

MAGNETIC FIELD CALCULATIONS FOR A LARGE APERTURE NARROW QUADRUPOLE *

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Abstract

In the design of high-intensity proton synchrotrons and accumulator rings, quadrupole magnets of narrower size in one of the transverse dimensions are often needed to accommodate the compact ring geometry, the various injection and extraction devices, and the large vacuum chamber aperture. The stringent limit on tolerable beam loss further demands a good magnetic field quality to minimize beam resonances caused by higher-order magnetic multipoles.

In this paper, we present results from magnetic field calculations performed on 2D and 3D models of a large-aperture narrow-quadrupoles that is suitable for a high intensity, low beam-loss accumulator rings. The pole face of the quadrupole has been optimized to minimize the integrated field of the first three allowed multipoles (12pole, 20pole and 28pole). The ratio of each integrated magnetic-multipole-strength to the integrated magnetic-quadrupole-strength at a radius of 85% of the quad's pole-tip-radius is less than 2×10^{-4} . Results from the calculations performed on the two-dimensional and three-dimensional models of the narrow quad are presented.

1 INTRODUCTION

In a published paper[1] we provide detailed information about the design of a "narrow quadrupole" that has been built to be used in the SNS accumulator ring[2]. The transverse dimension of the quadrupole on the horizontal plane has helped accommodate the various devices which are located at the injection and extraction regions[1] of the SNS accumulator ring. The main requirements for the magnetic design of the narrow quadrupole was to minimize the integrated strength of the 12pole magnetic multipole down to $\int B_{12pole} dz / \int B_{4pole} dz < 2 \times 10^{-4}$ at a radius of $r=10$ cm, and we did not regard the contributions from the 20pole and 28pole multipoles. Table I shows the integrated strength of these multipoles relative to the integrated quadrupole strength as were calculated by the 3-Dimensional model and also as were measured in the first built quadrupole.

In Table 1, $R_n = \int B_n(r,z) dz / \int B_1(r,z) dz$ where $B_n(r,z)$ is the n^{th} coefficient in the expansion of the radial field $B_r(r,z)$ $B_r(r,z) = \sum B_n(r,z) \cos[(n+1)\theta]$ { $n=1$ quad, $n=3$ oct ... } (1) The $B_r(r,z)$ is calculated and measured at a radius $r=10$ cm and the coefficients $B_n(r,z)$ are integrated along the z-axis which is along the beam direction.

In Table 1, the 2nd row shows the calculated quantity R_n , and the 3rd and 4th rows the measured quantity R_n for the quadrupole as it was delivered by the manufacturer

(3rd row) and with a minor modification (reduce the transverse pole length by 0.75 mm), (4th row). The modification resulted in reduction of the $R_{(12pole)}$ multipole strength down to 1.2×10^{-4} . The error in the measurements of the harmonics was $\pm 0.02 \times 10^{-4}$

Table 1: The Ratio R_n of few allowed multipoles. The quantity R_n is defined in the text.

	$R_5 (12pole)$	$R_9 (20pole)$	$R_{13} (28pole)$
CALC	-4.2×10^{-4}	-5.2×10^{-4}	-1.5×10^{-4}
MEAS#1	$+3.20 \times 10^{-4}$	-6.90×10^{-4}	-1.20×10^{-4}
MEAS#2	$+1.20 \times 10^{-4}$	-6.81×10^{-4}	-0.92×10^{-4}

Calculations performed on the SNS ring [3] showed that the measured strength of the (20 and 28)poles multipoles (see Table 1) of the narrow quadrupole are well below the limits that may bring the beam into resonance and cause significant beam emittance growth that will result in beam losses. Nevertheless we thought as a useful task to design a narrow quadrupole that minimizes the first three allowed multipoles (12,20,28)pole. The following sections are dealing with the design of such a narrow quadrupole.

2 THEORY FOR THE MAGNET DESIGN

Poisson's theorem states that the magnetic field vector \mathbf{B} (or any vector, regular at infinity) can be expressed as:

$$\mathbf{B}(\mathbf{x}) = (1/4\pi) \int \{ [\nabla'(\nabla' \cdot \mathbf{B}) - \nabla' \times (\nabla' \times \mathbf{B})] / |\mathbf{x} - \mathbf{x}'| \} d^3 \mathbf{x}' \quad (1)$$

By defining the magnetization vector

$$\mathbf{M} = \mathbf{B} - \mu_0 \mathbf{H} \quad (2)$$

and using the Maxwell equations $\nabla' \cdot \mathbf{B} = 0$ and $\nabla' \times \mathbf{H} = \mathbf{J}$ equation (1) becomes:

$$\mathbf{B}(\mathbf{x}) = (1/4\pi) \left\{ \int [\mu_0 \nabla' \times \mathbf{J} + \nabla' \times (\nabla' \times \mathbf{M})] / |\mathbf{x} - \mathbf{x}'| \right\} d^3 \mathbf{x}' \quad (3)$$

Equation (3) expresses the magnetostatic field $\mathbf{B}(\mathbf{x})$ as the contribution of two terms; one term corresponding to the currents distribution \mathbf{J} the other term to the magnetization \mathbf{M} of the materials. With $\mathbf{J}_M = (1/\mu_0)(\nabla' \times \mathbf{M})$ equation (3) can also be written (see ref. [4]) as:

$$\mathbf{B}(\mathbf{x}) = (\mu_0/4\pi) \left[\int \nabla' \times \mathbf{J} / |\mathbf{x} - \mathbf{x}'| d^3 \mathbf{x}' + \int \mathbf{J}_M \times \mathbf{r}_{ur} / |\mathbf{x} - \mathbf{x}'|^2 d^3 \mathbf{x}' + \int (\mathbf{M} \times \mathbf{n}_{out}) \times \mathbf{r}_{ur} / |\mathbf{x} - \mathbf{x}'|^2 d^2 \mathbf{x}' \right] \quad (4)$$

In equation (4) the second integral extends over the interior of the finite volume of the magnetic material, and the third integral over the surface enclosing the volume of the magnetic material (\mathbf{n}_{out} =normal to the surface). It is the contribution of the third integral that can affect the strength of the various allowed multipoles by altering the contour of the pole face. We assume that the value of the permeability μ of the iron at the vicinity of the pole surface has a value $\mu \gg 1$ for the third integral to have an effect on the magnetic multipoles. It is therefore possible to affect the magnetic field in the space of the beam by

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modifying the contour of the pole tip of the quadrupole. This approach [5] was followed in the design of narrow quadrupoles. In this paper we employ more sophisticated contour of the pole tip and we extend the calculations in three dimensions.

3 TWO-DIMENSIONAL MODELING

In this section we present the results of the two dimensional magnetic calculations as applied to three designs of the narrow quadrupole. The designs will be referred in the text as A,B and C. In each of the designs, discussed below, we provide enough information about the cross section of the narrow quadrupoles, for the reader to repeat the calculations using her/his preferred computer code for the electromagnetic design. In all three designs the strength of all allowed multipoles $B_{(12pole,20pole,28pole)}$ was calculated at $r=10cm$ and the $B_{(12pole,20pole,28pole)}/B_{quad}$ ratio was reduced below the value of 1×10^{-4} . The deviation of the narrow quad from the four fold symmetry, introduced multipoles like octupoles, 16poles etc. However the relative strength of each of the multipoles ($B_{(8pole,16pole,24pole)}/B_{quad}$) was below the value of 1×10^{-5} at $r=10 cm$. In each of the designs we kept intact both, the outside dimensions of the quadrupole (shown in Fig. 1) and the pole tip radius R_p and we only varied the width of the pole piece W , and the pole tip contour.

In order to keep the permeability of the iron at a reasonable large value $\mu \gg 1$, the quadrupole strength of each of the models was also kept at a value of ~ 4.2 [T/m]. All calculations were performed using the computer code for electromagnetics of Vector_Fields[6].

3.1 Narrow Quadrupole Design_A

The cross section of one of the pole pieces of the design_A quadrupole is shown in Fig 2. In this design we kept the contour of the pole tip similar to the contour of the narrow quad discussed in ref [1] but we increased the pole width (W) to a value of 19.8 cm, to achieve minimization of the 20pole and 28pole multipoles. The design was finally optimized by modifying the contour of the pole face, by varying the radii of curvature ρ_i, ρ_o and the location of the inflection points P_1, P_2 , shown in Fig 2. The optimization yielded a ratio B_n/B_{quad} of $< 1 \times 10^{-4}$ at a radius $r=10 cm$ for the (12, 20, 28)pole multipoles.

The increase of the pole width (W) however reduced the area of the current conductor which has to run at a higher current density (J) to achieve the quadrupole strength of ~ 4.2 [T/m]. An alternative design which satisfies the requirements of low relative strength $B_{npole}/B_{4pole} < 1 \times 10^{-4}$, for the (12, 20, 28)pole multipoles, and also provides more conductor area, is discussed in the next subsection.

3.2 Narrow Quadrupole Design_B

The cross section of this alternative design of a narrow quadrupole is shown in Fig. 3. In this design the width of the pole piece has been reduced to 17.6 cm but the overall

shape of the pole tip surface remained almost the same as in design_A with only small modifications of the location of the inflection points P_1, P_2 and radii of curvature ρ_i, ρ_o . These minor modifications reduced the relative strength B_{npole}/B_{4pole} of the (12, 20, 28)pole multipoles, below the required value of 1×10^{-4} at a radius $r=10 cm$.

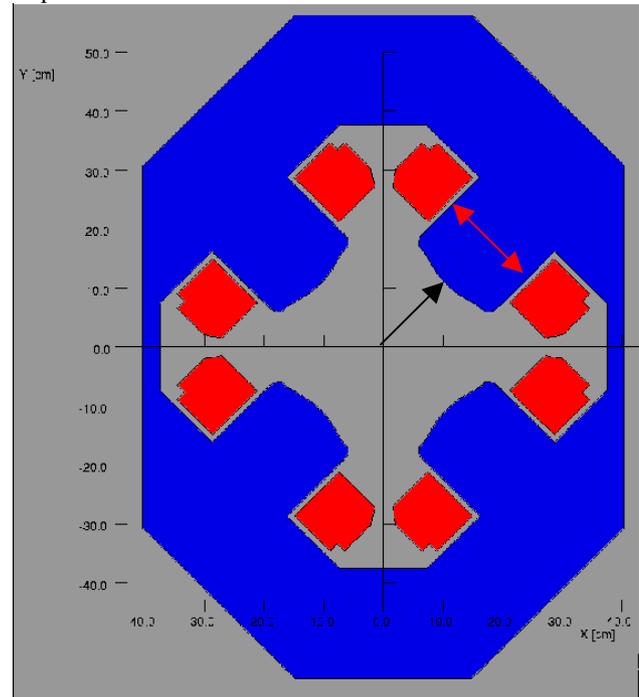


Figure 1: Cross section of the narrow quad. The outer dimensions were the same for all designs A,B, and C.

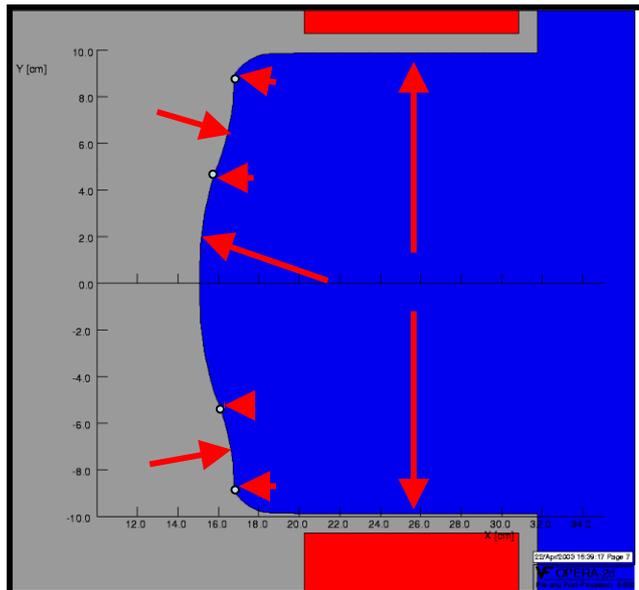


Figure 2. Cross section of pole piece corresponding to "design A" (see text). The inflection points P_1, P_2 , and the radii of curvature ρ_i, ρ_o were varied in order to minimize the strength of the (12,20,28)pole multipoles.

Compared with design A, this design allows for an increased copper area and the required gradient of ~ 4.2 [T/m] is achieved at a reduced current density, but the magnetic field \mathbf{B} inside the poles will be higher.

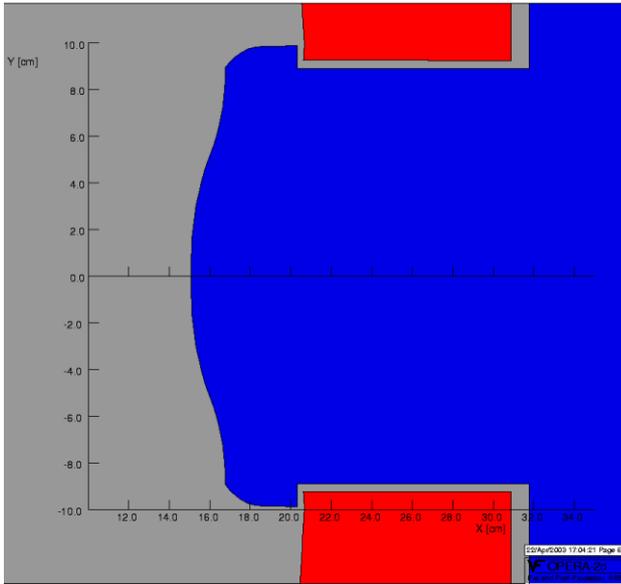


Figure 3. Cross section of pole piece for “design B” .

3.3 Narrow Quadrupole Design_C

This design combines the features of the design_A and design_B namely larger conductor area (same as in design_B) and lower value of the magnetic flux density \mathbf{B} in the pole pieces (as in design_A). The cross section is shown in figure 4. The relative strength B_{npole}/B_{4pole} of the (12,20,28)pole is minimized to values $< 1 \times 10^{-4}$. Compare the contour shape of design_C with that of designs_A or B.

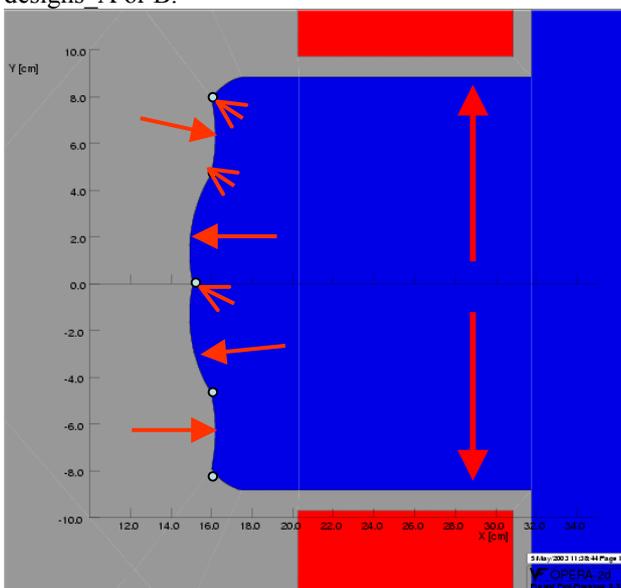


Figure 4. Cross section of the pole piece of “design_C”.

4 THREE-DIMENSIONAL MODELING

Practical considerations lead us to perform the three dimensional magnetic field calculations on the “design_C”. The goal was to minimize the relative integrated strength $\int B_{npole} dz / \int B_{4pole} dz$ of the (12,20,28)pole multipoles. The method of optimization was to chamfer the edges of the pole pieces at both, the entrance and exit of the magnet [1] as shown in figure 6. The “pole chamfering” reduced the integrated strength of the 12pole multipole but introduced some strength in the 20pole and 28pole multipoles. This strength was reduced by reshaping slightly the contour of the pole tip inside the magnet. The optimization yielded the following results:
 $\int B_{12pole} dz / \int B_{4pole} dz = 2 \times 10^{-6}$ $\int B_{20pole} dz / \int B_{4pole} dz = 4 \times 10^{-5}$
 $\int B_{28pole} dz / \int B_{4pole} dz = 5 \times 10^{-5}$ at $r=10$ cm.

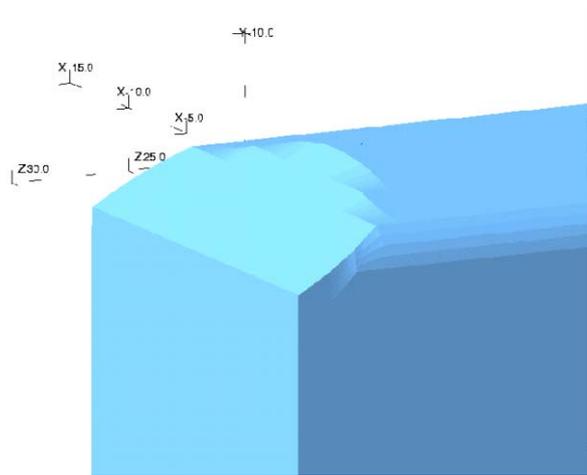


Figure 5 An isometric view of the chamfered ends of one of the pole-piece of the narrow quadrupole.

5 CONCLUSIONS

Two dimensional magnetic field calculations were performed on three models of a large aperture narrow quadrupole. Each of the models was optimized to minimize the relative strength B_{npole}/B_{4pole} of the (12, 20, 28)pole multipoles to values less than 1×10^{-4} at a radius $r=10$ cm. One of the models was optimized using 3D magnetic field calculations.

6 REFERENCES

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