

FEASIBILITY STUDY OF A TWO-MILE SUPERCONDUCTING LINAC*

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Summary. Assuming that investigations now underway will result in superconducting accelerator structures capable of withstanding gradients of 33 MeV/m, a feasibility study of a two-mile 100 GeV superconducting electron linac with 6% duty cycle has been made at frequencies of 1428 and 2856 MHz. Tentative machine parameters and the preliminary design of components and systems have been examined. These studies are based upon a traveling-wave accelerator structure with RF feedback in each 20-ft section.

Introduction

As SLAC is completing its third year of operation, the two-mile linear accelerator has either met or surpassed all its original design goals.¹ It is expected that the experimental physics program presently underway will exploit the usefulness of the machine for several years. Meanwhile, in keeping with new developments, the machine physicists at SLAC are looking into various long range possibilities of upgrading the accelerator. The feasibility study presented here is only one of several alternatives. Moreover, many of the design parameters given here may be modified several times during the course of further studies.

Selection of Basic Parameters

In this section, a set of basic parameters for a two-mile 100 GeV superconducting accelerator is chosen. To the maximum extent consistent with technical and physics requirements, these parameters are selected to be compatible with the characteristics and specifications of the utility systems and physical plant of the existing SLAC two-mile accelerator. In the paragraphs which follow, the factors influencing the choices of the basic parameters are discussed. These parameters are tabulated in Table I.

Energy

The energy goal of 100 GeV has been chosen since this energy is sufficiently higher than the 20 GeV capability of the present SLAC accelerator to be of great interest in physics research and yet within anticipated fiscal and technical limitations. At 100 GeV in a length of 3 km, the energy gradient of the proposed conversion is 33 MeV/m. The theoretical limit for niobium, the presently favored material for such an accelerator, is from 50 to 70 MeV/m depending on the choice of accelerating mode. This limit is due to the critical magnetic field. It is likely that the actual limit will be lower, either because of field emission or because the practical critical magnetic field for an RF cavity is less than that measured with test samples at low frequencies.

Although to date no accelerator cavity tests have achieved 33 MeV/m, the most recent measurements with niobium are very encouraging. These measurements made at the High Energy Physics Laboratory of Stanford

TABLE I
PARAMETERS OF A TWO-MILE 100 GeV SUPERCONDUCTING
ACCELERATOR AT 2 FREQUENCIES

Parameter	f = 1428 MHz	f = 2856 MHz
Length	3000 m	3000 m
r_0/Q	$2.15 \times 10^3/\text{m}$	$4.3 \times 10^3/\text{m}$
r_0	$3.44 \times 10^{13}/\text{m}$	$1.72 \times 10^{13}/\text{m}$
Q	16×10^9	4.0×10^9
Loaded Energy (max)	100 GeV	100 GeV
V/L	33 MeV/m	33 MeV/m
Duty Cycle	1/16	1/16
Peak Beam Current	48 μA	48 μA
Average Beam Current	3 μA	3 μA
Peak Beam Power	4.8 MW	4.8 MW
Average Beam Power	0.3 MW	0.3 MW
Number of Klystrons	240	240
Peak Power per Klystron	20 kW	20 kW
Average Power per Klystron	1.25 kW	1.25 kW
Type of RF Structure	TW with RF feedback loop	TW with RF feedback loop
Number of Accelerator Sections	480	480
Length of Accelerator Section	20 ft	20 ft
Power Dissipated in Accelerator	6000 W (2 W/m)	12,000 W (4 W/m)
Maximum RF Pulse Length	1.00 μs	0.25 μs
Filling Time (to 63.2%)	72 ms	18 ms
Maximum Beam Pulse Length	0.95 μs	0.24 μs
Time-Off Between RF Pulses	15.0 s	3.75 s
Accelerator Attenuation		
Factor (τ)	4.62×10^{-7} nepers	37.0×10^{-7} nepers
Feedback Attenuation Factor (ρ)	0.462×10^{-7} nepers	3.7×10^{-7} nepers
Bridge Ratio (g)	2.19×10^4	0.546×10^4
P_0 at Maximum Efficiency	219 MW	54.6 MW

* These values of parameters in this table are based upon operation at 100 GeV. The ranges of these parameters for other beam energies are shown in Figs. 1 and 2 assuming constant RF dissipation in the accelerator.

University,² translate to about 21 MeV/m for a $2\pi/3$ traveling-wave structure.

The filling time of a superconducting accelerator is very long (typically 10 to 100 ms) compared to conventional "room-temperature" accelerators (0.5 to 5 μs). The maximum RF pulse length shown in Table I is limited by the allowable temperature rise in the accelerator structure during the pulse. Practically speaking, an RF source capable of producing full power for a period on the order of a second must, for cathode emission and heat dissipation purposes, be designed to operate in a CW manner. Moreover, the existing ac distribution system in the SLAC Klystron Gallery has sufficient capacity to supply the required power continuously. Therefore, the adopted design approach is based on initial operation at a 1/16 duty cycle but with ac and RF capability for a 100% duty cycle. Then, if warranted at a later date, the duty cycle at energies above 25 GeV could be increased by adding more refrigeration units to the system.

Choice of Frequency

Theoretically, operation of a superconducting accelerator at the lower microwave frequencies is favored³ because of the higher Q's and the resulting lower dissipative losses predicted at these reduced frequencies. The minimum feasible frequency is probably limited by the increasing physical size of the structure, the

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increasing filling time, and the more stringent phasing and impedance matching tolerances as the frequency is reduced. Moreover, in the case of the projected superconducting accelerator for SLAC, there is already in existence an expensive high power waveguide system which operates at a particular frequency, viz., 2856 MHz. Cost savings would obviously result if the same frequency were chosen for the superconducting accelerator so that the existing waveguide system could be used. However, it is not yet clear whether the choice of a relatively high frequency would result in sufficiently higher refrigeration costs to outweigh the cost savings just discussed. For this reason, it has been decided to carry out preliminary design studies and cost analyses at two frequencies, 1428 and 2856 MHz, to determine whether either frequency has a definite technical and/or cost advantage over the other. The parameters shown in Table I have therefore been calculated for both frequencies. The basic criteria (beam energy, beam current, beam power, and duty cycle) are the same at each frequency. However, the filling times, pulse lengths, and feedback loop characteristics (described later) are different at the two frequencies. Theoretical scaling laws for Q and shunt impedance r_0 versus frequency have been used (see next paragraph). Definitive experiments to determine the frequency dependence of these quantities at temperatures near absolute zero have not yet been made.

r_0/Q , r_0 , and Q

The starting points in determining these parameters are: (1) the measured improvement in Q at 1300 MHz and 1.85°K by a factor of 10^6 compared to the Q in the same structure at room temperature^{*}; and (2) the room temperature values at 2856 MHz of $Q = 1.3 \times 10^4$ and $r_0 = 56 \text{ M}\Omega/\text{m}$. The following scaling laws are then used to calculate the values given in Table I: $r_0 \propto \omega^{-1}$, $Q \propto \omega^{-2}$, and the improvement factor (value of Q at superconducting temperature divided by value at room temperature) $\propto \omega^{-3/2}$.

The room temperature values of 2856 MHz can probably be improved by using a somewhat more optimized cavity geometry but such potential improvements have not been taken into account here.

Selection of Accelerator Type

The ratio of peak-to-average fields in a traveling-wave accelerator is significantly less³ than in a standing-wave accelerator (a factor of ≈ 1.6 improvement appears to be achievable). For this reason, the tentative decision has been made to adopt a traveling-wave design with RF feedback. A schematic of one section of such an accelerator is shown in Fig. 1. The residual RF power at the end of the accelerator is fed back to the input end where it is combined with the source power P_S by means of a suitable waveguide bridge. The combined power P_0 is then fed into the accelerator. The bridge ratio, g , is defined as the ratio of the powers which the bridge is designed to combine. When the ratio of the feedback power to the source power is equal to the bridge ratio g and when the feedback phase is properly adjusted, P_0

will be $(1 + g)$ times the source power and the power P_L to the external load will be zero. For this to occur, it is necessary that $P_S(1 + g)/x^2 = g P_S$, or $g = (x^2 - 1)^{-1}$ where $x^2 = P_0/P_F$ is the attenuation due to beam loading and to wall losses in the accelerator structure and feedback loop. When this condition is met, the steady state power build-up ratio in the accelerator will be $P_0/P_S = 1 + g = [1 - (1/x^2)]^{-1}$. Thus, a large build-up ratio results when x^2 is small (close to unity), i. e., when a large fraction of the input RF power is fed back to the bridge. The design value of the bridge ratio must be correspondingly high as indicated by the condition given above. It is generally sufficient to design the accelerator section with a fixed value of g which is strictly correct only at the design level of beam loading. For other levels of beam loading, the energy gain will be somewhat less than the maximum obtainable (by varying g to meet the optimization condition) and P_L will not be zero. In this case, the beam energy is a linear function of beam current in accordance with the equation³ $V \approx (8\tau g r_0 l P_S)^{1/2} - 2\tau g r_0 l i$ where r_0 is the shunt impedance per unit length in ohms/m, τ is the RF attenuation in the accelerator structure in nepers, l is the length of the structure, and i is the peak beam current. The above equation is valid when the losses in the accelerator structure (τ) and feedback loop (γ) are very small and $1/2g \gg (\tau + \gamma)$.

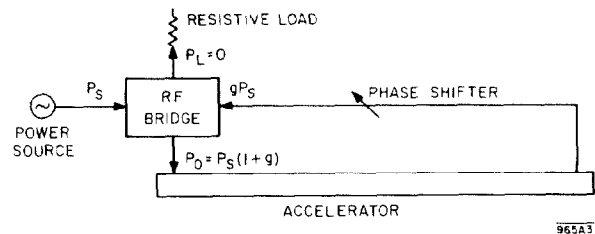


FIG. 1--Schematic of a TW accelerator section with RF feedback. The values shown are for the condition of maximum gain, i. e., $P_L = 0$.

Duty Cycle

The possibility of achieving a high duty cycle is one of the attractive features of a superconducting accelerator. However, adopting a 100% duty cycle for a projected 100 GeV two-mile machine would lead to excessive refrigeration costs on the basis of present-day cost experience and estimates. Therefore, the duty cycle of the superconducting machine has been limited to 1/16 (~6%) when the machine is operating at the full 100 GeV energy level. Even so, this is 100 times the duty cycle of the existing SLAC accelerator and would result in greatly improved experimental statistics.

When the energy is less than 100 GeV, the duty cycle can be increased without exceeding the capacity of the refrigeration system. For constant RF dissipation in the machine at a given beam energy of 25 GeV or lower, the duty cycle can be 100% if desired without exceeding refrigeration specifications.

Average Beam Power

For initial operation at 1/16 duty cycle, an average beam power of 0.3 MW (maximum) has been assumed at the design beam energy of 100 GeV. Since a beam

*This improvement factor is based upon measurements made at W. W. Hansen Laboratories of Physics at Stanford University.

conversion efficiency of $\approx 100\%$ should be achieved in the superconducting accelerator, a total RF power of 0.3 MW average and 4.8 MW peak must be delivered to the accelerator.

The peak beam current at the 100 GeV design energy level can be obtained by dividing the peak beam power by the beam energy, i.e., $i = (4.8 \times 10^6) / (100 \times 10^9) = 48 \mu\text{A}$.

For an accelerator designed to produce 100 GeV with maximum transfer of RF power to the beam, the no-load (zero current) energy will be 200 GeV for the same value of input RF power. As the beam current is increased from zero, the beam energy for this design decreases linearly as $V = 200 [1 - (i/96)]$ GeV, and would become zero (provided beam break-up or other phenomena did not destroy the beam) at a beam current equal to twice the 48 μA design value.

For constant RF dissipation in the machine at a given beam energy V , the permissible duty cycle varies as V^{-2} . The maximum values of duty cycle, peak beam current and average beam current vs beam energy are shown in Fig. 2. The maximum values of duty cycle, peak beam power, and average beam power versus beam energy are shown in Fig. 3. While the maximum design average beam power is limited to 300 kW at the maximum energy level of 100 GeV, the beam power can be higher at lower energies. A maximum design average beam power of 2.1 MW can be obtained at a beam energy of 25 GeV, a peak beam current of 84 μA , and unity duty cycle. This is the highest average beam power which can be achieved with this machine without increasing either the level of RF power or the capacity of the refrigeration system. At energies above 100 GeV all the curves of Figs. 2 and 3 are shown dotted since 100 GeV is assumed to be the maximum energy achievable in this machine.

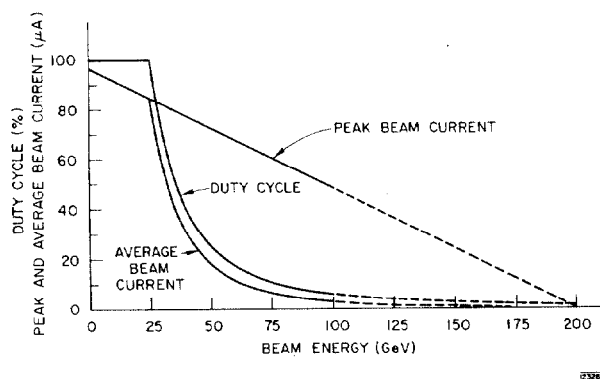


FIG. 2--Maximum values of duty cycle, peak beam current, and average beam current vs beam energy. The region above 100 GeV is expected to be beyond the operating limit.

Description of Components and Systems

In order to make the maximum use of the existing facilities, it is assumed that the superconducting accelerator will be placed in the present 10,000 foot-long tunnel. The present division of the accelerator and systems into 30 sectors is also expected to be retained.

Several of the components and systems of the accelerator are described below.

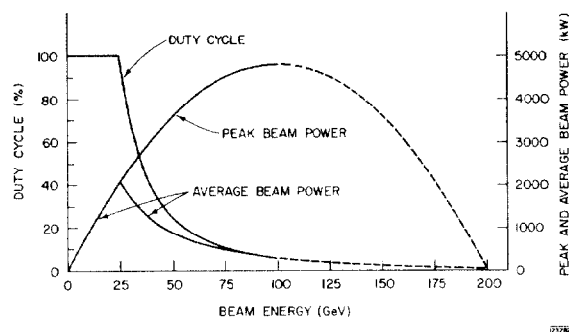


FIG. 3--Maximum values of duty cycle, peak beam power, and average beam power vs beam energy.

The Accelerating Structure

It is assumed that the accelerating structure will be a uniformly periodic disk-loaded waveguide. From the expression for beam energy given earlier it can be shown that the condition for maximum energy and efficiency is $2\tau g r_0 = P_s / i^2$. Then, $V = (2\tau g r_0 P_s)^{1/2}$ where the power dissipated, P_d , is given by $2\tau g P_s$ or $V^2 / r_0 \ell$. From the definition of $\tau = \omega \ell / 2v_g Q$ where v_g is the group velocity, one can also write $2\tau g = \omega \ell / v_g Q$ and the filling time $t_f = (2Q) / \omega (2\tau g) = 2g\ell / v_g$.

The combination of the above expressions and the properties of the superconducting materials set the goals of the design. Maximizing r_0 minimizes the power lost and the refrigeration requirements. Since $2\tau g$ appears everywhere as a product, it is seen that for given values of r_0 , Q , ω and ℓ , one can maintain $2\tau g$ constant by simply adjusting the ratio g/v_g . This means that within the range where r_0 and Q vary slowly, the group velocity can be chosen on the basis of other criteria and g can then be adjusted accordingly. The criteria for choosing v_g are that the ratios of peak-to-average E and H fields in the structure be minimized in order not to exceed the limits imposed by field emission and critical magnetic field.

The structure optimization program is proceeding on two fronts. On the one hand, two computer programs are presently being developed to study the dependences of r_0 , Q , E_{pk}/E_0 and H_{pk}/E_0 on cavity shape and v_g . One is a standing-wave calculation using the LALA program initiated by the Los Alamos group.⁴ The second is a traveling-wave calculation initiated at SLAC. On the other hand, there is an experimental program to verify some of the measurements and try out some new ideas.

As to the dependence of H_{pk}/E_0 , an estimate can be obtained from a simple TM_{01} waveguide. At cut-off, $H_{\phi\text{max}}/E_{z\text{max}} \approx 20$ gauss/MV/m. Since the fundamental space harmonic content, $a_0^2/\Sigma a_n^2$, is larger for the $\pi/2$ than for the $2\pi/3$ mode, it is possible that it will be easier to stay within the critical field limits by choosing the $\pi/2$ mode. Tentative gradients that may be reached with various structures and materials are shown in Table II.

TABLE II
Approximate Predictable Gradients for Various Modes and Materials
as Limited by Critical Magnetic Field. SLAC goal is 33 MeV/m.

Material at 1.85°K	B_c (max) (gauss)	Corresponding Maximum Gradients			
		TM_{01}, TW (MeV/m)	$TW^3, \pi/2$ (MeV/m)	$TW^4, 2\pi/3$ (MeV/m)	TM_{01}, SW^5 (MeV/m)
Lead	Theoretical: 500	~25	~21	~18	~12
	Achieved: ¹ ~400	~20	~17	~14	~10
Niobium	Theoretical: 1650	~82	~68	~59	~41
	Achieved: ² ~600	~30	~25	~21	~15
Technetium	Theoretical: 1350	~67	~56	~48	~34
	Achieved: Unknown				

¹ See for example, R.B. Wilson and H.A. Schwetman, Superconducting Accelerators, IEEE Transactions on Nuclear Science, NS-12, No. 3, (June 1965).

² J. Turneure, Private Communication, W. W. Hansen Laboratories, Stanford University, Stanford, California, (January 1969).

³ Based on assumption that space harmonic content in modes higher than fundamental is 20%.

⁴ Based on assumption that space harmonic content in modes higher than fundamental is 40%.

⁵ Based on assumption that $E_{TW}/E_{SW} = 2$ for the TM_{01} mode. This assumption results in a ratio of ~1.6 for E_{TW} for the $\pi/2$ mode to E_{SW} for the TM_{01} mode.

In parallel with this optimization program, studies are also proceeding on the susceptibility of these structures to beam break-up, RF transients, coupler asymmetry and the general problem of tolerances and mechanical deformations.

The Drive and Phasing System

This system can be divided conceptually into two sub-systems, one upstream and one downstream of the 20 kW klystron. The upstream part will hopefully require only minor departures from the present SLAC drive system.⁵ A more stable master oscillator at 476 MHz with a stability of $1:10^{10}$ will be needed because of the bandwidth imposed by the loaded Q ($\sim 10^8$) of the structure. The present drive lines will probably be retained and only the pulsed sub-booster amplifier will be eliminated (see Fig. 4). The varactors which multiply the 476 MHz frequency up to 1428 or 2856 MHz may be combined with simple solid state amplifiers and installed, one per sector or one per klystron.

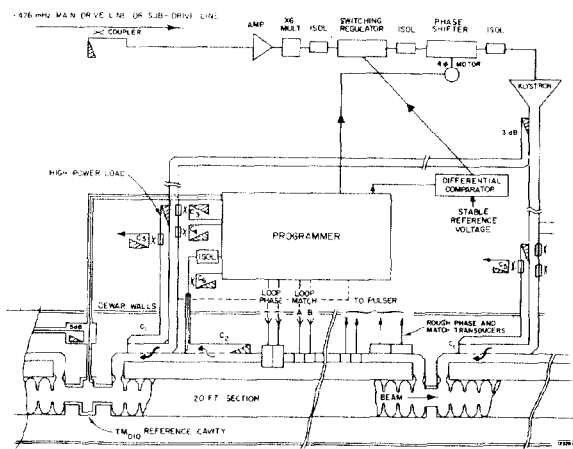


FIG. 4--Block diagram of the microwave systems (sample system shown to illustrate principles)

The klystron amplifiers presently being planned are rated at 20 kW output and are expected to be about 50% efficient. They use the same dc power supplies as the

present klystrons but do not use the pulsed modulators. The dc high voltage is kept on continuously during operation. Only the RF drive is gated on and off to vary the duty cycle. Physically, the tubes are of about the same size as the present 25 MW tubes because the average power rating is about the same. Each klystron will have its individual isolator, diode switching regulator, phase shifter and attenuator. The rectangular waveguide, downstream of the klystron, will remain as presently installed if the 2856 MHz frequency is retained.

From this point on, the system departs noticeably from the present SLAC system.^{6,7} Briefly, two automatic phasing systems and one impedance matching system are necessary. To minimize energy deviations, the first phasing system must keep the electric length around the loop an integral number of wavelengths to within 1 millidegree at 2856 MHz or 1/4 millidegree at 1428 MHz. The initial criterion for correct loop-phasing is the minimization of power flow into the high-power coupler load as monitored by coupler C₅.

Once the loop is adjusted, the output from a microwave bridge which compares the phase outputs of couplers C₂ and C₃ is nulled and any subsequent deviation is corrected by the programmer by means of the "loop-phase" transducers. The second phasing system utilizes an inline reference cavity which monitors beam phase. Again, the klystron is initially phased with respect to the beam and subsequently, a phase bridge using signals from C₃ and the cavity, makes corrections on the input phase shifter of the klystron by means of a four-phase motor.

The purpose of the automatic impedance-matching system is to maintain the voltage reflection coefficient in the feedback loop below 4×10^{-5} at 2856 MHz or below 1×10^{-5} at 1428 MHz. After an initial match is achieved, this function is performed by a third bridge which compares the phase outputs of couplers C₂ and C₄ and provides signals to the "loop-match" transducers.

Finally, there will be an RF monitoring system which will compare the RF amplitude in the ring as coupled at C₂ and compare it to a stable reference voltage through a differential comparator. Any deviation from a preset level will translate itself into a correcting signal applied to the switching diode regulator, as shown.

Dewar System

The helium dewar module will have the length of one sector, about 323 feet. It will be assembled in 20 foot-long sections. Each dewar section will contain one accelerator sections which will receive 1/2 the power from one klystron. The accelerator is the innermost of the three separate vacuum envelopes shown in Fig. 5. The midpoint of the 20 foot-long section is the axial anchor point and is also the location of all instrumentation and RF waveguide feed-throughs. This arrangement limits the effects of differential thermal contraction.

The second envelope shown in Fig. 5 is the helium dewar. The dewar is completely filled with liquid helium. The surface level of the liquid is in an elbow which extends a couple of feet above the accelerator axis. The

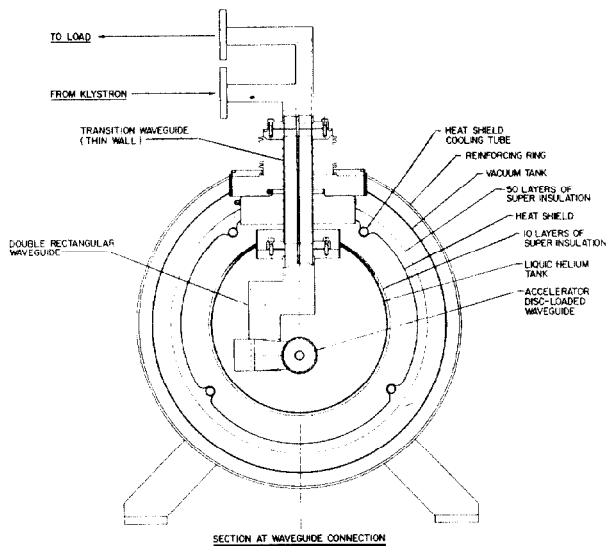


FIG. 5--Cross section of accelerator dewar system (sample system shown to illustrate principles)

heat transfer properties of He II permit the heat from a whole sector to be removed by pumping from a small area at one end of the sector. It is fortunate that this is the case because the accelerator tunnel is built with a slope amounting to 19 inches in the length of a sector. Thus the diameter of the dewar would become very large if a surface of evaporation has to be maintained over its entire length.

The shield between the helium dewar and the outer jacket is cooled by helium gas at about 80°K rather than by liquid nitrogen. The heat absorbed by the shield comes from radiation from the outer jacket and by conduction through waveguides and supports. This heat amounts to about 20 kW total for the smaller 2856 MHz design and about 25 kW at 1428 MHz. Even so, the power required to cool the shield amounts to only 2 to 4 percent of the power needed by the refrigeration system.

Refrigeration System

An example of a helium refrigeration system is shown in Fig. 6. The last stage of the refrigeration system, which is the liquefier, is located near the helium dewar. The refrigerator unit shown in Fig. 6 contains heat exchangers and turbine expanders to supply cold gas for two sectors. The total low temperature refrigeration load is estimated at 14 kW for the 2856 MHz accelerator. For 1428 MHz it is 8.7 kW. The compressors consume the largest fraction of the power. These units are concentrated at one location. Only warm gas is transported to and from the compressors. The vacuum pumps are located in the klystron gallery above the accelerator. These pumps are the next largest user of electrical power. The power consumption is shown in Table III for both proposed operating frequencies. The pumps compress the helium from 15 torr at the surface of the liquid helium to one atmosphere for transport to the compressors.

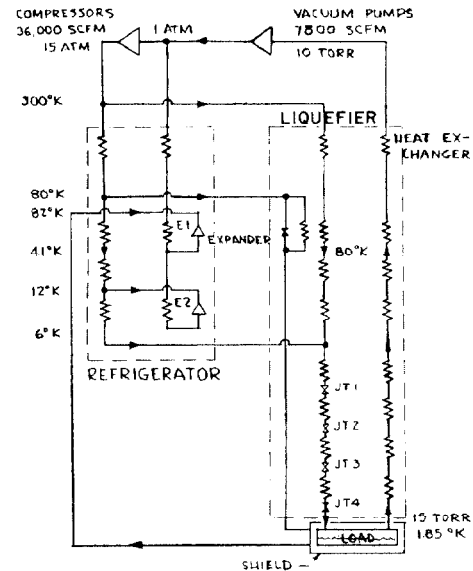


FIG. 6--Helium refrigeration system (sample system shown to illustrate principles)

Table III

MAJOR POWER REQUIREMENTS		
Frequency	2856 MHz	1428 MHz
Compressors	12×10^6 W	7.6×10^6 W
Vacuum pumps	10×10^6 W	6.4×10^6 W
Klystron power	12×10^6 W	12×10^6 W
Total	34×10^6 W	26×10^6 W

Acknowledgements

Many members of the SLAC staff have participated in the feasibility study of which only a part has been presented here. In particular, we would like to mention the work of Ron Cochran on the mechanical system and Gordon Ratliff and Steven St. Lorant on the refrigeration system. In addition, we are grateful for helpful discussions with the Stanford High Energy Physics Laboratory Group.

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