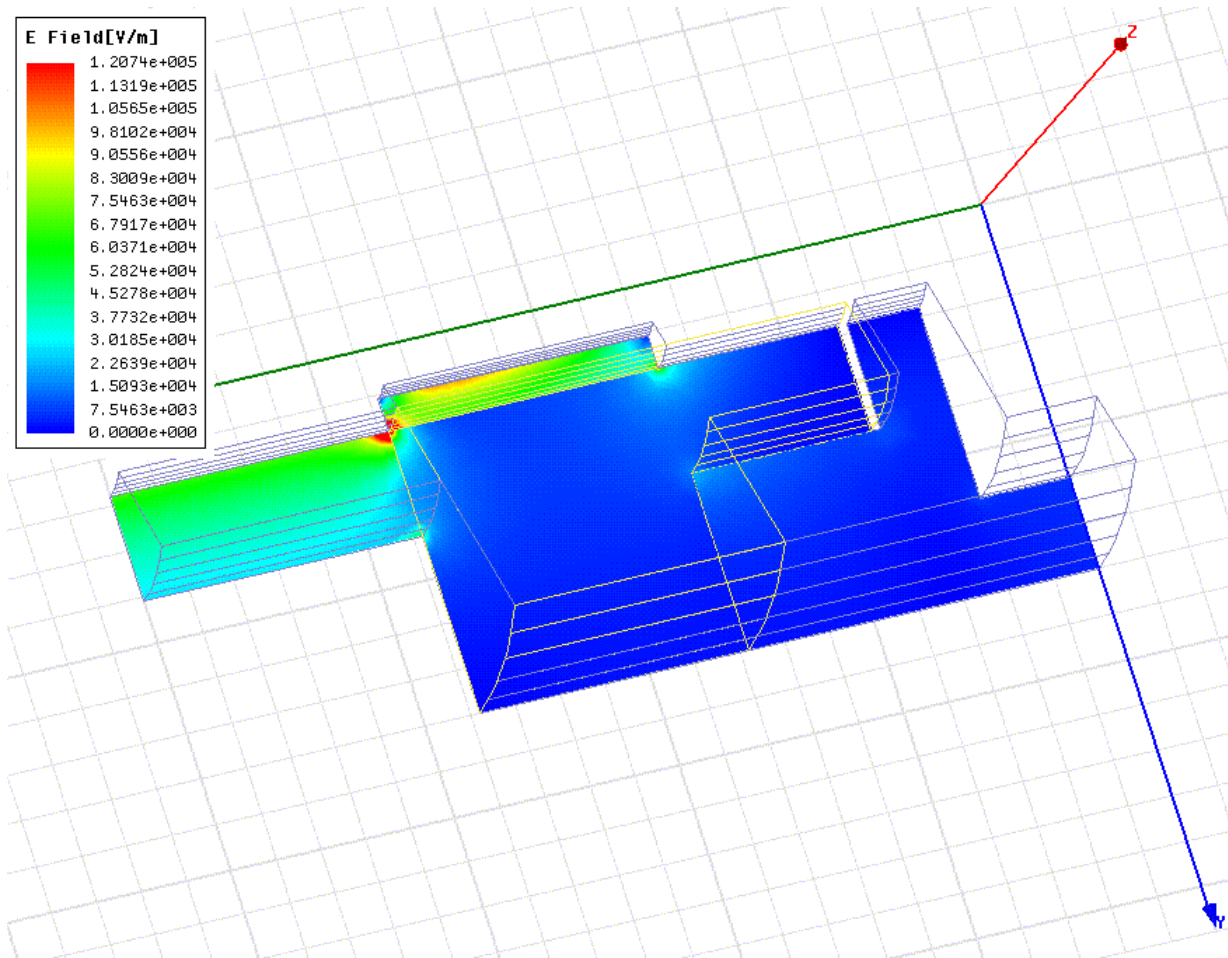
Tips

- **Using normal SMA-J connector (or N-J, too) usually encounter difficulty in RF contact at the split-pin of the center conductor due to brazing process – No spring action might be expected after the process.**
 - Using P-structure.. not easy to handle..
 - Use pin on the J-structure (“Reverse type J”)
- **Gold-plating of at the RF connector is desirable**
 - Not easy to maintain the thickness of the gold during brazing process (Gold will be diffused into the bulk metal..)

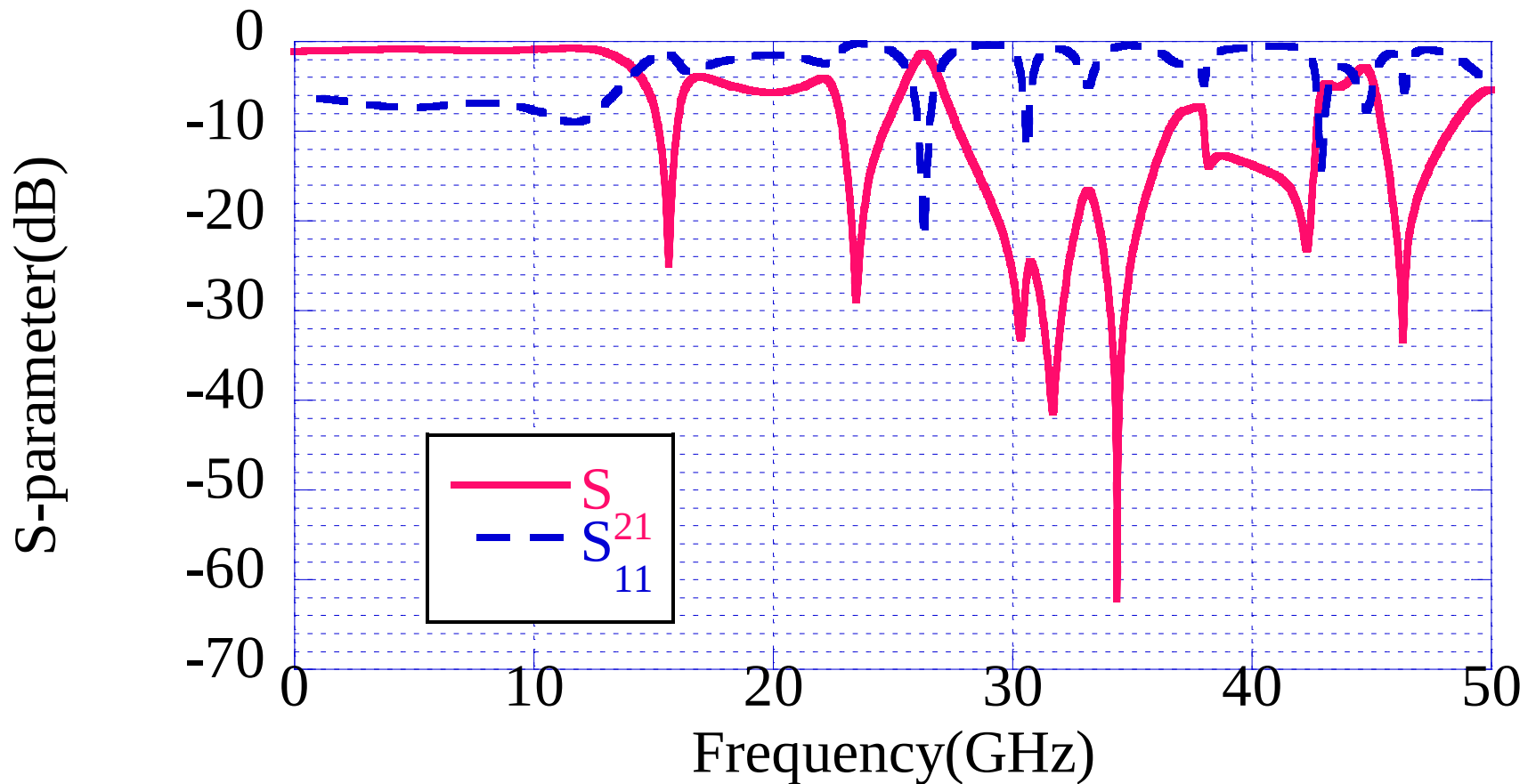
SMA-J and SMA-J(R)



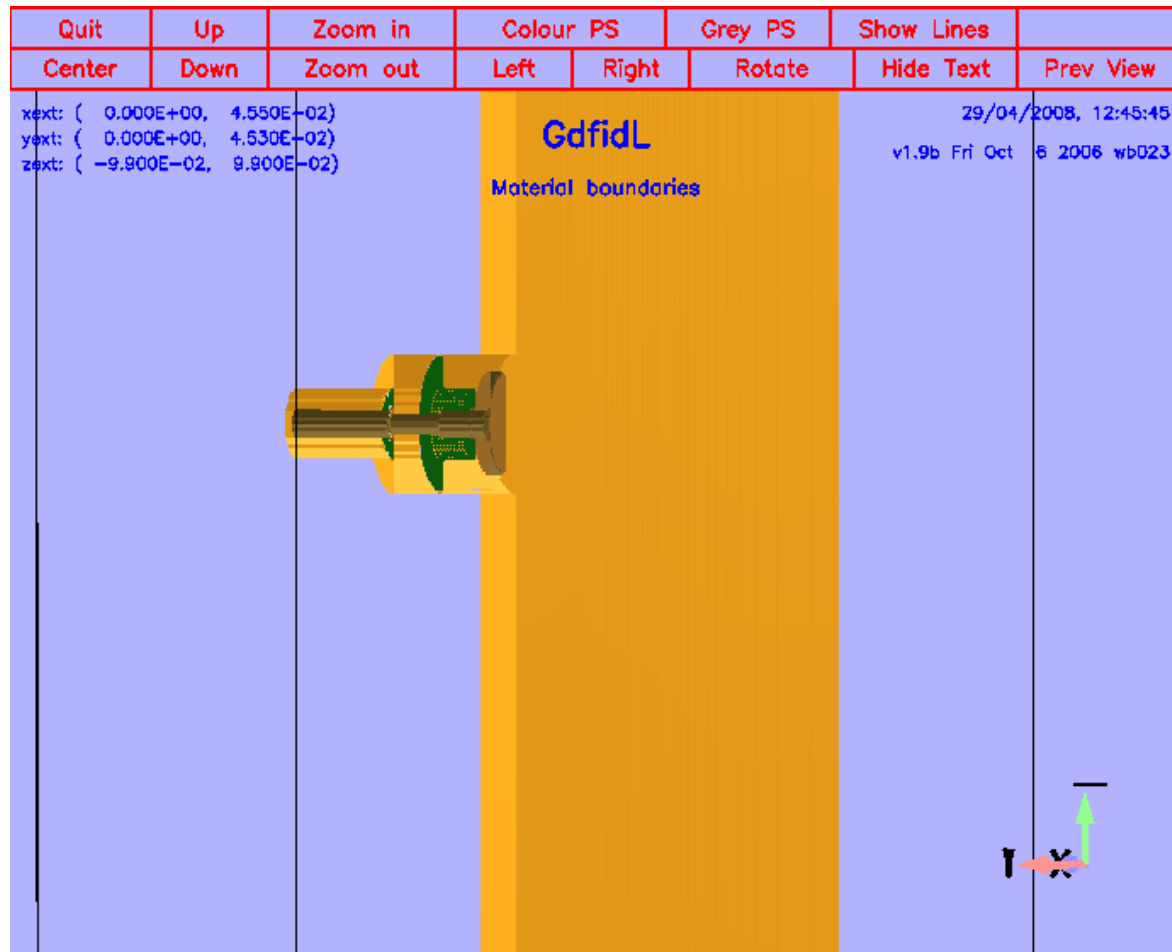
HFSS model



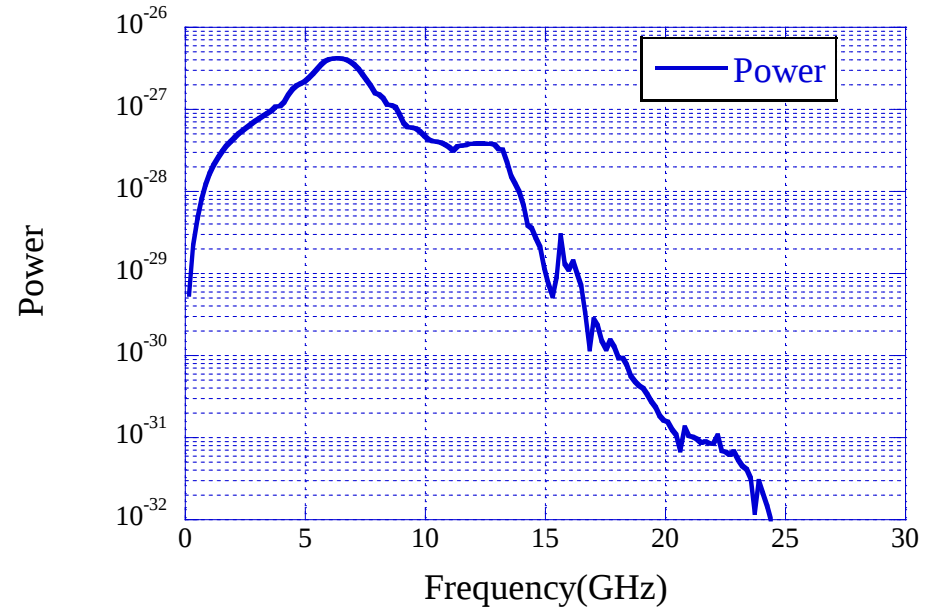
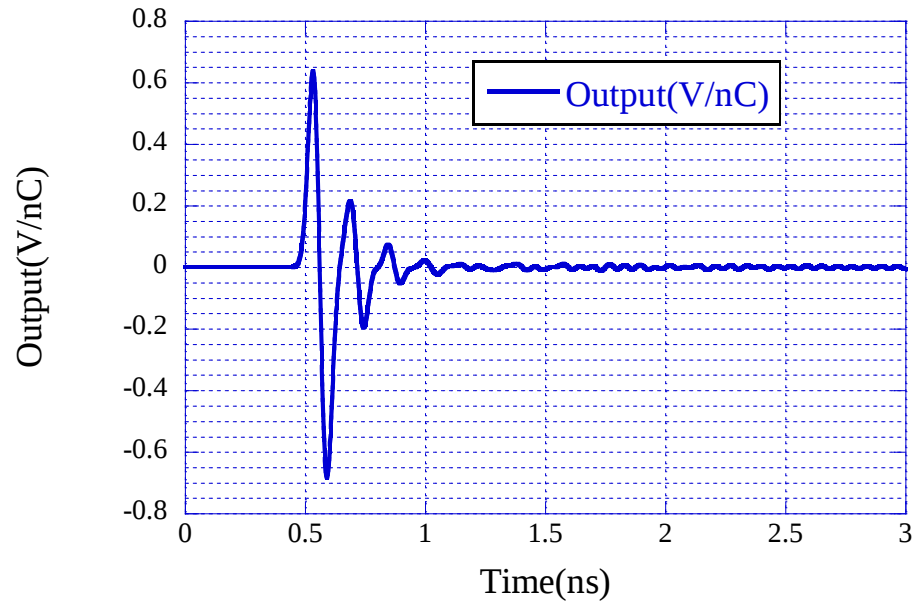
Simulated S-Parameters



GdfidL model



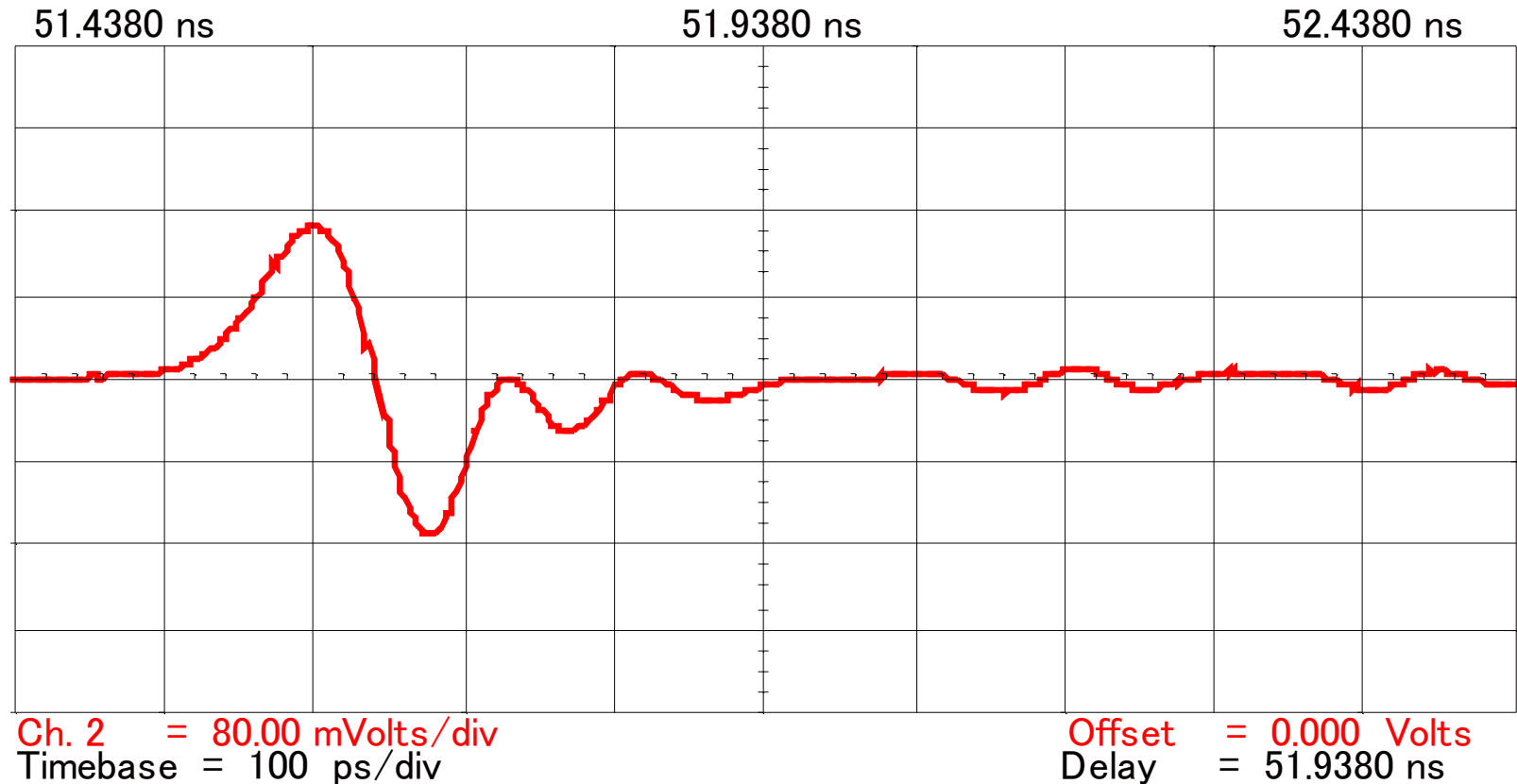
Simulated Port output



Tips

- **Need to pay attention to the possible trapped modes in the sealing structure, which may damage the seal on high beam current operation.**
 - Use Eigen-mode solver to estimate the frequency of the trapped mode.
 - Fine frequency sweep to find dip structure in S_{11}
 - Coupling to coaxial mode might be difficult.
- **Need to pay attention to the multipactoring around the button gap, especially for high single-bunch current machine.**
- **Longitudinal beam coupling impedance might not be negligible on large circumference, high beam current machine.**

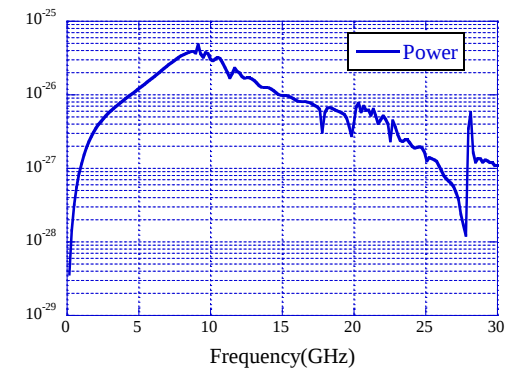
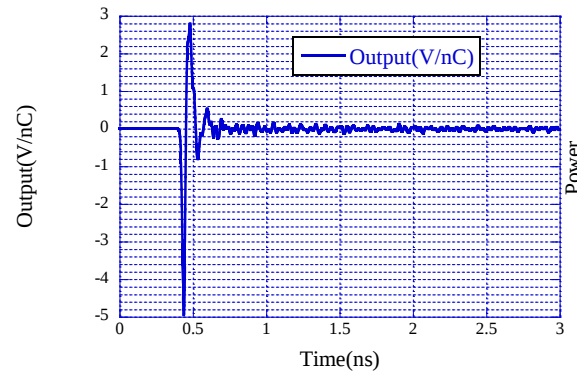
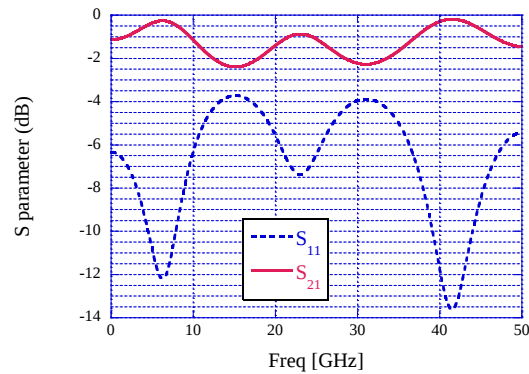
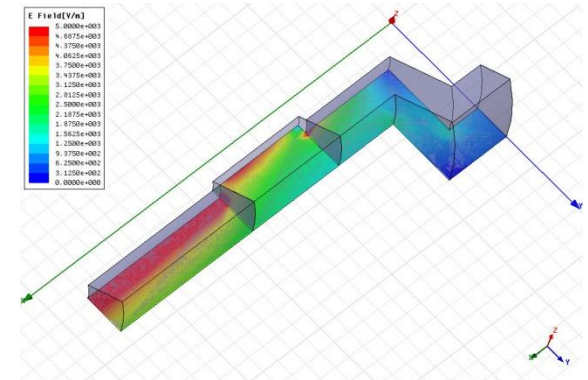
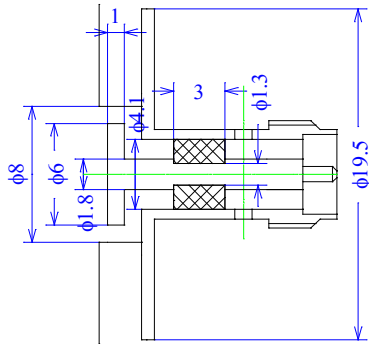
Measured signal (20GHz scope)



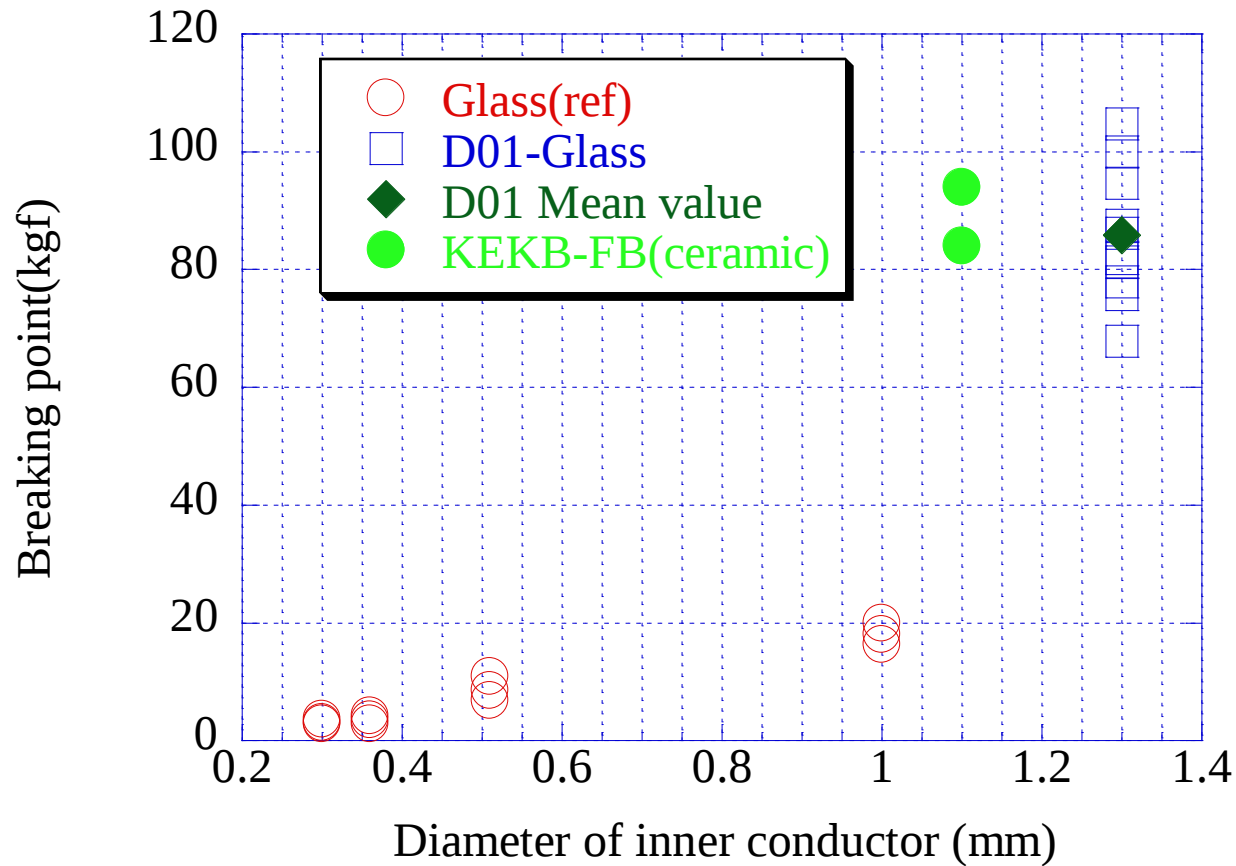
9nC/bunch LER 43dB attenuation, LER upstream

Maximum beam current at KEKB: $2A + 1.3A = 3.3A$ with large horizontal offset

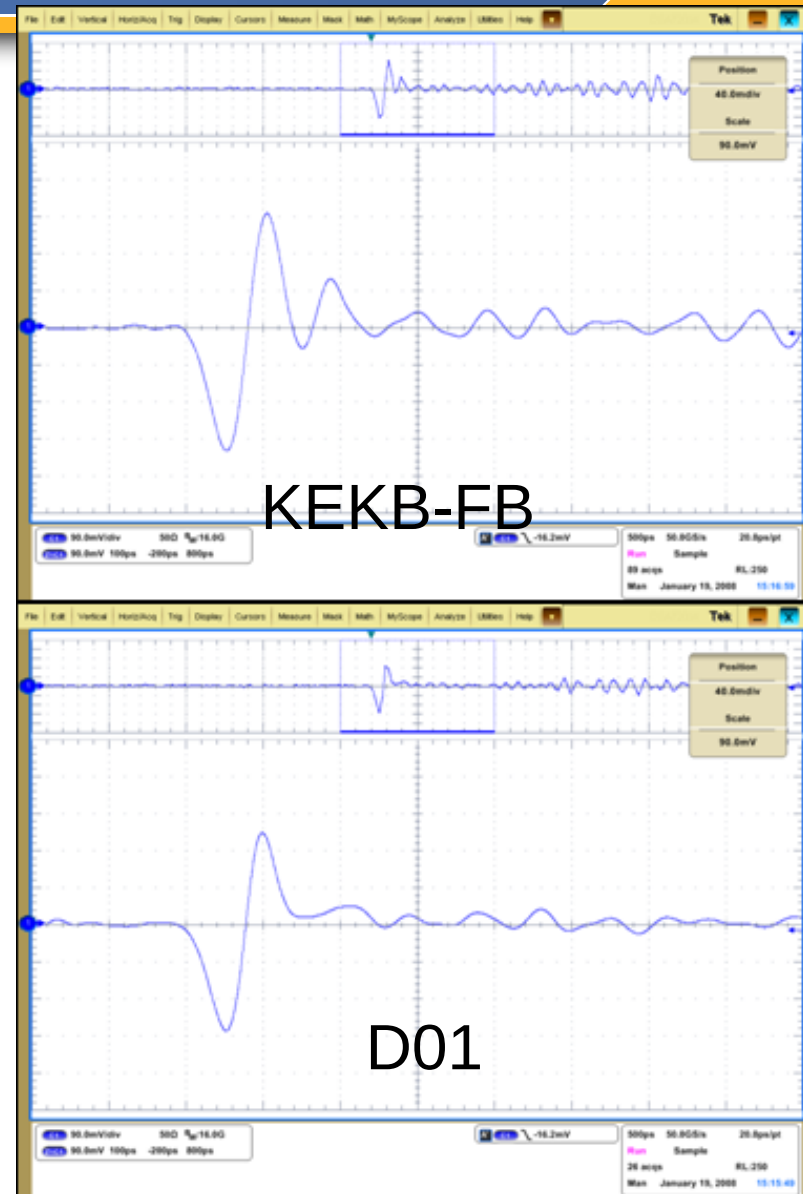
Feedthrough with low $\epsilon_r(\sim 4)$ sealing



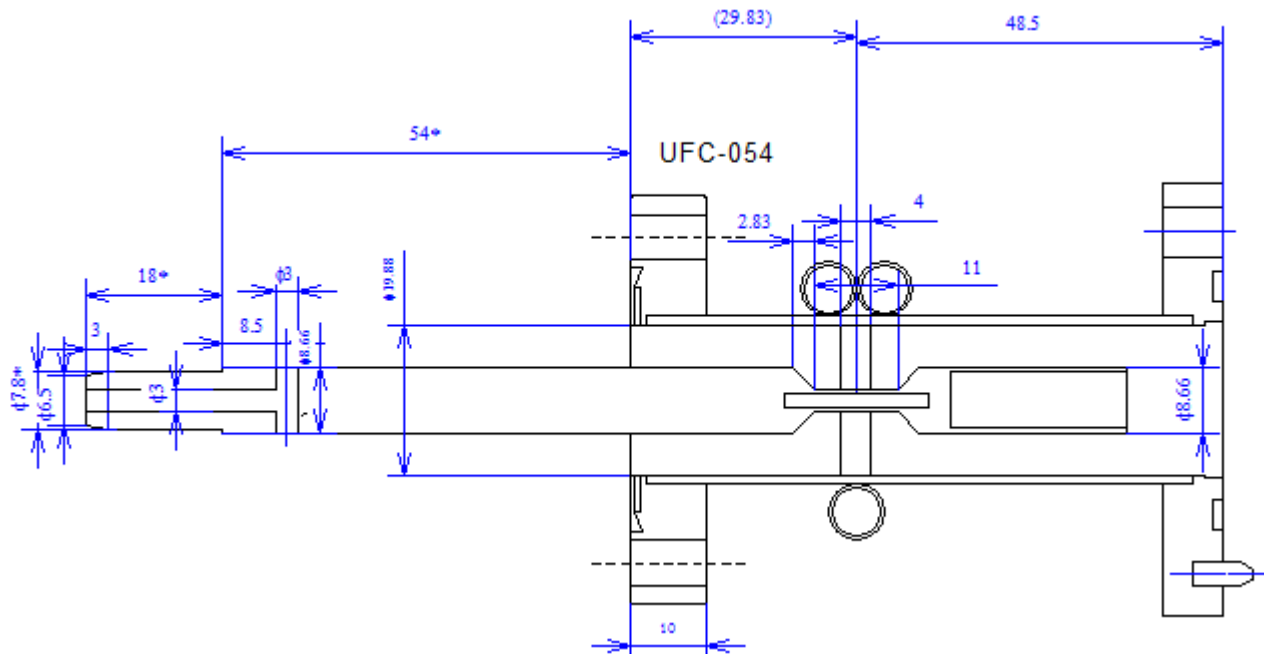
Center pin pulling-out test



Beam test using Linac short pulse



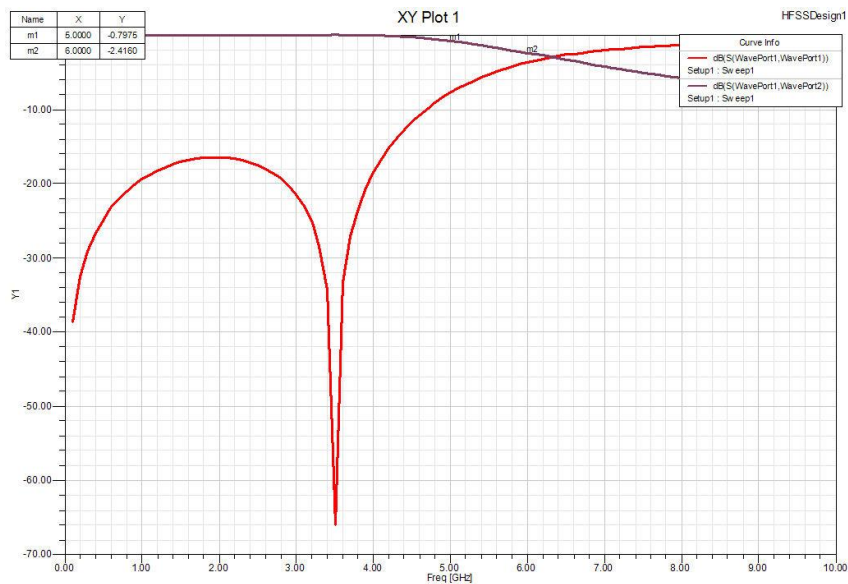
High power feedthrough



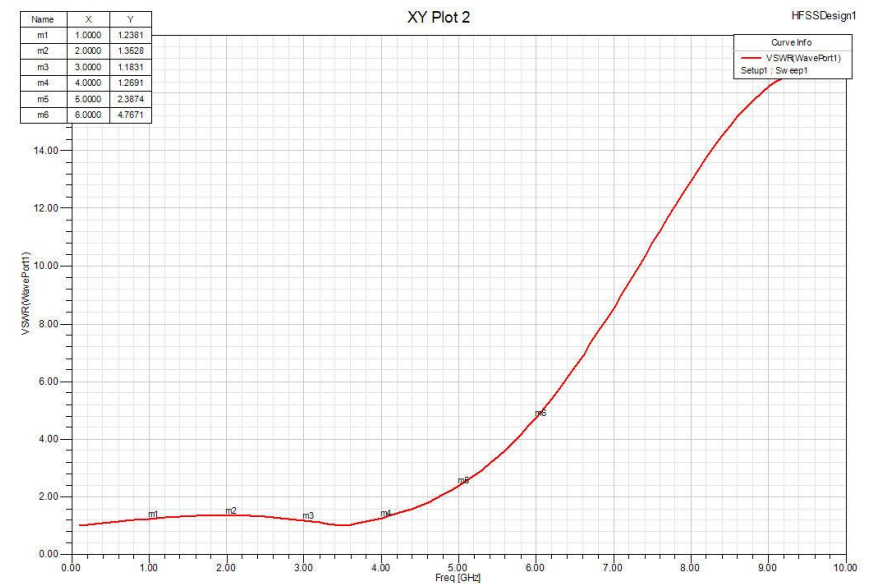
- Max CW power $\sim 5\text{kW}$
- Not so wideband (DC–3GHz)
- SWR through pass band < 1.1

Simulated S-Parameters

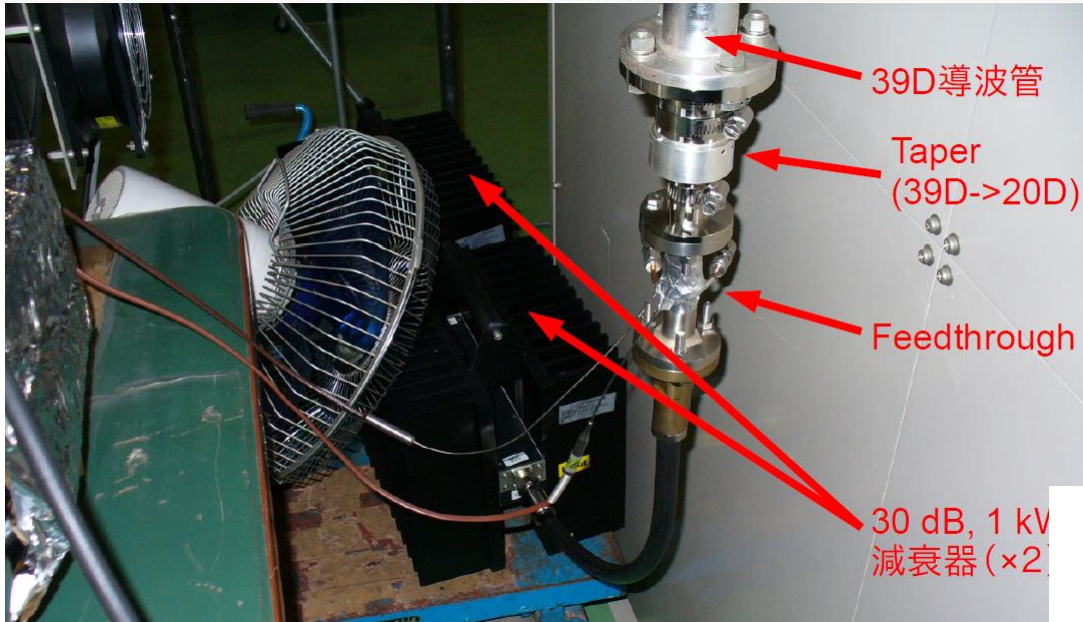
S21 and S11



VSWR(S11)



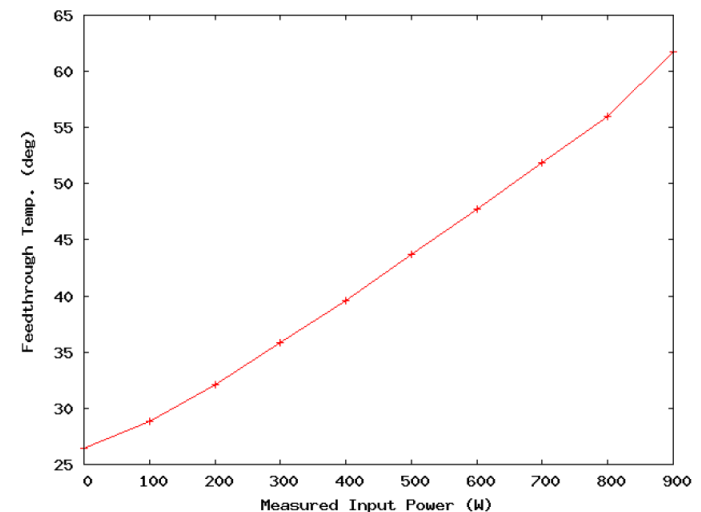
High power test



1.3GHz power
source

- **Temperature rise :4 deg/100W**
(without water cooling)

Feedthrough温度対入力パワー



冷却水がなくても900 Wまでは異常無し。

Tips

- To enhance thermal conductivity between the center conductor and the outer wall, it might be considered to fill BN ($\epsilon_R \sim 4$) at the air side.
- Si_3N_4 has lower ϵ_R and has better (lower) loss tangent
 - Expensive
 - Mechanically weaker than alumina-ceramic

Summary

- **Briefly reviewed the transmission-line theory**
 - S-Parameters, SWR, Time-domain response
- **Introduced E-M simulation software**
 - HFSS (Frequency domain)
 - GdfidL (Time domain)
- **Shown several examples**
 - KEKB-FB button electrode
 - SuperKEKB-FB button electrode
 - High power feedthrough for SuperKEKB longitudinal kicker.

Reference

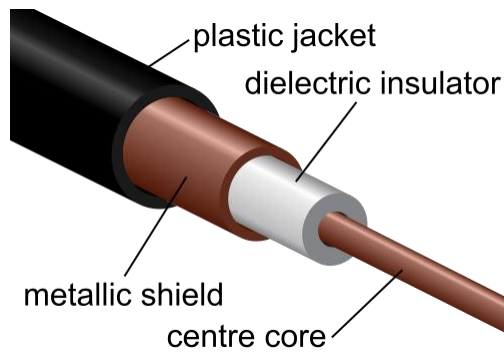
- **“Microwave Engineering Passive Circuits”,**
by Peter A. Rizzi, ISBN 0135867029
- **“Introduction to Microwave Theory”,**
by Harry A. Atwater, ISBN 0898741920
- **Ansys HFSS,**
 - <http://www.ansys.com/Products/Simulation+Technology/Electromagnetics/High-Performance+Electronic+Design/ANSYS+HFSS>
- **GdfidL, <http://www.gdfidl.de/>**
- **Kyocera Corporation**
 - <http://global.kyocera.com/prdct/fc/product/usage/vacuum/index.html>
- **Orient Microwave Corp.**
 - <http://www.orient-microwave.com/english/index.html>
- **SUHNER**
 - <http://ipaper.ipapercms.dk/HUBERSUHNER/Technologies/Radiofrequency/RFCConnectorsEN/>



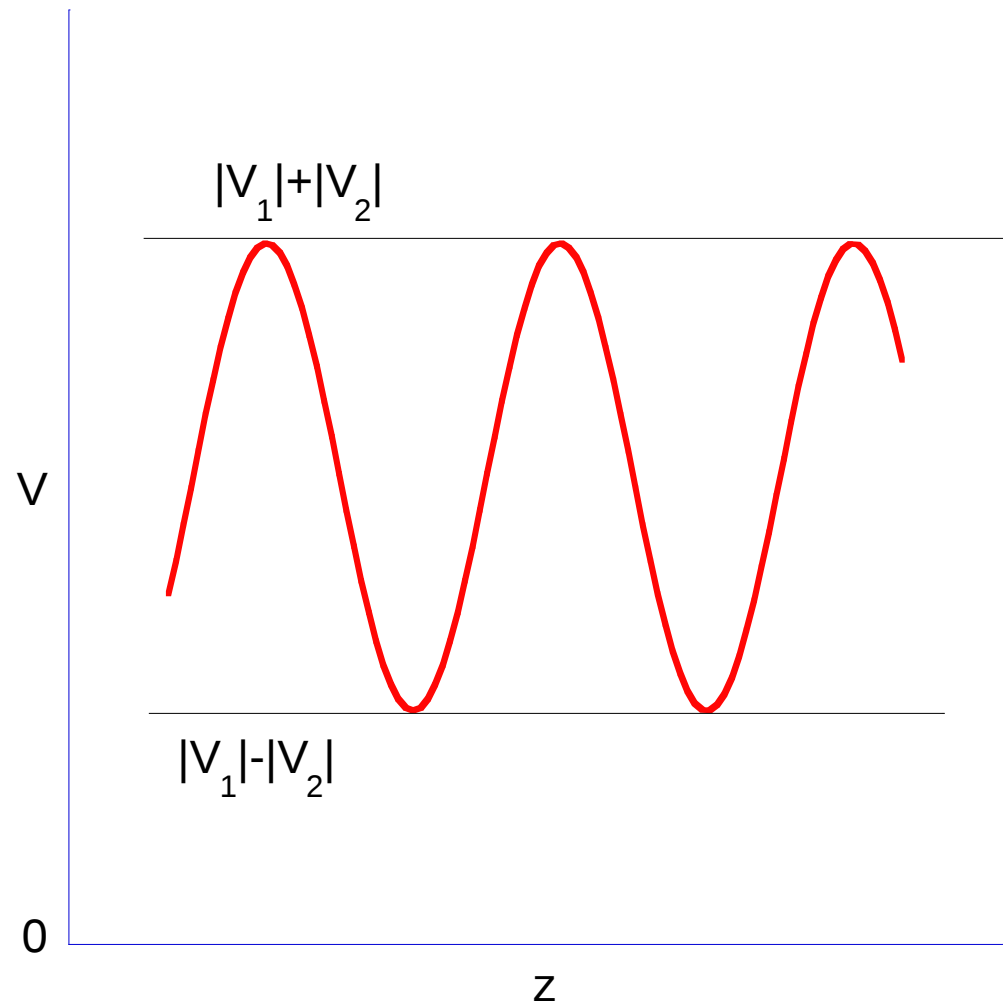
backup

Transmission line

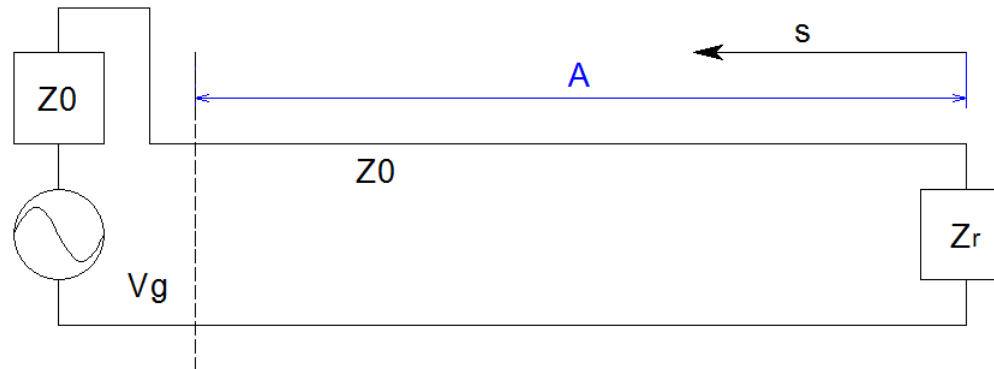
- **Coaxial cable**



Standing wave

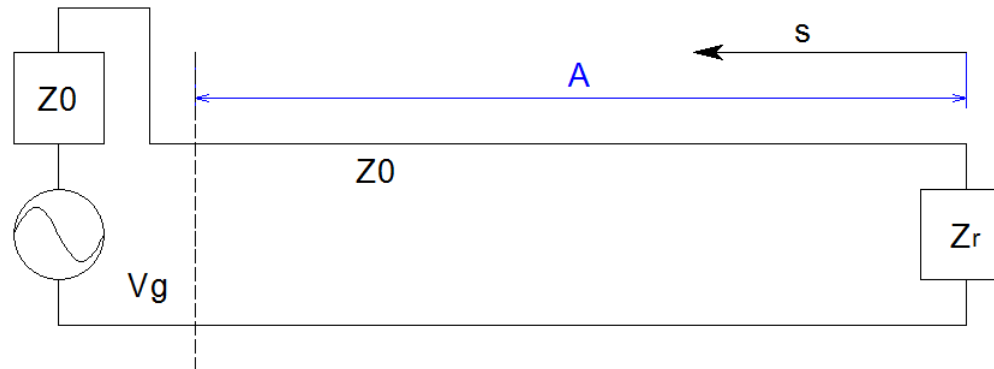


Standing wave



$$\begin{aligned}
 V &= V_1 e^{-j\beta(A-s)} + V_2 e^{j\beta(A-s)} = V_1' e^{j\beta s} + V_2' e^{-j\beta s} \\
 &= V_1' e^{j\beta s} (1 + \rho e^{-2j\beta s})
 \end{aligned}$$

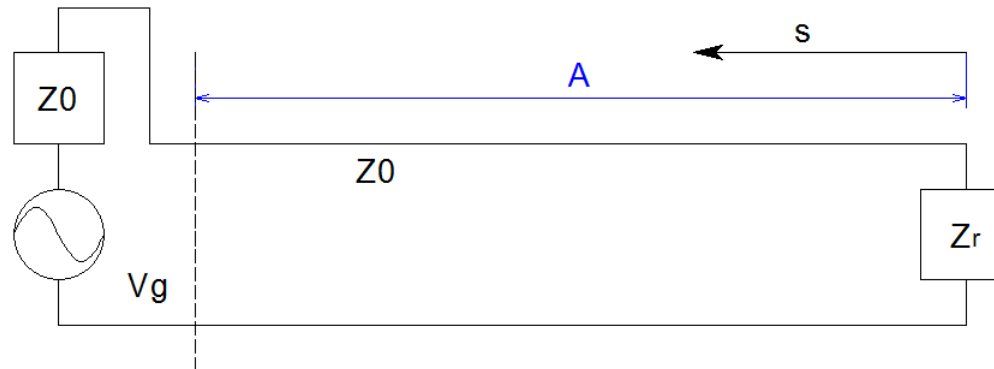
Standing wave



$$\begin{aligned}
 V &= V_1 e^{-j\beta(A-s)} + V_2 e^{j\beta(A-s)} = V_1' e^{j\beta s} + V_2' e^{-j\beta s} \\
 &= V_1' e^{j\beta s} (1 + \rho e^{-2j\beta s})
 \end{aligned}$$

- ρ : reflection constant

Standing wave



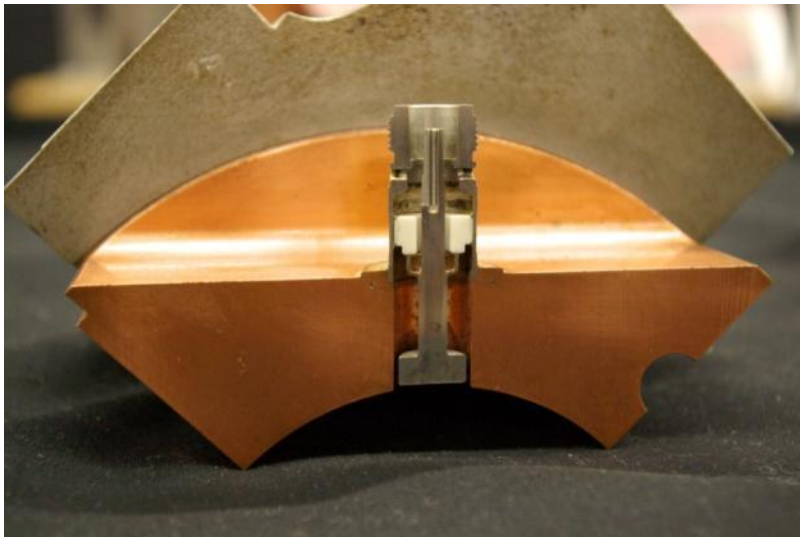
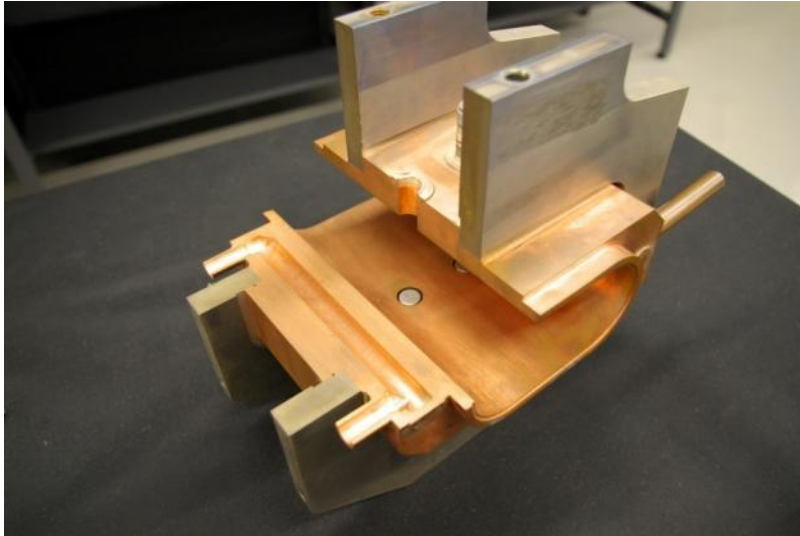
$$\begin{aligned}
 V &= V_1 e^{-j\beta(A-s)} + V_2 e^{j\beta(A-s)} = V_1' e^{j\beta s} + V_2' e^{-j\beta s} \\
 &= V_1' e^{j\beta s} (1 + \rho e^{-2j\beta s})
 \end{aligned}$$

- ρ : reflection constant

$$\rho = \frac{V_2'}{V_1'} = |\rho| e^{j\varphi}$$

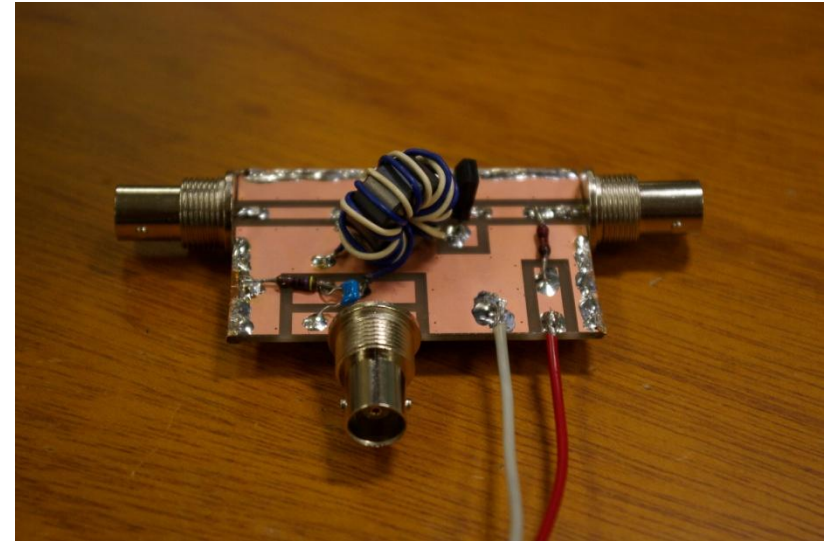
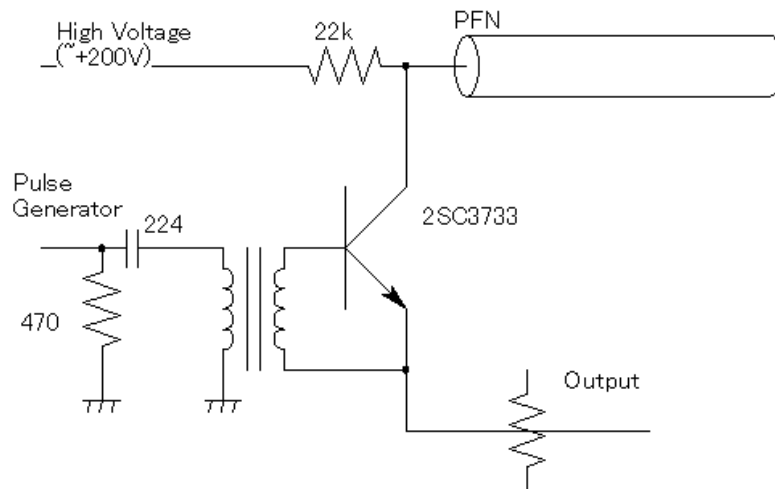
$$V = V_1' e^{j\beta s} (1 + |\rho| e^{j(\varphi - 2\beta s)})$$

例：ボタン電極型ビーム位置モニタ

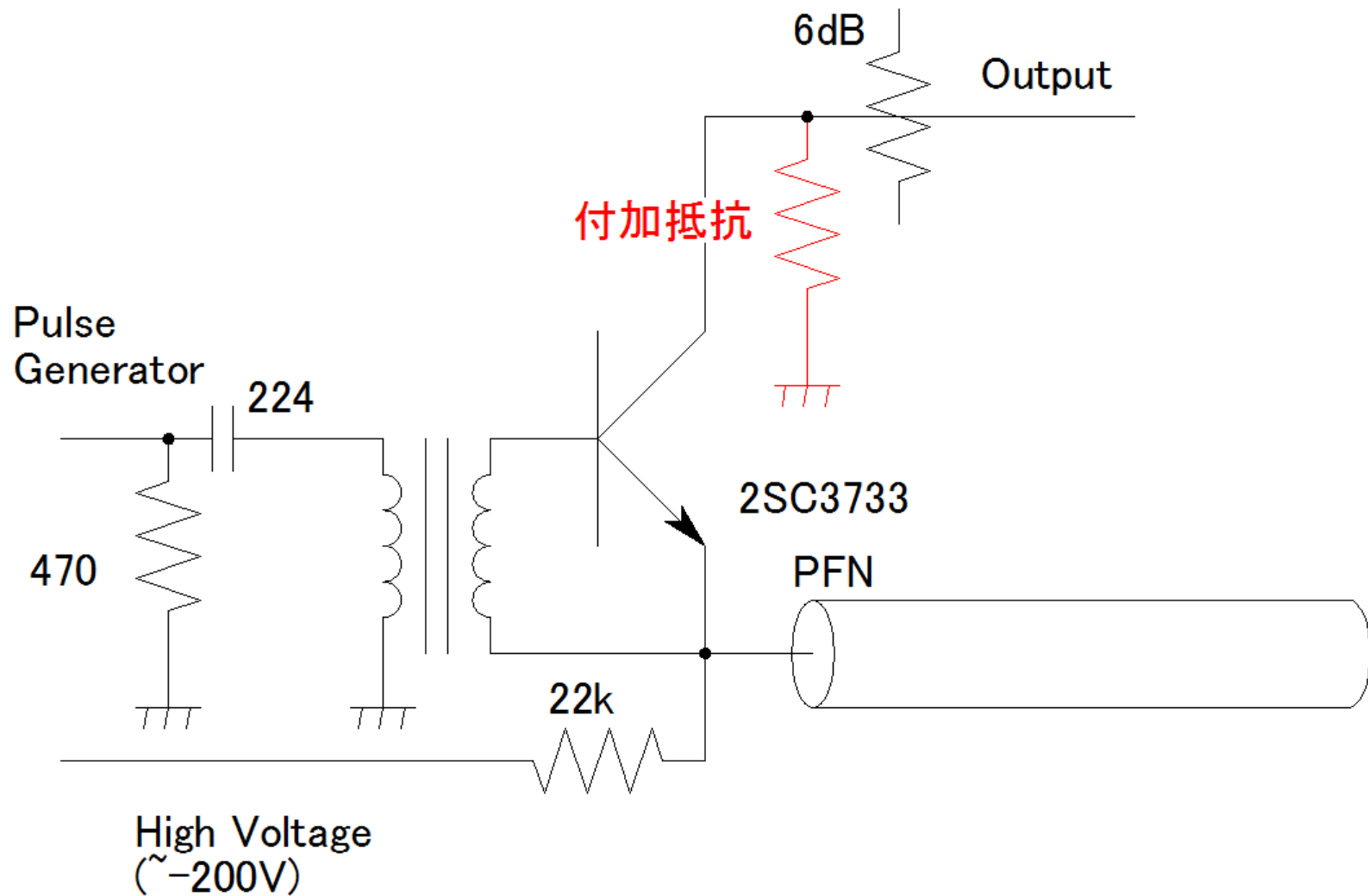


四つの電極に出てきた電荷の差から、バンチの重心位置を求めるためのモニターです。レンズの働きをする四極電磁石に固定してあります。

アバランシェランジスタパルサー(正極性)

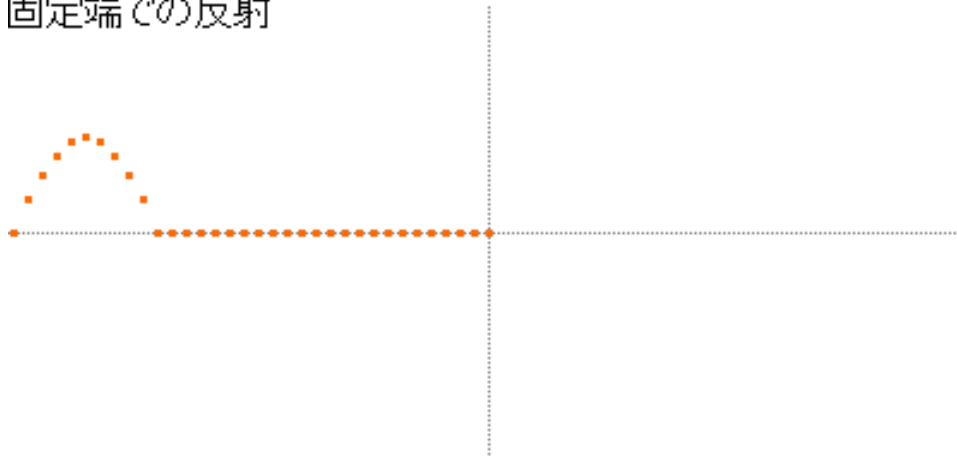


アバランシェパルサー(負極性)

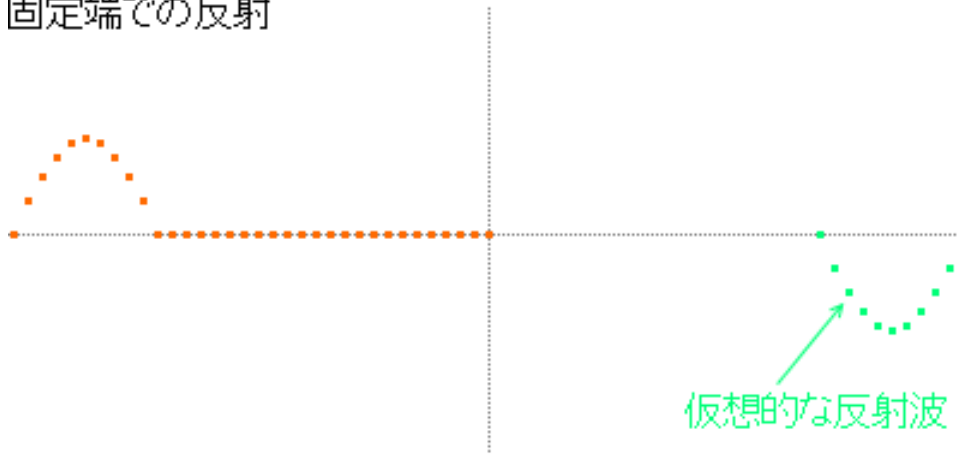


固定端での反射

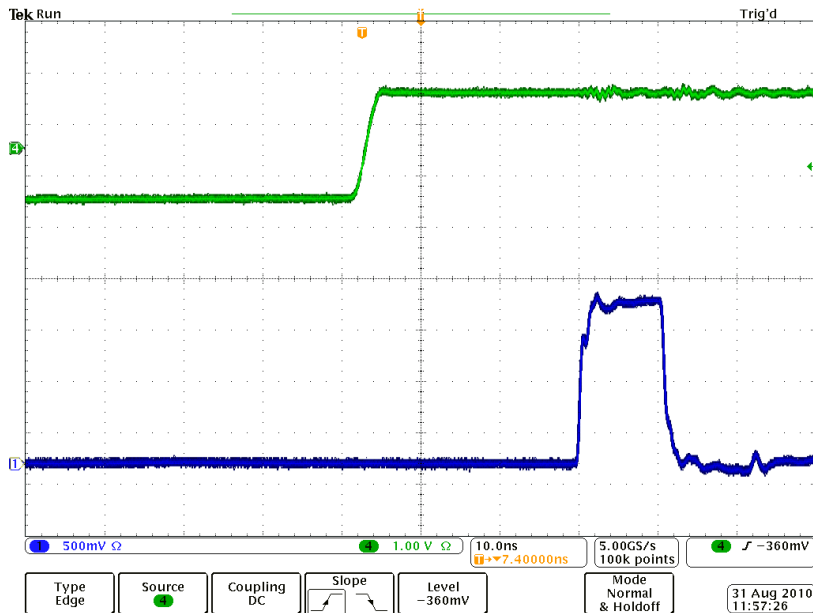
固定端での反射



固定端での反射



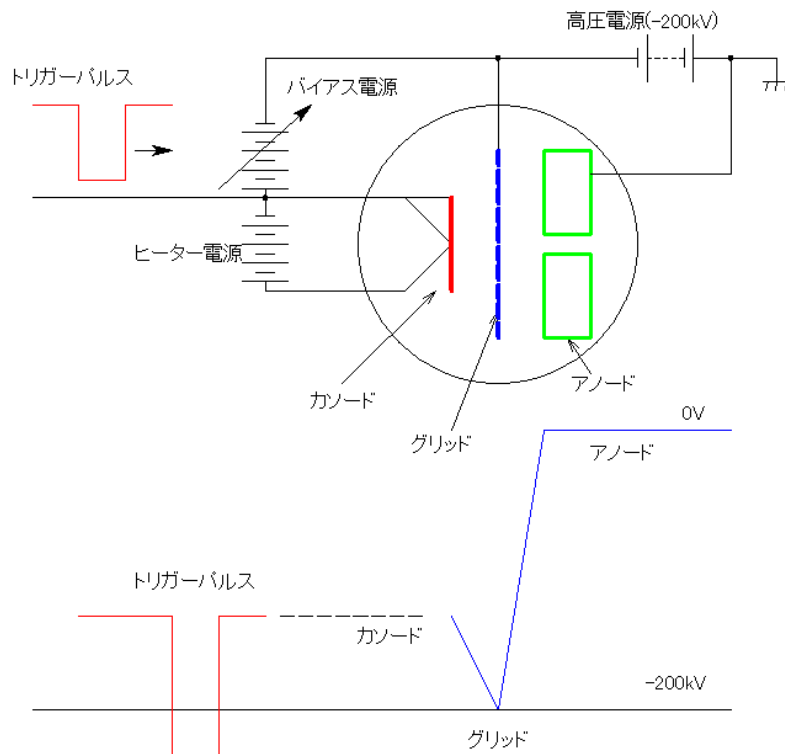
出力は



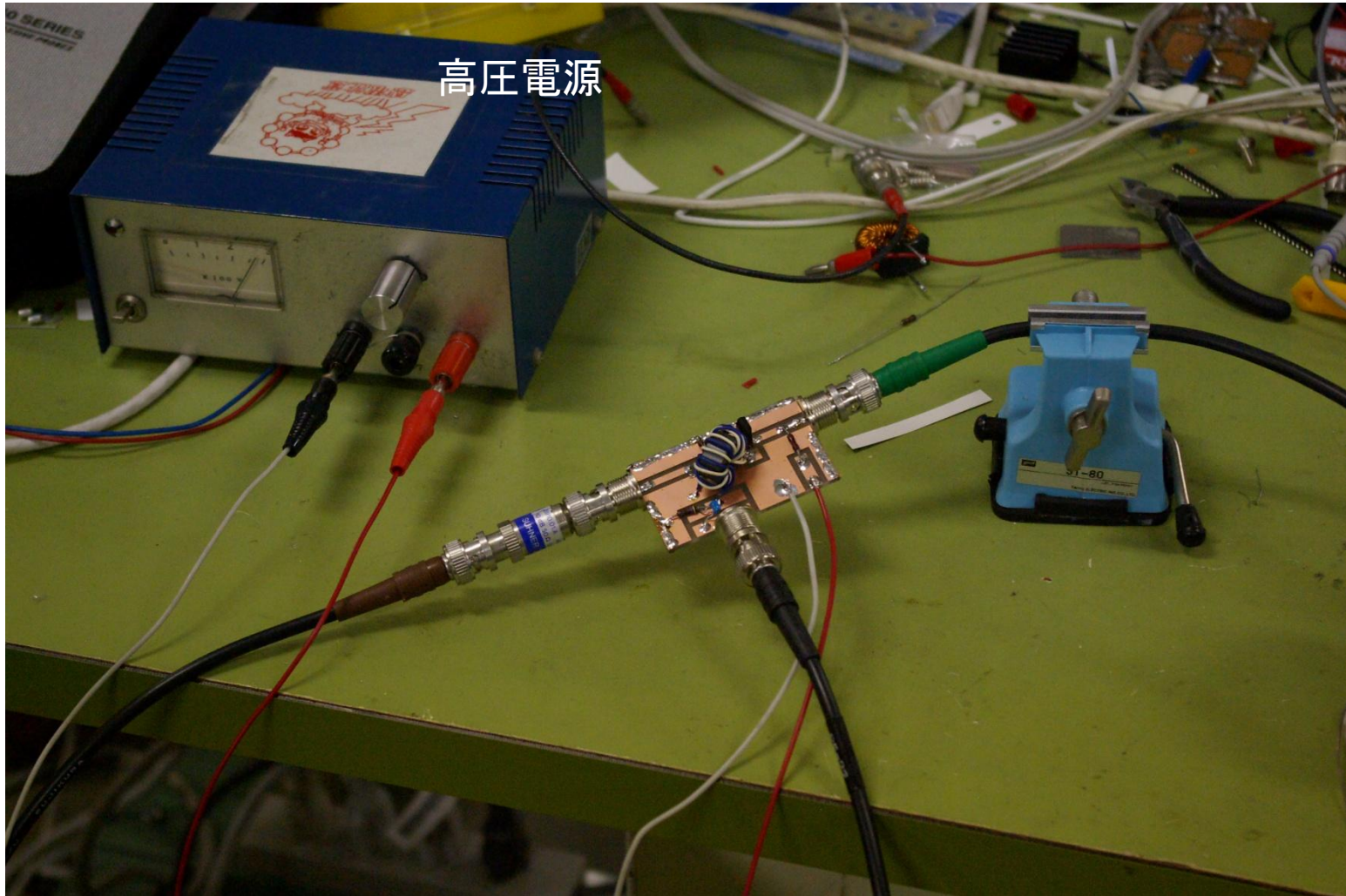
- トリガーが入ってトランジスタがONになり、約10ns後にOFFになっている。
- このとき、PFNには1mの同軸線をつけていた。ここにたまっていた高電圧が抜けて、パルス高が0に戻った。

アバランシェパルサーの使用例

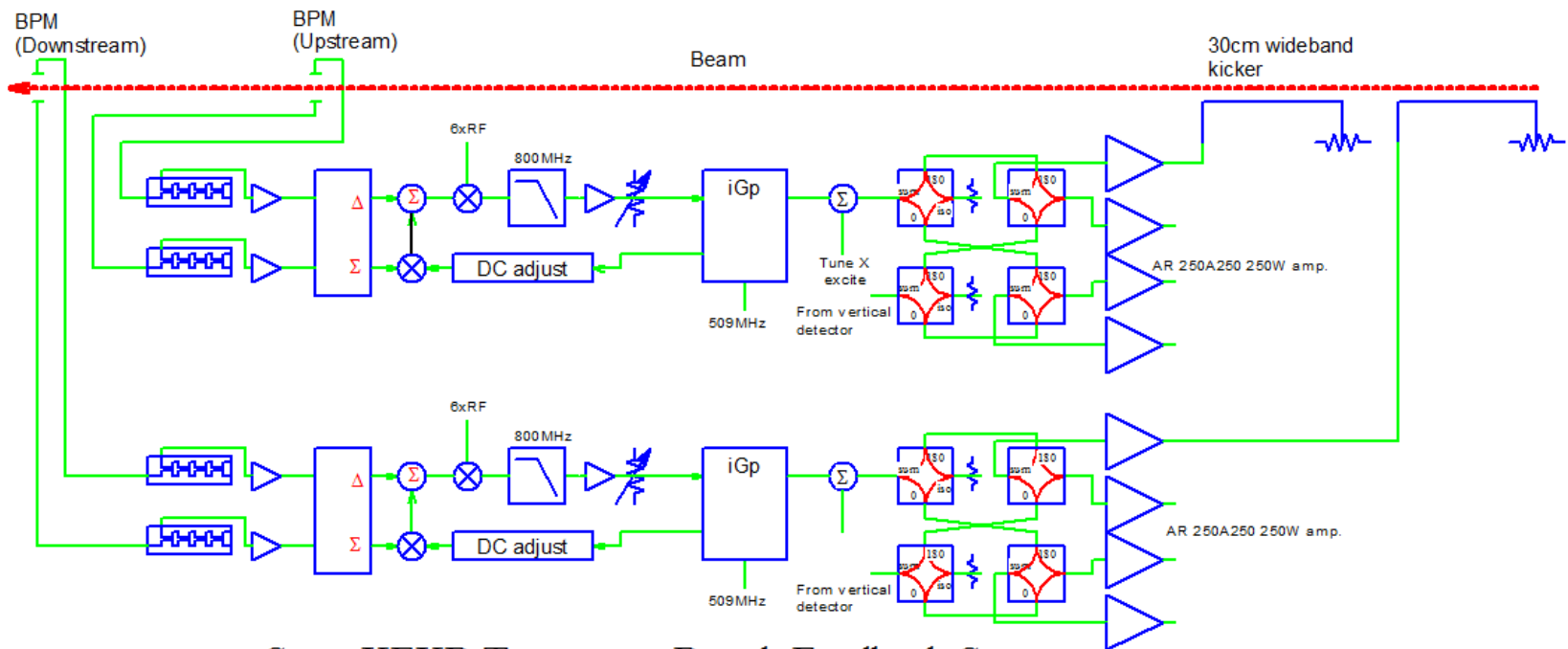
- 線形加速器の熱電子銃から出る電子ビームの長さを非常に狭く(1ns以下とか)したいとき、アバランシェパルサーを使ってビームが出るタイミングを制御します



高圧電源

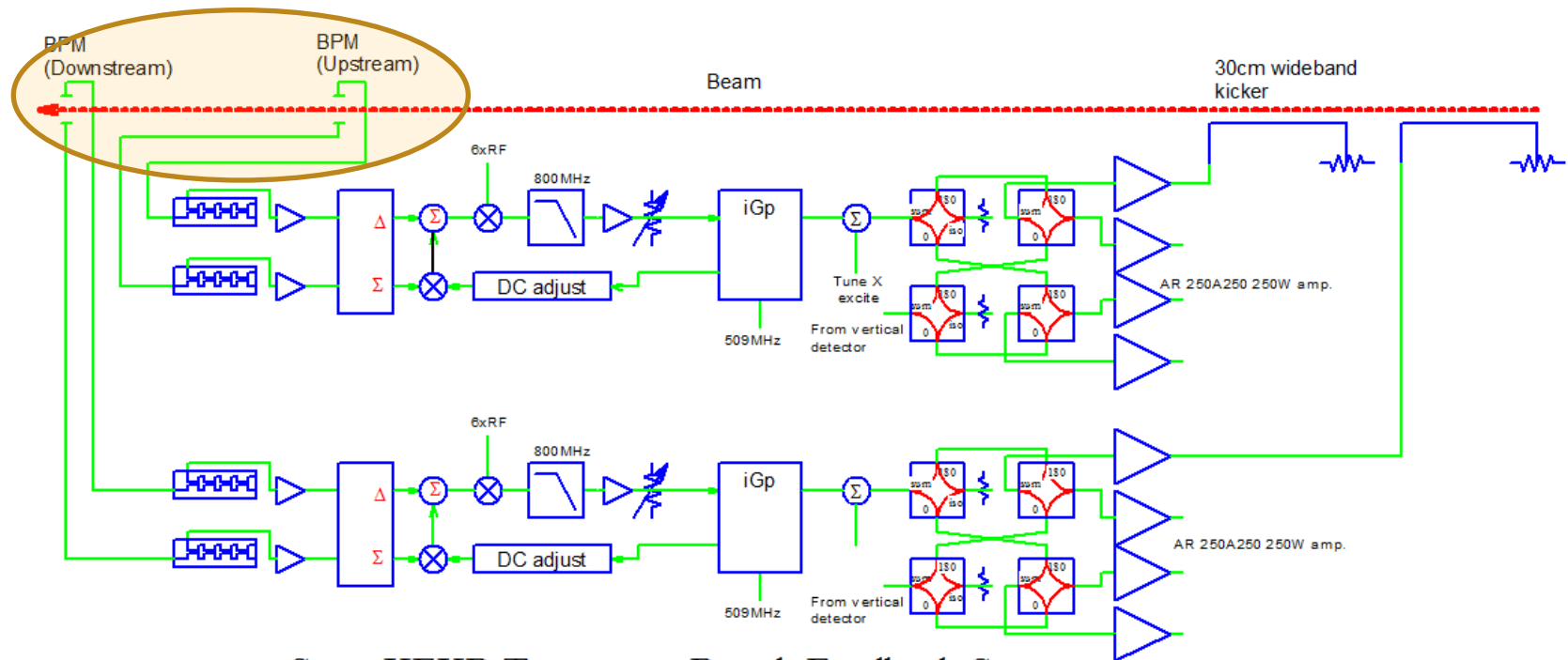


SuperKEKB Transverse FB plan



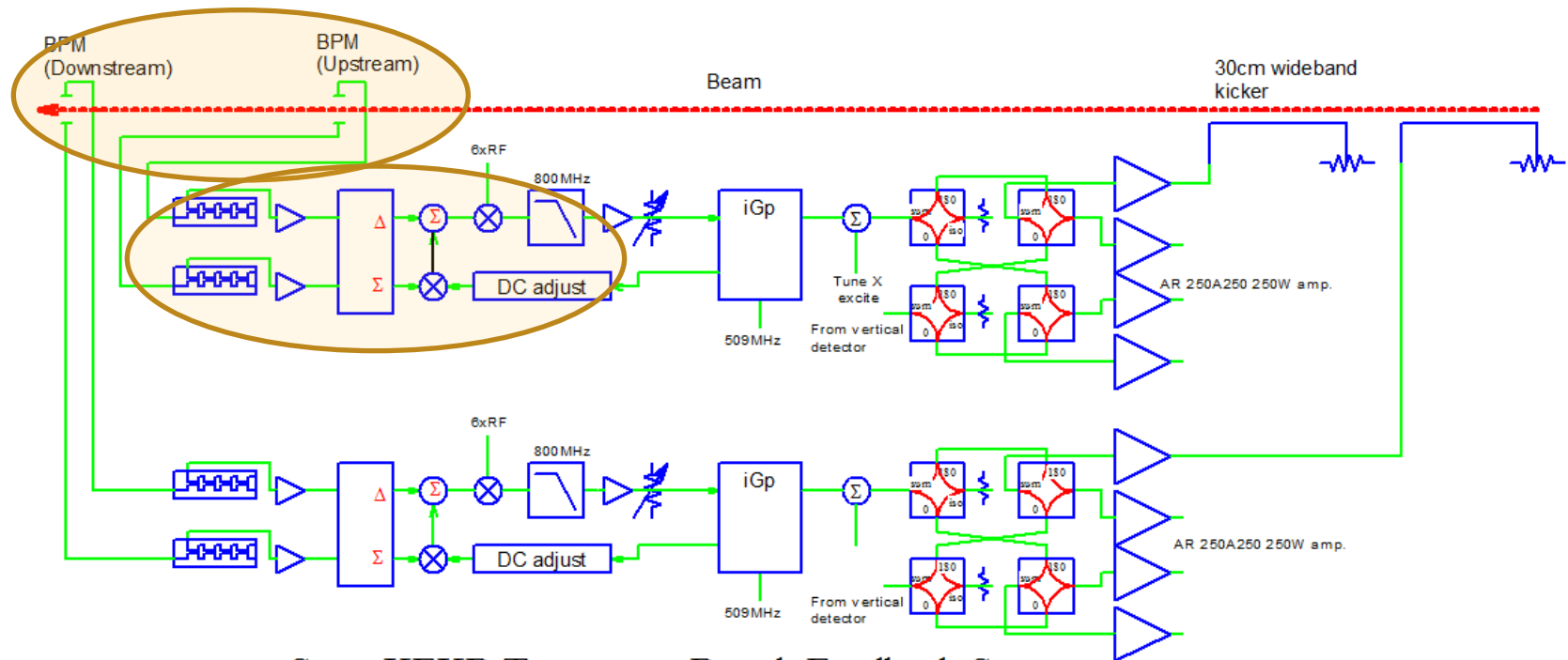
SuperKEKB Transverse Bunch Feedback System

SuperKEKB Transverse FB plan



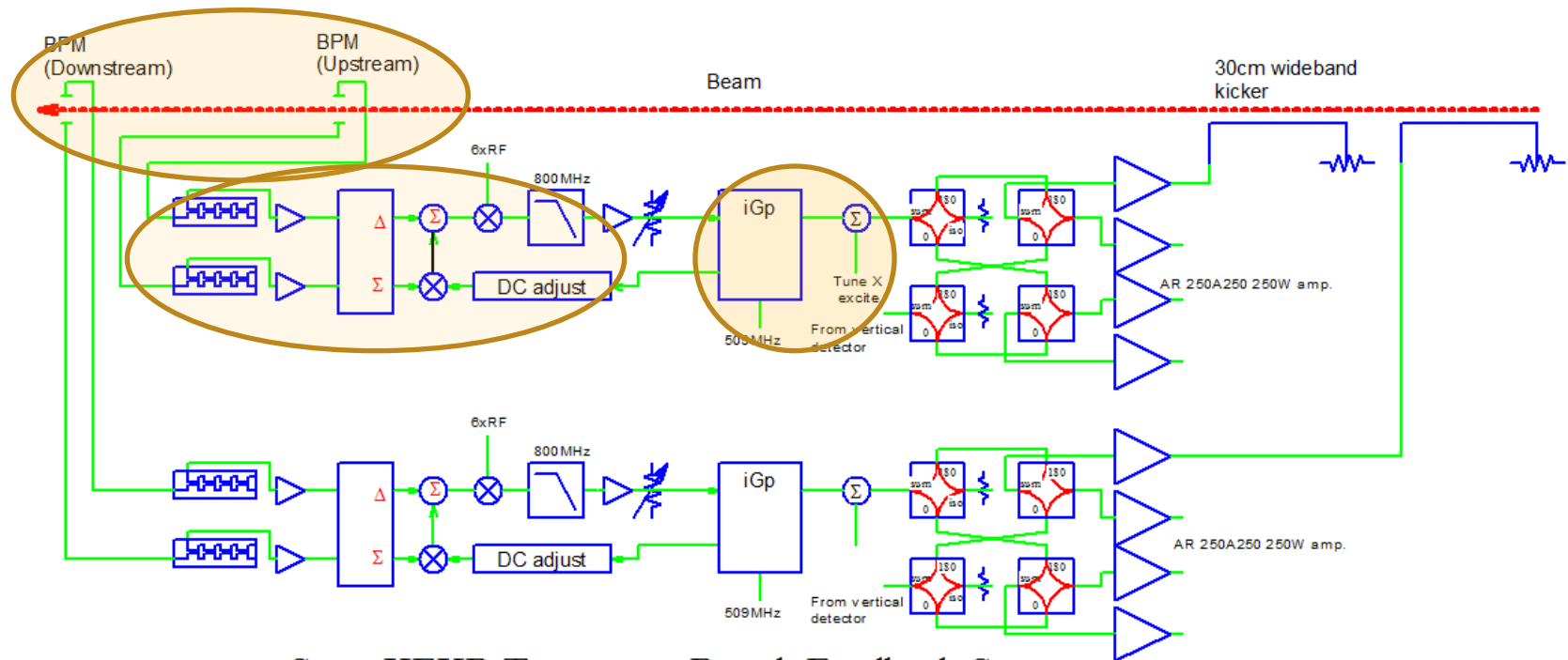
SuperKEKB Transverse Bunch Feedback System

SuperKEKB Transverse FB plan



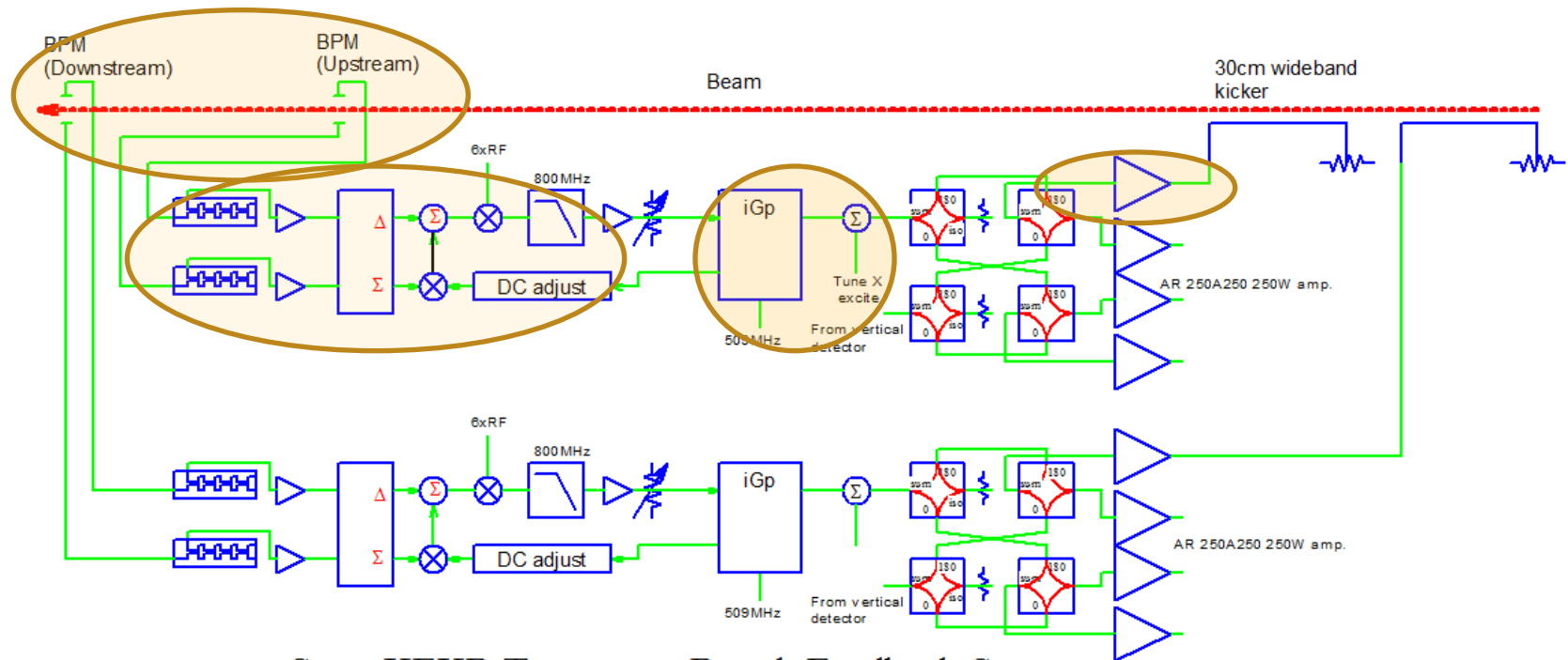
SuperKEKB Transverse Bunch Feedback System

SuperKEKB Transverse FB plan



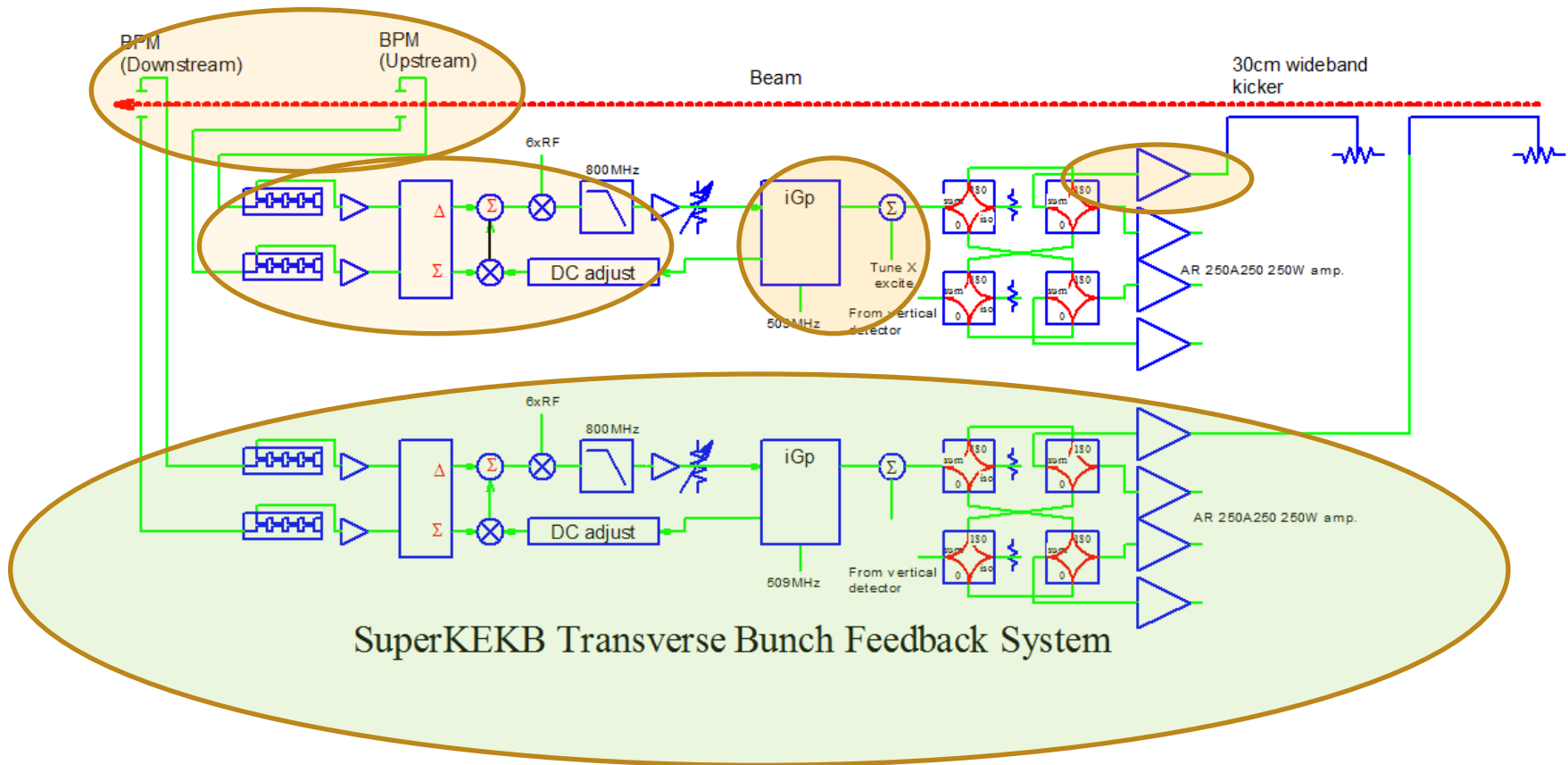
SuperKEKB Transverse Bunch Feedback System

SuperKEKB Transverse FB plan

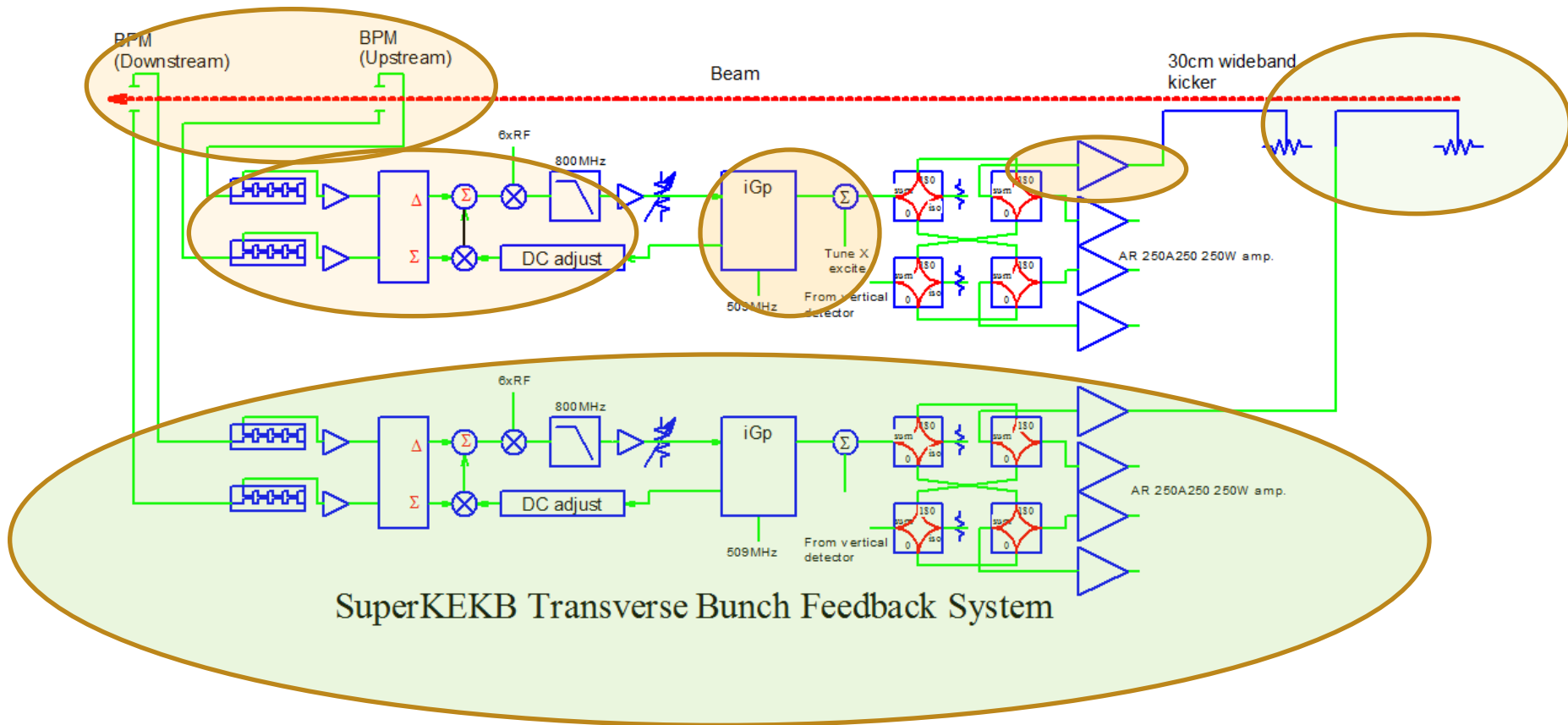


SuperKEKB Transverse Bunch Feedback System

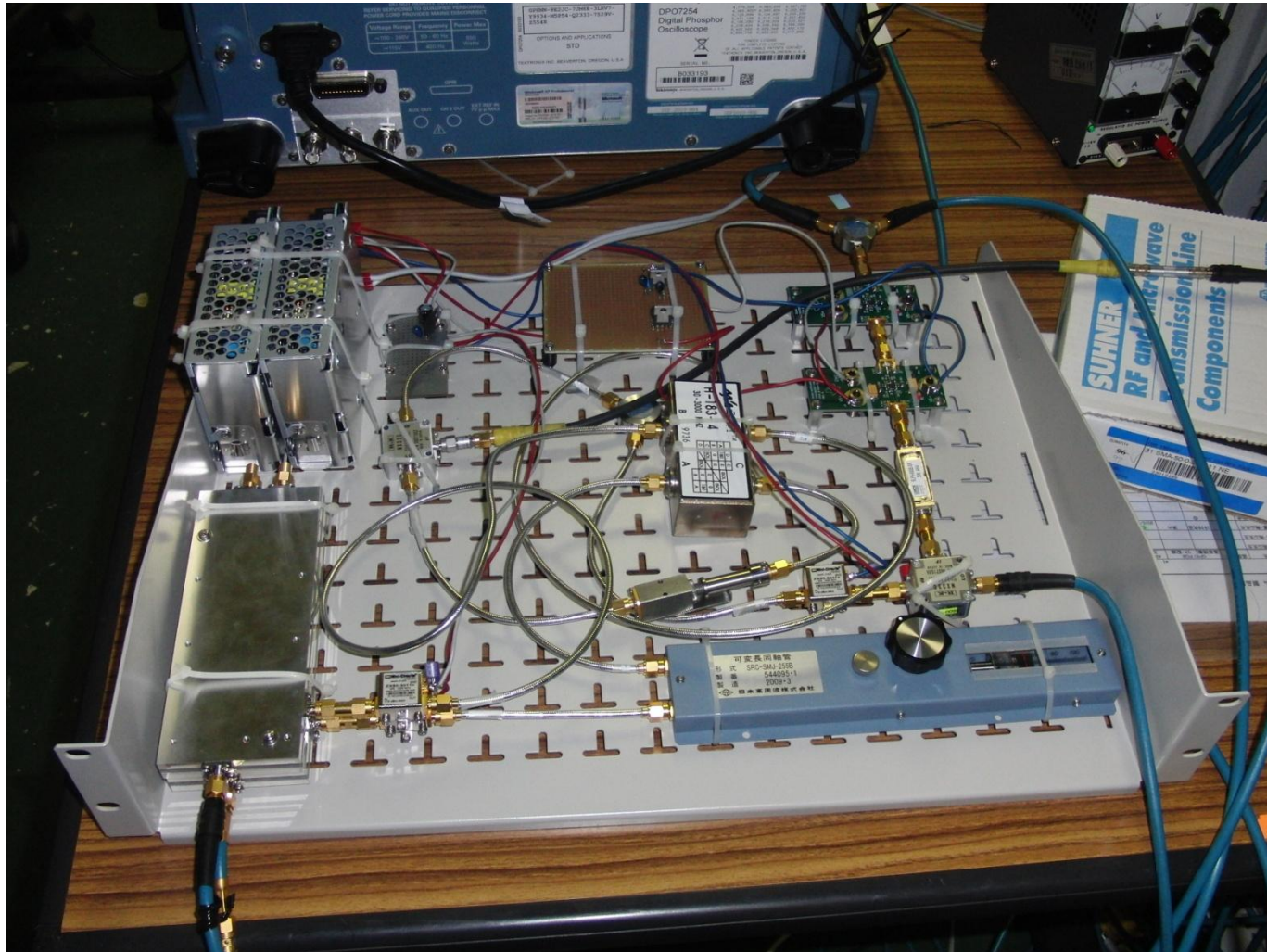
SuperKEKB Transverse FB plan



SuperKEKB Transverse FB plan



Bunch position detector prototype



Vacuum chamber

- Aluminum alloy antechamber
- cutoff frequency < 1 GHz

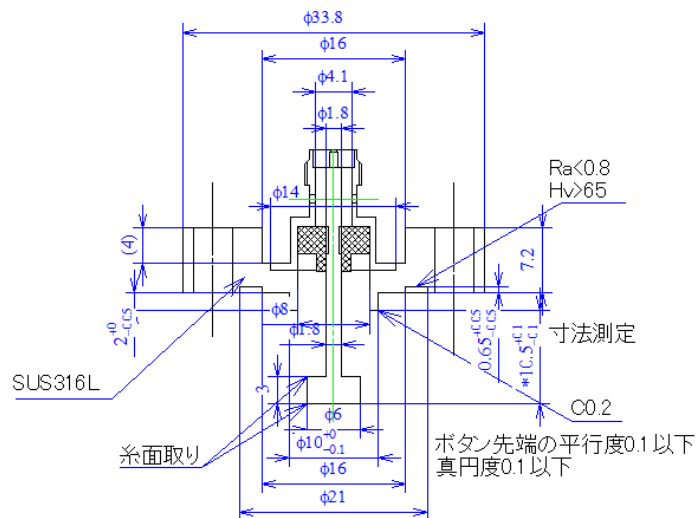
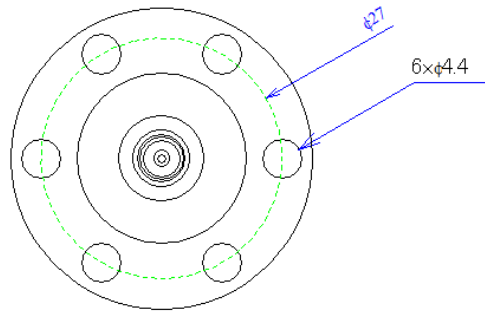
Aluminum-alloy duct



Aluminum-alloy duct



BPM head



SuperKEKB用BPM model-E1
 作図: M.Tobiyama 6/Oct/2006
 修正: M.Tobiyama 8/Nov/2006



Impedance/button output simulation

