

PULSED UNDULATORS FOR HIGH EFFICIENCY FEL OSCILLATORS USABLE IN THE VISIBLE SPECTRUM

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ABSTRACT The problem for reaching high efficiencies in RF-linac driven FEL's, is to cross the perturbed region between the small signal and the large-signal operation due to large phase shifts variations.

We consider the FEL as a Traveling-Wave-Tube (TWT) or the reverse of a tapered buncher as in an electron accelerator. The equivalent transverse gradient is

$$E_{||} = \frac{K}{2\gamma} E_{\perp}$$

The difference with a TWT is that the EM beam instead of being guided inside a waveguide, is in free space

A- ASSUMPTIONS :

1-multi-pass operation : the build-up of the oscillation requires 100 to 200 turns (order of 1μs)

2-gain per pass 1.02-1.03 in gradient to compensate losses and useful power : $E = \text{Const}$

3-efficiency : energy transfer - electrons-to-EM - at each pass: For a 20% efficiency, the product

$$P(1 + \frac{K^2}{2})$$

has to change about 40-50% in value along the undulator during the steady-state operation.

4-spontaneous emission does not contribute to a substantial amount to the exchange of energy between the electron beam and the EM wave. Harmonics can be considered, eventually as a nuisance for the mirrors, or as a low power source.

5-Coupling impedance :

Two conditions are necessary for a substantial exchange of energy between beam and RF :

a/ the beam current must carry a large component at the frequency of operation, -means good bunching-

b/ the line must create a coupling impedance Z which, in the case of a perfect undulator and a transverse wave, in ideal conditions, is simply

$$Z = \frac{K}{\gamma} \sqrt{\frac{\mu_0}{\epsilon_0}} = \frac{K}{\gamma} \eta$$

This is true as well for the fundamental or harmonics if we would like to generate them with an acceptable efficiency.

B-EQUATIONS

p is the period and K (.94 B.p) the conventional undulator factor. They are defined for one full period

Since the period and the magnetic field may vary, we introduce a coefficient proportional to the inverse of phase velocity, versus a reference given by p_0 and K_0 :

$$\alpha^2 = p_0 (1 + \frac{K_0^2}{2}) / p (1 + \frac{K^2}{2})$$

The abscissa is referred to the period or fraction of period : x is the independant variable

$$dz = p(x) dx$$

and we use the two equations for phase and energy along the line :

$$\frac{d\Psi}{dx} = 2\pi \left[(\alpha - 1 + \alpha u) \frac{\alpha + 1 + \alpha u}{\alpha^2 (1 + u)^2} \right]$$

$$\frac{du}{dx} = -\frac{K^2}{1 + \frac{K^2}{2}} \cdot \frac{E\lambda}{m_0 c^2} \cdot \frac{1}{\alpha^2 (1 + u)^2} \cdot \sin \Psi \quad u = \frac{\gamma}{\gamma_0} - 1$$

In these equations, the wavelength enters as the scaling parameter $E_0 \lambda$.

The "transfer coefficient" a is the sum of the values of u over all phases calculated at each period.

C- UNIFORM UNDULATOR and UNIFORM-TAPERED :

Fig 1 and 2 summarize the results in those two conventional situations. The data come in two sets of curves : evolution of phase and energy along the line for each value of the RF level ($E\lambda$), and value of the transfer efficiency $\gamma E\lambda$.

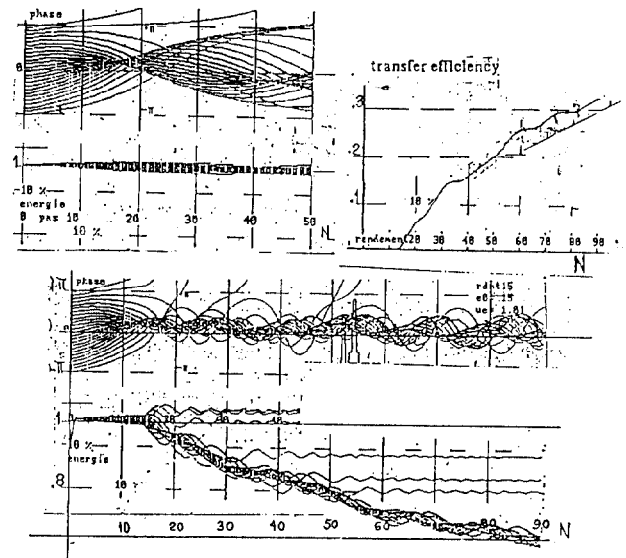


FIG 1. Phase and energy diagrams.
TOP : uniform undulator $E\lambda = 1\text{kV}$
BOTTOM : Buncher + tapered undulator $E\lambda = 15\text{kV}$
the efficiency displayed here is 33% with a 90 periods undulator, including 14 periods for the buncher

For the uniform undulator, and $E\lambda < .5 \text{ kV}$, the transfer coefficient is in $(E\lambda)^2$, (slope 2 on the graphic Fig 2) as in the conventional small-signal theory.

The efficiency levels off at about 1-1.5% and starts to drop if the RF level is made to increase above 1 kV.

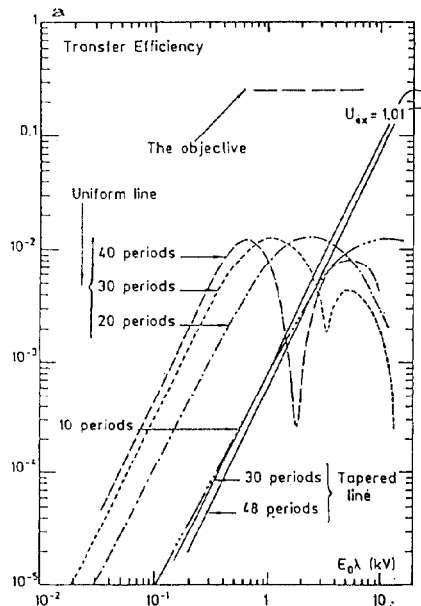


FIG 2 Transfer coefficient vs $(E\lambda)$

With the tapered undulator, and values of the parameter $E\lambda$ above a few kVolts, we are in the "large-signal" conditions. The bunching is nearly complete. The efficiency of transfer is then roughly proportional to $E\lambda$ instead of $(E\lambda)^2$ as in the small signal situation.

The beam intensity which can sustain the oscillation is:

$$I_{(Amp)} = \frac{b}{\xi} \cdot \frac{1}{4\eta} \cdot \frac{e}{m_0 c^2} \cdot \frac{2}{1 + \frac{K^2}{2}} \cdot \gamma \cdot \frac{\pi \epsilon_n}{\gamma \lambda} \cdot ff \quad \xi = \frac{a}{N(E\lambda)^2}$$

ff : form factor of the RF beam (order of 1.4)

b : loss coefficient per turn of the RF (3-5%)

It is well known and can be seen on Fig 2, that such a tapered line has no gain at small signal (slope below .5 kV larger than 2) : oscillation does not start !

D- BUNCHER+STRONG TAPER +UNIFORM+SLIGHT TAPER

A high gain structure must be added, of course not *ahead* of the tapered part (it would behave as above)!, but *after*. And we arrive at a geometry similar to the one already described and tested⁹.

At small signal, the tapered part acts as a delay line between the buncher and the uniform line. The small signal coming out of the buncher is amplified and must arrive at the correct phase in the line. In the large signal operation, buncher + tapered portion behave as previously.

However there are very strict limitations in this operation : The number of periods of each part is critical, in order to get the optimum phase shift.

For monoenergetic electrons, there always remains a "hole" in the efficiency curve, between .5 to 1 kV,

and the system would stop building up. Fortunately, this hole is not exactly at the same place for different electron energies, within 0.4%. So, if the available current in the beam were sufficient, this critical transition could be passed

E- PROGRAMMING THE PROFILE DURING BUILD UP OF OSCILLATION

This will be the correct solution : change the line profile, so that the profile is continuously adjusted to the level of RF power. In a TWT, this, of course, is impossible. But here, the build up covers 100 turns or more, which represents a duration of one, or a few, microseconds which is enough to program the currents which create the magnetic field in the undulator.

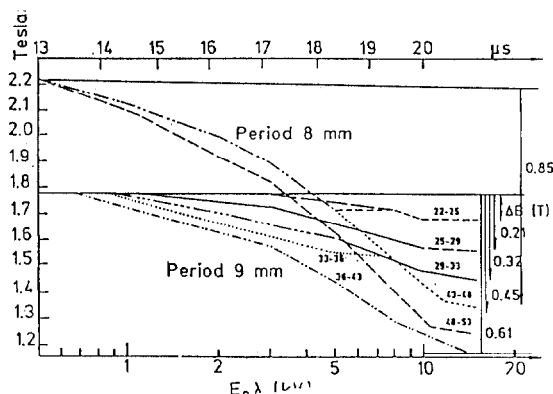


FIG 3

Variation of the magnetic field vs $(E\lambda)$ and N .

For $(E\lambda) < .5$ kV, there is no taper. Above .5 kV, the diagram is swept during the few microseconds of rise time.

Tapering starts just one or two periods after the buncher.

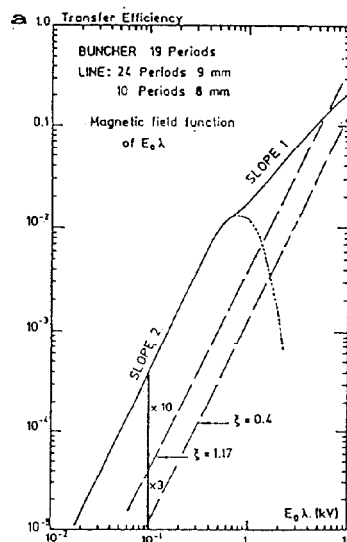


FIG 4

Transfer coefficient vs $(E\lambda)$. The transition between small and large signal is smooth, and efficiencies larger than 20% can theoretically be obtained with 35 - 40 periods (not including the buncher).

Of course, this implies that the undulator has no magnetic material, and is short enough so that the reactances are kept small.

Fig 3 gives the slopes of the taper vs the RF level.

In the large signal region , $(E\lambda)$ increases of an equal amount per turn , and since each turn has the same duration , dE/dt is a constant .

Fig 4 is a typical plot of the transfer coefficient , for a 34 period-line preceeded by a 19 period-buncher , which shows a 20% efficiency at a value of $(E\lambda)$ not too high for the mirrors .

It can be seen that the bunching is of the order of 80% , so that side-bands excitation has a very low efficiency ∞ and the bunches are 30-40 degrees in phase so that the space-charge fields stay below a few percent of the main RF field ∞ . The "transfer coefficient" can be of the order of .4-.5 % per period

F- TECHNOLOGY

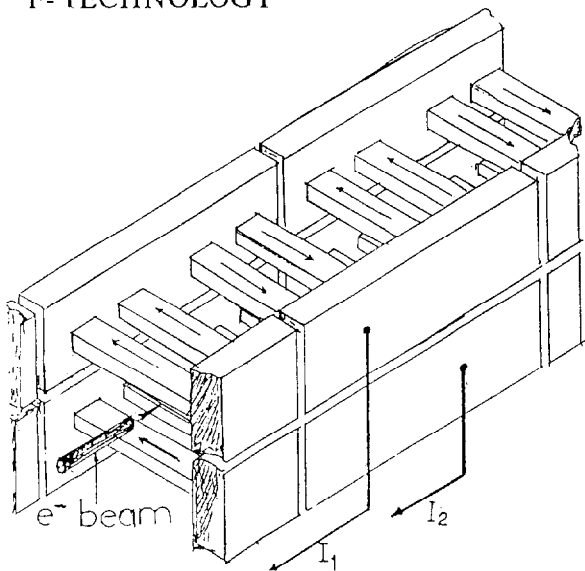


FIG 5 Schematic view of an undulator
I1 and I2 : Derivation currents to control $B(z)$ vs time.

The major difficulty is obviously to provide sufficient cooling capacity . The undulator presents itself as two layers of " bars " (1 or 2 mm cross-section , and 10 to 20 mm long) . Copper bars can be cooled at the two ends if 100-150 degrees C can be tolerated between the middle and the ends . The cooling pipes are four- titanium or stainless steel -spirals , cooling the four ends of the successive bars . The period of the spirals is the same as for the undulator . Pipes are available down to 2 mm in diameter , so that the mechanical limit of this type of technology is around 2.5 mm period for the undulator.

The maximum current which can be passed through such a "bar" is

$$I_{(Amp)} \cong 433 \cdot \sqrt{\frac{s}{l} \cdot \frac{\Delta T}{f \tau}}$$

where : s and l are the cross-section and length of the bar (mm)

f τ is the duty factor

ΔT is the maximum temperature difference.

for a temperature difference of 100 °C and a duty factor of 1/100 , the period is limited down to 7 mm for $K=1.4$ and 3mm with 200 °C and 1/300 duty factor

In the end , the major difficulty will be in the mirrors

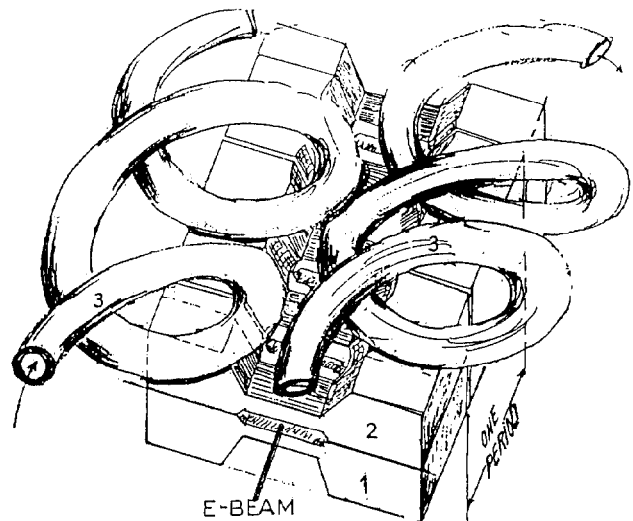


FIG 6 Arrangement of the cooling pipes

G- HARMONICS ENHANCEMENT

If the "bars" corresponding to each half period are divided into two or more "sub-bars" in parallel , as in Fig 5 , it is possible , by changing the relative distances , to have a K_n value of the coupling parameter even larger for the harmonics than for the fondamental as it is shown in Fig 7 .

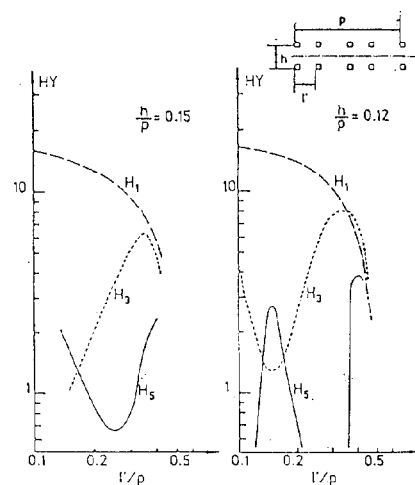


FIG 7 Coupling factor K for the fundamental and harmonics 3 and 5 , for the structure represented on Fig 5 : Two bars in parallel for each half period .

* Variable-Wiggler FEL oscillator

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** Development of side-bands in tapered and untapered FEL's
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