Results of Magnetic Measurements and Field Integral Compensation for the Elliptical Multipole Wiggler

D. Frachon, P. M. Ivanov, E. A. Medvedko, I. Vasserman, O. Despe, Y. G. Kang
APS, Argonne National Laboratory, Argonne, IL 60439

I. ABSTRACT

A prototype of the Elliptical Multipole Wiggler (EMW) [1] has been assembled, tested and tuned at the APS. This prototype has a period of 160 mm with 7 poles for the hybrid structure and 10 poles for the electromagnet part of the EMW. The hybrid structure of the EMW produces a vertical magnetic field of 0.83 T with $K_y=12$ for a gap of 27 mm, and the electromagnetic structure provides a horizontal field change up to 100 Hz with a maximum field of 0.12 T ($I=0.6$ kA, $K_x=1.6$). The current pulse has a trapezium-type shape with a switching time to change the current polarity of about 2 ms. The measurements and tuning were done for direct current (DC) mode and alternating current (AC) mode. Fine adjustment during the test at the NSLS X-ray ring using the BPMs and active correction system allowed to achieve about 1 μm of beam distortion. It corresponds to the peak-to-peak variations during the time less than ±0.5 G-cm and ±100 G-cm$^2$ of the first and second horizontal field integrals respectively.

II. INTRODUCTION

Two different correction systems were used to adjust the first and the second field integrals dependence on time. A passive correction system includes manually adjustable gaps for the end poles. The adjustment of the first horizontal field integral was performed by moving the gap of each end pole in the opposite direction. The adjustment of the second horizontal field integral was performed by moving the gap of each end pole in the same direction without distortion of the first field integral due to the antisymmetric configuration of the device. The active correction system is based on the use of a set of two trim magnets mounted on each end of the device. The magnet coils are fed by a power supply with an arbitrary function generator [2].

III. MAGNETIC MEASUREMENT TECHNIQUE

The conventional Hall probe technique was used for measurements of the magnetic field distribution in the longitudinal direction. The rotating coil magnetic measurement technique was used for field integral measurements. Actually the rotating mode was necessary only for the DC mode. For the AC mode, the FAST 16-1 ADC board was used with a sampling time of 0.1 ms for a frequency of 100 Hz. The CTM05 board was installed on the PC bus and used to generate pulses to define the frequency of the power supply and to trigger the FAST 16-1 ADC board to synchronize measurements. The integration of the signal from the coil provides the flux dependence on time through the coil. The most important part of the EMW measurements is the field integral dependence of electromagnetic structure on time for the AC mode. Measurements of the second field integral by the usually applied technique of measuring the field map are very time consuming. The reason is that such measurements are based on the step-by-step motion of magnetic sensors along the main axis of the device and on time-dependent measurements at each point. That is why a novel technique using a twisted long coil was used for fast and precise magnetic measurements of second field integrals [3].

IV MAGNETIC FIELD MEASUREMENTS

The results of Hall probe measurements of the vertical and horizontal field distributions are shown in Fig. 1.

![Fig. 1. Horizontal and Vertical Field vs. Z](image_url)
the calculation of the second horizontal field integral from the Hall probe data is shown in Fig. 2.

Fig. 2. Vertical Trajectory in the DC mode

V. FIRST FIELD INTEGRAL MEASUREMENTS

A conventional long coil with parallel wires was used for first field integral measurements. For measurements of hybrid structure and of the electromagnet in the DC mode, coil rotating at 360° was used. Actually these measurements were complementary to the main set of measurements that was done in the AC mode without rotating the coil. The results of the AC first horizontal field integral measurements are shown in Fig. 3 with active correction on and off. The first horizontal field integral change during the cycle is less than 10 G-cm without active correction and less than 1 G-cm with active correction switched on.

Fig. 3. First Horizontal Field Integral vs. Time. f = 100 Hz

The time dependence of the field integrals can be divided into two parts:
1. DC part. This part is a result of different signs of the current and can be easily adjusted by choosing the proper gaps for the end poles. This part manifests mainly for low frequencies (less than 10 Hz).
2. AC part. The length of this part is much longer than the switching time (2 ms) and is about 50 ms. It is due to an eddy-current-produced delay of the field penetrating to the air space. Small differences in the design produce different delay times for different parts of the device and result in field integral dependence on time. The only way to correct for this is to apply an active system fed by a special power supply with an arbitrary function generator. Both parts of the first field integral time dependence can be easily seen at a frequency of 1 Hz (Fig. 4) without dynamic correction. Only the DC part was adjusted here by means of a passive end-correction system.

Fig. 4. First Horizontal Field Integral vs. Time. f = 1 Hz

The dependence of the first and second vertical field integrals on time exists due to the saturation effect of the electromagnetic field on hybrid structure. There was no active correction system for the vertical direction at the time of the measurements, and the change in the field integral therefore is bigger than that for the horizontal direction. The first vertical field integral change during the cycle is less than 45 G-cm. The results of the measurements of the AC first vertical field integral are shown in Fig. 5.

Fig. 5. First Vertical Field Integral vs. Time. f = 100 Hz

VI. SECOND FIELD INTEGRAL MEASUREMENTS

Coil twisted by 180° was used to obtain the second field integral from these measurements. At such a configuration, the expression for the second field integral dependence on the measured magnetic flux and the first field integral is [3]:

\[ I_2(L) = -\frac{\Phi}{\Theta} + L \cdot I_1(L), \]

where: \( I_1(L) \) and \( I_2(L) \) are the...
first and second field integrals, respectively; $\Phi$ is the magnetic flux through the coil; $L$ is the half length of the coil; and $\Theta$ is the crossing angle of the coil. This expression becomes especially simple and allows one to achieve the most reliable and precise results in the case in which the first field integral is equal to zero.

The results of second field integral measurements are shown in Fig. 5 at a frequency of 100 Hz. Change in the second horizontal field integral during the cycle is less than 4000 G-cm$^2$ without active correction and about 1000 G-cm$^2$ with active correction switched on. Change in the second vertical field integral change during the cycle is less than 3000 G-cm$^2$.

Fig. 6. Second Horizontal and Vertical Field Integral Dependence on Time. $f= 100$ Hz

VII. INTEGRATED MULTIPOLE COMPONENTS

There were no special requirements for multipole components, but this question is rather important from the point of view of beam life time and beam dimensions in the storage ring. So a set of measurements was performed in order to obtain the dependence of the first field integral on the horizontal position (X) for both the AC mode and the DC mode. The DC mode originates mostly from the hybrid structure, and the AC mode originates from the electromagnetic part and time-dependent part of the hybrid structure due to a saturation effect induced by the electromagnet. The results of measurements of the DC mode obtained from the rotating coil measurements are shown in Table 1.

Table 1. Integrated Multipole Components at DC mode

<table>
<thead>
<tr>
<th>Skew components</th>
<th>Quadrupole (G)</th>
<th>Sextupole (G/cm)</th>
<th>Octupole (G/cm$^2$)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Normal Components</td>
<td>183.</td>
<td>29.</td>
<td>29.4</td>
</tr>
</tbody>
</table>

The results of the time-dependent part of the multipole components are shown in Table 2. For each X position, the dependence of the first field integral on time was obtained and the RMS value was calculated. The results, shown in Table 2, correspond to dependence of these RMS values on X.

Table 2. Integrated Multipole Components in the AC mode

<table>
<thead>
<tr>
<th>Skew Components</th>
<th>Quadrupole (G)</th>
<th>Sextupole (G/cm)</th>
<th>Octupole (G/cm$^2$)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Normal Components</td>
<td>-1.38</td>
<td>11.6</td>
<td>-13.8</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Normal Components</th>
<th>Quadrupole (G)</th>
<th>Sextupole (G/cm)</th>
<th>Octupole (G/cm$^2$)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Normal Components</td>
<td>0.13</td>
<td>4.39</td>
<td>0.91</td>
</tr>
</tbody>
</table>

VIII. CONCLUSION

The first tests of the EMW at the NSLS X-ray ring with frequencies of 2 Hz and 100 Hz were successful and showed rather good performance of the device [1]. Further improvements in vertical magnetic field are possible with the help of an additional set of trim magnets. This system is under construction now and will be incorporated into the EMW later.

IX. ACKNOWLEDGMENTS

Work performed under contracts W-31-109-ENG-38 and DEAC-02-76-CH-00016 of the U.S. Department of Energy.

IX. REFERENCES

1. E. Gluskin et al. "The Elliptical Multipole Wiggler Project," This conference
